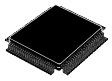
32-bit Arm® Cortex®-M7 280 MHz MCUs, 128-Kbyte Flash memory, 1.4-Mbyte RAM, 46 com. and analog interfaces, SMPS, crypto



LQFP64
(10 x 10 mm)
LQFP100
(14 x 14 mm)
LQFP144
(20x20 mm)
LQFP176
(24 x 24 mm)



UFBGA169
(7 x 7 mm)
UFBGA176+25
(10x10 mm)

Product summary	
STM32H7B0xB	STM32H7B0AB, STM32H7B0IB, STM32H7B0RB, STM32H7B0ZB, STM32H7B0VB

Features

Includes ST state-of-the-art patented technology

Core

- 32-bit Arm® Cortex®-M7 core with double-precision FPU and L1 cache: 16 Kbytes of data and 16 Kbytes of instruction cache allowing to fill one cache line in a single access from the 128-bit embedded flash memory; frequency up to 280 MHz, MPU, 599 DMIPS/ 2.14 DMIPS/MHz (Dhrystone 2.1), and DSP instructions

Memories

- 128 Kbytes of flash memory plus 1 Kbyte of OTP memory
- ~1.4 Mbytes of RAM: 192 Kbytes of TCM RAM (inc. 64 Kbytes of ITCM RAM + 128 Kbytes of DTCM RAM for time critical routines), 1.18 Mbytes of user SRAM, and 4 Kbytes of SRAM in Backup domain
- 2x Octo-SPI memory interfaces with on-the-fly decryption, I/O multiplexing and support for serial PSRAM/NOR, Hyper RAM/flash frame formats, running up to 140 MHz in SRD mode and up to 110 MHz in DTR mode
- Flexible external memory controller with up to 32-bit data bus:
  - SRAM, PSRAM, NOR flash memory clocked up to 125 MHz in Synchronous mode
  - SDRAM/LPSDR SDRAM
  - 8/16-bit NAND flash memories
- CRC calculation unit

Security

- ROP, PC-ROP, active tamper, secure firmware upgrade support, Secure access mode

General-purpose input/outputs

- Up to 138 I/O ports with interrupt capability
  - Fast I/Os capable of up to 133 MHz
  - Up to 164 5-V-tolerant I/Os

Low-power consumption

- Stop: down to 32 µA with full RAM retention
- Standby: 2.8 µA (Backup SRAM OFF, RTC/LSE ON, PDR OFF)
- V_{BAT}: 0.8 µA (RTC and LSE ON)

Clock management

- Internal oscillators: 64 MHz HSI, 48 MHz HSI48, 4 MHz CSI, 32 kHz LSI
- External oscillators: 4-50 MHz HSE, 32.768 kHz LSE
- 3× PLLs (1 for the system clock, 2 for kernel clocks) with fractional mode

Reset and power management

- 2 separate power domains, which can be independently clock gated to maximize power efficiency:
  - CPU domain (CD) for Arm® Cortex® core and its peripherals, which can be independently switched in Retention mode
  - Smart run domain (SRD) for reset and clock control, power management and some peripherals
- 1.62 to 3.6 V application supply and I/Os
- POR, PDR, PVD and BOR
- Dedicated USB power embedding a 3.3 V internal regulator to supply the internal PHYs
- Dedicated SDMMC power supply
- High power efficiency SMPS step-down converter regulator to directly supply V_{CORE} or an external circuitry
- Embedded regulator (LDO) with configurable scalable output to supply the digital circuitry
- Voltage scaling in Run and Stop mode
- Backup regulator (~0.9 V)
- Low-power modes: Sleep, Stop and Standby
- V_{BAT} battery operating mode with charging capability
- CPU and domain power state monitoring pins

Interconnect matrix

- 3 bus matrices (1 AXI and 2 AHB)
- Bridges (5× AHB2APB, 3× AXI2AHB)

5 DMA controllers to unload the CPU

- 1× high-speed general-purpose master direct memory access controller (MDMA)
- 2× dual-port DMAs with FIFO and request router capabilities
- 1× basic DMA with request router capabilities
- 1x basic DMA dedicated to DFSDM

Up to 35 communication peripherals

- 4× I2C FM+ interfaces (SMBus/PMBus)
- 5× USART/5x UARTs (ISO7816 interface, LIN, IrDA, modem control) and 1x LPUART
- 6× SPIs, including 4 with muxed full-duplex I2S audio class accuracy via internal audio PLL or external clock and 1 x SPI/I2S in LP domain (up to 125 MHz)
- 2x SAIs (serial audio interface)
- SPDIFRX interface
- SWPMI single-wire protocol master interface
- MDIO Slave interface
- 2× SD/SDIO/MMC interfaces (up to 133 MHz)
- 2× CAN controllers: 2 with CAN FD, 1 with time-triggered CAN (TT-CAN)
- 1× USB OTG interfaces (1HS/FS)
- HDMI-CEC
- 8- to 14-bit camera interface up to 80 MHz
- 8-/16-bit parallel synchronous data input/output slave interface (PSSI)

11 analog peripherals

- 2× ADCs with 16-bit max. resolution (up to 24 channels, up to 3.6 MSPS)
- 1× analog and 1x digital temperature sensors
- 1× 12-bit single-channel DAC (in SRD domain) + 1× 12-bit dual-channel DAC
- 2× ultra-low-power comparators
- 2× operational amplifiers (8 MHz bandwidth)
- 2× digital filters for sigma delta modulator (DFSDM), 1x with 8 channels/8 filters and 1x in SRD domain with 2 channels/1 filter

Graphics

- LCD-TFT controller up to XGA resolution
- Chrom-ART graphical hardware Accelerator (DMA2D) to reduce CPU load
- Hardware JPEG Codec
- Chrom-GRC™ (GFXMMU)

Up to 19 timers and 2 watchdogs

- 2× 32-bit timers with up to 4 IC/OC/PWM or pulse counter and quadrature (incremental) encoder input (up to 280 MHz)
- 2× 16-bit advanced motor control timers (up to 280 MHz)
- 10× 16-bit general-purpose timers (up to 280 MHz)
- 3× 16-bit low-power timers (up to 280 MHz)
- 2× watchdogs (independent and window)
- 1× SysTick timer
- RTC with sub-second accuracy and hardware calendar

Cryptographic acceleration

- AES chaining modes: ECB,CBC,CTR,GCM,CCM for 128, 192 or 256
- HASH (MD5, SHA-1, SHA-2), HMAC
- 2x OTFDEC AES-128 in CTR mode for Octo-SPI memory encryption/decryption
- 1x 32-bit, NIST SP 800-90B compliant, true random generator

Debug mode

- SWD and JTAG interfaces
- 4 KB Embedded Trace Buffer

96-bit unique ID

All packages are ECOPACK2 compliant

1 Introduction

This datasheet provides the ordering information and mechanical device characteristics of the STM32H7B0xB microcontrollers.

This document should be read in conjunction with the STM32H7B0xB reference manual (RM0455). The reference manual is available from the STMicroelectronics website .

For information on the device errata with respect to the datasheet and reference manual, refer to the STM32H7B0xB errata sheet (ES0478), available on the STMicroelectronics website .

For information on the Arm® Cortex®-M7 core, refer to the Cortex®-M7 Technical Reference Manual, available from the www.arm.com website

Note: Arm is a registered trademark of Arm Limited (or its subsidiaries) in the US and/or elsewhere.

2 Description

STM32H7B0xB devices are based on the high-performance Arm® Cortex®-M7 32-bit RISC core operating at up to 280 MHz. The Cortex® -M7 core features a floating point unit (FPU) which supports Arm® double-precision (IEEE 754 compliant) and single-precision data-processing instructions and data types. STM32H7B0xB devices support a full set of DSP instructions and a memory protection unit (MPU) to enhance application security.

STM32H7B0xB devices incorporate high-speed embedded memories with a flash memory of 128 Kbytes, around 1.4 Mbyte of RAM (including 192 Kbytes of TCM RAM, 1.18 Mbytes of user SRAM and 4 Kbytes of backup SRAM), as well as an extensive range of enhanced I/Os and peripherals connected to four APB buses, three AHB buses, a 32-bit multi-AHB bus matrix and a multi layer AXI interconnect supporting internal and external memory access.

All the devices offer two ADCs, two DACs (one dual and one single DAC), two ultra-low power comparators, a low-power RTC, 12 general-purpose 16-bit timers, two PWM timers for motor control, three low-power timers, a true random number generator (RNG), and a cryptographic acceleration cell and a HASH processor. The devices support nine digital filters for external sigma delta modulators (DFSDM). They also feature standard and advanced communication interfaces.

- Standard peripherals
  - Four I2Cs
  - Five USARTs, five UARTs and one LPUART
  - Six SPIs, four I2Ss in full-duplex mode. To achieve audio class accuracy, the I^2S peripherals can be clocked via a dedicated internal audio PLL or via an external clock to allow synchronization.
  - Two SAI serial audio interfaces, out of which one with PDM
  - One SPDIFRX interface
  - One single wire protocol master interface (SWPMI)
  - One 16-bit parallel synchronous slave interface (PSSI) sharing the same interface as the digital camera)
  - Management Data Input/Output (MDIO) slaves
  - Two SDMMC interfaces (one can be supplied from a supply voltage separate from that of all other I/Os)
  - A USB OTG high-speed with full-speed capability (with the ULPI)
  - One FDCAN plus one TT-CAN interface
  - Chrom-ART Accelerator
  - HDMI-CEC
- Advanced peripherals including
  - A flexible memory control (FMC) interface
  - Two octo-SPI memory interface with on-the-fly decryption (OTFDEC)
  - A digital camera interface for CMOS sensors (DCMI)
  - A graphic memory management unit (GFXMMU)
  - An LCD-TFT display controller (LTDC)
  - A JPEG hardware compressor/decompressor

Refer to Table 1. STM32H7B0xB features and peripheral counts for the list of peripherals available on each part number.

STM32H7B0xB devices operate in the –40 to +85 °C ambient temperature range from a 1.62 to 3.6 V power supply. The supply voltage can drop down to 1.62 V by using an external power supervisor (see Section 3.5.2 Power supply supervisor) and connecting the PDR_ON pin to V_{SS}. Otherwise the supply voltage must stay above 1.71 V with the embedded power voltage detector enabled.

The USB OTG_HS/FS interfaces can be supplied either by the integrated USB regulator or through a separate supply input.

A dedicated supply input is available for one of the SDMMC interface for package with more than 100 pins. It allows running from a different voltage level than all other I/Os.

A comprehensive set of power-saving mode allows the design of low-power applications.

The CPU and domain states can be directly monitored on some GPIOs configured as alternate functions.

STM32H7B0xB devices are offers in several packages ranging from 64 pins to 225 pins/balls. The set of included peripherals changes with the device chosen.

These features make the STM32H7B0xB microcontrollers suitable for a wide range of applications:

- Motor drive and application control
- Medical equipment
- Industrial applications: PLC, inverters, circuit breakers
- Printers, and scanners
- Alarm systems, video intercom, and HVAC
- Home audio appliances
- Mobile applications, Internet of Things
- Wearable devices: smart watches.

Figure 1. STM32H7B0xB block diagram shows the general block diagram of the device family.

Table 1. STM32H7B0xB features and peripheral counts

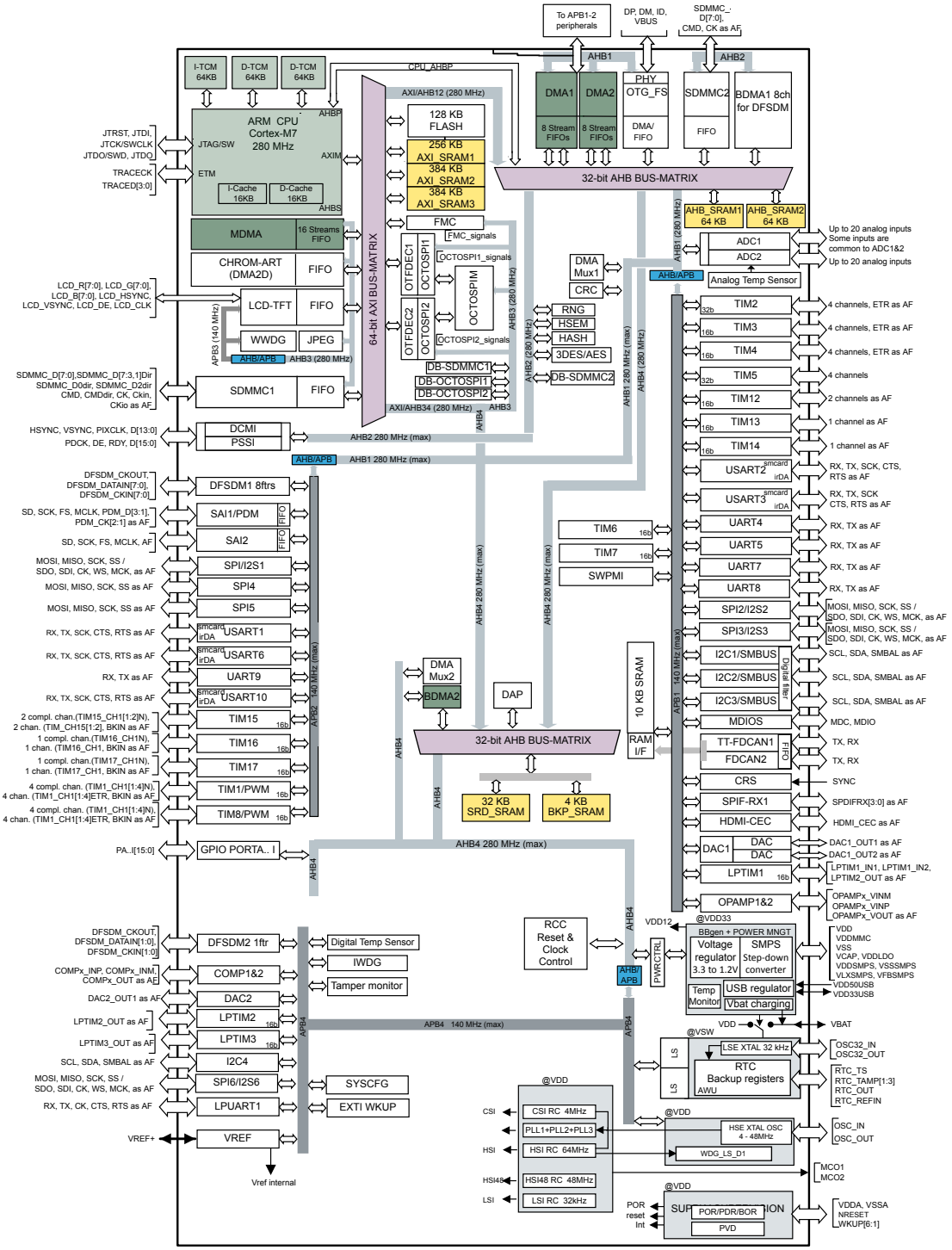
Peripherals		SMPS [1]		no-SMPS			
		STM32H7B0IBK	STM32H7B0ABI	STM32H7B0IBT	STM32H7B0ZBT	STM32H7B0VBT	STM32H7B0RBT
Flash memory (Kbytes)		128					
SRAM in Kbytes	SRAM on AXI	1024					
	SRAM on AHB (CD domain)	128					
	SRAM on AHB (SRD domain)	32					
TCM RAM in Kbytes	ITCM RAM (instruction)	64					
	DTCM RAM (data)	128					
Backup SRAM (Kbytes)		4					
FMC	Interface	1					
	NOR Flash memory/RAM controller	x [2]		x	x[2]	-	
	Multiplexed I/O NOR Flash memory	x		x		x[2]	-
	16-bit NAND Flash memory	x		x		x[2]	-
	SDRAM controller	x[2]		x	x[2]	-	
Octo-SPI interfaces[3]		2[4]		2	2[4]	1 Octo-SPI	1 Quad-SPI
Timers	General-purpose	10					
	Advanced-control (PWM)	2					
	Basic	2					
	Low-power	3					
Window watchdog / independent watchdog		1/1					
Real-time Clock (RTC)		1					
Tamper pins [5]	Passive	2		3	2		
	Active	1		2	1		
Random number generator		1					
Cryptographic accelerator		1					
Hash processor (HASH)		1					
On-the-fly decryption	for external Octo-SPI memory	2		2		2	2[2]

Peripherals		SMPS [1]		no-SMPS			
		STM32H7B0IBK	STM32H7B0ABI	STM32H7B0IBT	STM32H7B0ZBT	STM32H7B0VBT	STM32H7B0RBT
Communi-cation interfaces	SPI/I2S [6]	6/4		6/4		5/4	4/4
	I2C	4					3
	USART/UART /LPUART	5/5 /1		5/5 /1		5[2]/5 /1	4[2]/3[2] /1
	SAI/PDM	2/1		2/1		2[2]/1	1[2]/-
	SPDIFRX	4 inputs		4 inputs			
	SWPMI	1					
	MDIOS	1					
	SDMMC	2		2		2[7]	
	FDCAN/TT-CAN	1/1					1/1[2]
	USB OTG_HS ULPI, OTG_FS PHY	1	1	1	1 [8]		1 [9]
Digital camera interface/PSSI [10]		1/1					1/1
LCD-TFT display controller		1					
JPEG Codec		1					
Chrom-ART Accelerator (DMA2D)		1					
Graphic memory management unit (GFXMMU)		1					
HDMI CEC		1					
DFSDM		2					
Number of filters for DFSDM1/DFSDM2		8/1		8/1			7/1
ADCs	8 to 16 bits	2					
	Number of channels	24	24	20[11]		16[11]	
DACs	12 bits	2					
	Number of channels	3 (1 single channel + 1 dual-channel interfaces)					
Comparators		2		2			1
Operational amplifier		2		2			1
GPIOs		128	121	138	112	80	49
	Wakeup pins	4		6	4		

Peripherals	SMPS [1]		no-SMPS			
	STM32H7B0IBK	STM32H7B0ABI	STM32H7B0IBT	STM32H7B0ZBT	STM32H7B0VBT	STM32H7B0RBT
Maximum CPU frequency (MHz)	280					
SMPS step-down converter	1		-			
USB internal regulator	1		-			
USB separate supply pad	1				-	
VDDMMC separate supply pad	1				-	
VREF+ separate pad and internal buffer	1		1		1	-
Operating voltage	1.62 to 3.6 V [12]					
Operating temperatures	Ambient temperature range: −40 to 85 °C					
	Junction temperature range: −40 to 130 °C[13]					
Packages	UFBGA176+25	UFBGA169	LQFP176	LQFP144	LQFP100	LQFP64
Bootloader	USART, I2C, SPI, USB-DFU, FDCAN	USART, I2C, SPI, USB-DFU, FDCAN	USART, I2C, SPI, USB-DFU, FDCAN	USART, I2C, SPI, USB-DFU		

1. *The devices with SMPS correspond to commercial code STM32H7B0xIxxQ.*
2. *For limitations on peripheral features depending on packages, check the available pins/balls in Table 7. STM32H7B0xB pin/ball definition.*
3. *To maximize the performance, the I/O high-speed at low-voltage feature (HSLV) must be activated when V_{DD} < 2.7 V. This feature is not available on all I/Os (see Table 88. OCTOSPI characteristics in SDR mode, and Table 89. OCTOSPI characteristics in DTR mode (with DQS)/Octal and Hyperbus).*
4. *The I/O high-speed at low-voltage feature (HSLV) at V_{DD} < 2.7 V is not available for OCTOSPIM_P2.*
5. *A tamper pin can be configured either as passive or active (not both).*
6. *SPI1, SPI2, SPI3 and SPI6 interfaces give the flexibility to work in an exclusive way in either SPI mode or I2S audio mode.*
7. *Dedicated I/O supply pad (VDDMMC) or external level shifter are not supported.*
8. *The ULPI interface is supported. PC2 and PC3 are available on PC2_C and PC3_C, respectively, by closing the internal analog switch (see Table 7. STM32H7B0xB pin/ball definition).*
9. *The ULPI interface is not supported.*
10. *DCMI and PSSI cannot be used simultaneously since they share the same circuitry.*
11. *For limitations on fast pads or channels depending on packages, check to the available pins/balls in Table 7. STM32H7B0xB pin/ball definition.*
12. *V_{DD}/V_{DDA} can drop down to 1.62 V by using an external power supervisor (see Section 3.5.2 Power supply supervisor) and connecting PDR_ON pin to V_{SS}. Otherwise the supply voltage must stay above 1.71 V with the embedded power voltage detector enabled.*
13. *The junction temperature is limited to 105 °C in VOS0 voltage range.*

Figure 1. **STM32H7B0xB block diagram**



3 Functional overview

3.1 Arm® Cortex®-M7 with FPU

The Arm® Cortex®-M7 with double-precision FPU processor is the latest generation of Arm processors for embedded systems. It was developed to provide a low-cost platform that meets the needs of MCU implementation, with a reduced pin count and optimized power consumption, while delivering outstanding computational performance and low interrupt latency.

The Cortex®-M7 processor is a highly efficient high-performance featuring:

- Six-stage dual-issue pipeline
- Dynamic branch prediction
- Harvard architecture with L1 caches (16 Kbytes of I-cache and 16 Kbytes of D-cache)
- 64-bit AXI4 interface
- 64-bit ITCM interface
- 2x32-bit DTCM interfaces

The following memory interfaces are supported:

- Separate Instruction and Data buses (Harvard Architecture) to optimize CPU latency
- Tightly Coupled Memory (TCM) interface designed for fast and deterministic SRAM accesses
- AXI Bus interface to optimize Burst transfers
- Dedicated low-latency AHB-Lite peripheral bus (AHBP) to connect to peripherals.

The processor supports a set of DSP instructions which allow efficient signal processing and complex algorithm execution.

It also supports single and double precision FPU (floating point unit) speeds up software development by using metalanguage development tools, while avoiding saturation.

Refer to Figure 1. STM32H7B0xB block diagram for the general block diagram of the STM32H7B0xB family.

Note: Cortex®-M7 with FPU core is binary compatible with the Cortex®-M4 core.

3.2 Memory protection unit (MPU)

The memory protection unit (MPU) manages the CPU access rights and the attributes of the system resources. It has to be programmed and enabled before use. Its main purposes are to prevent an untrusted user program to accidentally corrupt data used by the OS and/or by a privileged task, but also to protect data processes or read-protect memory regions.

The MPU defines access rules for privileged accesses and user program accesses. It allows defining up to 16 protected regions that can in turn be divided into up to 8 independent subregions, where region address, size, and attributes can be configured. The protection area ranges from 32 bytes to 4 Gbytes of addressable memory.

When an unauthorized access is performed, a memory management exception is generated.

3.3 Memories

3.3.1 Embedded flash memory

The STM32H7B0xB devices embed up to 128 Kbytes of flash memory that can be used for storing programs and data.

The flash memory is organized as 137-bit flash words memory that can be used for storing both code and data constants. Each word consists of:

- One flash word (4 words, 16 bytes or 128 bits)
- 9 ECC bits.

The Flash memory is organized as follows:

- 128 Kbytes of user Flash memory, containing 16 user sectors of 8 Kbytes each
- 128 Kbytes of System Flash memory from which the device can boot.
- 1 Kbyte of OTP (one-time programmable) memory containing option bytes for user configuration.

3.3.2 Secure access mode

In addition to other typical memory protection mechanism (RDP, PCROP), STM32H7B0xB devices embed the Secure access mode, an enhanced security feature. This mode allows developing user-defined secure services by ensuring, on the one hand code and data protection and on the other hand code safe execution.

Two types of secure services are available:

- STMicroelectronics Root Secure Services:

These services are embedded in System memory. They provide a secure solution for firmware and third-party modules installation. These services rely on cryptographic algorithms based on a device unique private key.

- User-defined secure services:

These services are embedded in user flash memory. Examples of user secure services are proprietary user firmware update solution, secure flash integrity check or any other sensitive applications that require a high level of protection.

The secure firmware is embedded in specific user flash memory areas configured through option bytes.

Secure services are executed just after a reset and preempt all other applications to guarantee protected and safe execution. Once executed, the corresponding code and data are no more accessible.

The above secure services are available only for Cortex®-M7 core operating in Secure access mode. The other masters cannot access the option bytes involved in Secure access mode settings or the flash secured areas.

3.3.3 Embedded SRAM

All devices feature:

- 1 Mbyte of AXI-SRAM mapped onto AXI bus matrix in CPU domain (CD) split into:
  - AXI-SRAM1: 256 Kbytes
  - AXI-SRAM2: 384 Kbytes
  - AXI-SRAM3: 384 Kbytes
- 128 Kbytes of AHB-RAM mapped onto AHB bus matrix in CPU domain (CD) split into:
  - AHB-SRAM1: 64 Kbytes
  - AHB-SRAM2: 64 Kbytes
- 32 Kbytes of SRD-SRAM mapped in Smart Run Domain (SRD)
- 4 Kbytes of backup SRAM

The content of this area is protected against possible unwanted write accesses, and is retained in Standby or V_{BAT} mode.

- RAM mapped to TCM interface (ITCM and DTCM):

Both ITCM and DTCM RAMs are 0 wait state memories that are accessible from the CPU or the MDMA (even in Sleep mode) through a specific AHB slave of the CPU(AHBP).

- 64 Kbytes of ITCM-RAM (instruction RAM)

This RAM is connected to ITCM 64-bit interface designed for execution of critical real-times routines by the CPU.

- 128 Kbytes of DTCM-RAM (2x 64 Kbyte DTCM-RAMs on 2x32-bit DTCM ports)

The DTCM-RAM could be used for critical real-time data, such as interrupt service routines or stack/heap memory. Both DTCM-RAMs can be used in parallel (for load/store operations) thanks to the Cortex®-M7 dual issue capability.

3.4 Boot modes

At startup, the boot memory space is selected by the BOOT pin and BOOT_ADDx option bytes, allowing to program any boot memory address from 0x0000 0000 to 0x3FFF FFFF which includes:

- All flash address space
- All RAM address space: ITCM, DTCM RAMs and SRAMs
- The system memory bootloader

The boot loader is located in non-user System memory. It is used to reprogram the flash memory through a serial interface (USART, I2C, SPI, USB-DFU, FDCAN). Refer to *STM32 microcontroller system memory boot mode application note* (AN2606) for details.

3.5 Power supply management

3.5.1 Power supply scheme

- V_{DD} = 1.62 to 3.6 V: external power supply for I/Os, provided externally through V_{DD} pins.
- V_{DDLDO} = 1.62 to 3.6 V: supply voltage for the internal regulator supplying V_{CORE}
- V_{DDA} = 1.62 to 3.6 V: external analog power supplies for ADC, DAC, Reset blocks, RCs and PLL.
- $V_{DD33USB}$ and $V_{DD50USB}$:

$V_{DD50USB}$ can be supplied through the USB cable to generate the $V_{DD33USB}$ via the USB internal regulator. This allows supporting a V_{DD} supply different from 3.3 V.

The USB regulator can be bypassed to supply directly $V_{DD33USB}$ if V_{DD} = 3.3 V.

- V_{DDMMC} = 1.62 to 3.6 V external power supply for independent I/Os. V_{DDMMC} can be higher than V_{DD}. VDDMMC pin should be tied to VDD when it is not used.
- V_{BAT} = 1.2 to 3.6 V: power supply for the V_{SW} domain when V_{DD} is not present.
- V_{CAP}: V_{CORE} supply, which value depends on voltage scaling (0.74 V, 0.9 V, 1.0 V, 1.1 V, 1.2 V or 1.3 V). It is configured through VOS bits in PWR_CR3 register. The V_{CORE} domain is split into two domains the CPU domain (CD) and the Smart Run Domain (SRD).
  - CD domain containing most of the peripherals and the Arm® Cortex®-M7 core
  - SRD domain containing some peripherals and the system control.
- V_{DDSMPS} = 1.62 to 3.6 V: step-down converter power supply
- V_{LXSMPS} = V_{CORE} or 1.8 to 2.5 V: external regulated step-down converter output
- V_{FBSMPS} = V_{CORE} or 1.8 to 2.5 V: external step-down converter feedback voltage sense input

Note: For I/O speed optimization at low V_{DD} supply, refer to Section 3.8 General-purpose input/outputs (GPIOs).

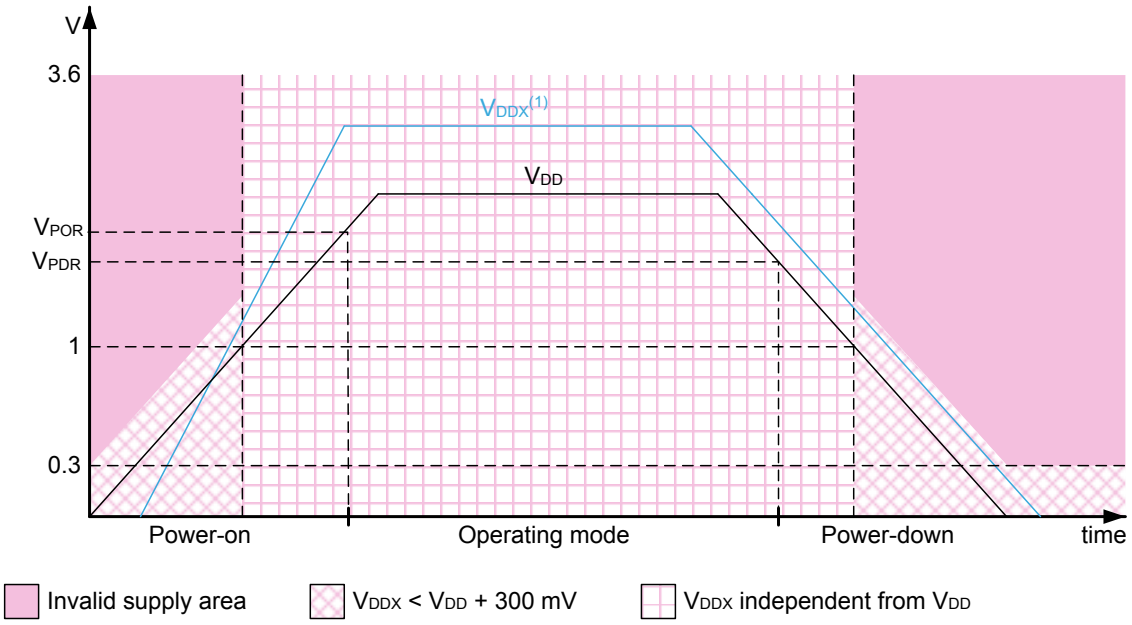
The features available on the device depend on the package (refer to Table 1. STM32H7B0xB features and peripheral counts).

During power-up and power-down phases, the following power sequence requirements must be respected (see Figure 2. Power-up/power-down sequence):

- When V_{DD} is below 1 V, other power supplies (V_{DDA}, $V_{DD33USB}$ and $V_{DD50USB}$) must remain below V_{DD} + 300 mV.
- When V_{DD} is above 1 V, all power supplies are independent (except for V_{DDSMPS}, which must remain at the same level as V_{DD}).

During the power-down phase, V_{DD} can temporarily become lower than other supplies only if the energy provided to the microcontroller remains below 1 mJ. This allows external decoupling capacitors to be discharged with different time constants during the power-down transient phase.

Figure 2. **Power-up/power-down sequence**



1. V_{DDx} refers to any power supply among V_{DDA}, $V_{DD33USB}$ and $V_{DD50USB}$.
2. V_{DD} and V_{DDSMPS} must be wired together into order to follow the same voltage sequence.

3.5.2 Power supply supervisor

The devices have an integrated power-on reset (POR)/ power-down reset (PDR) circuitry coupled with a Brownout reset (BOR) circuitry:

- Power-on reset (POR)

  The POR supervisor monitors V_{DD} power supply and compares it to a fixed threshold. The devices remain in reset mode when V_{DD} is below this threshold,
- Power-down reset (PDR)

  The PDR supervisor monitors V_{DD} power supply. A reset is generated when V_{DD} drops below a fixed threshold.

  The PDR supervisor can be enabled/disabled through PDR_ON pin.
- Brownout reset (BOR)

  The BOR supervisor monitors V_{DD} power supply. Three BOR thresholds (from 2.1 to 2.7 V) can be configured through option bytes. A reset is generated when V_{DD} drops below this threshold.
- Programmable voltage detector (PVD)

  The PVD monitors the V_{DD} power supply by comparing it with a threshold selected from a set of predefined values.

  It can also monitor the voltage level of the PVD_IN pin by comparing it with an internal V_{REFINT} voltage reference level.
- Analog voltage detector (AVD)

  The AVD monitors the V_{DDA} power supply by comparing it with a threshold selected from a set of predefined values.
- V_{BAT} threshold

  The V_{BAT} battery voltage level can be monitored by comparing it with two thresholds levels.
- Temperature threshold

  A dedicated temperature sensor monitors the junction temperature and compare it with two threshold levels.

3.5.3 Voltage regulator

The same voltage regulator supplies the two power domains (CD and SRD). The CD domain can be independently switched off.

Voltage regulator output can be adjusted according to application needs through six power supply levels:

- Run mode (VOS0 to VOS3)
  - Scale 0 and scale 1: high performance
  - Scale 2: medium performance and consumption
  - Scale 3: optimized performance and low-power consumption
- Stop mode (SVOS3 to SVOS5)
  - Scale 3: peripheral with wakeup from stop mode capabilities (UART, SPI, I2C, LPTIM) are operational
  - Scale 4 and 5 where the peripheral with wakeup from Stop mode is disabled

  The peripheral functionality is disabled but wakeup from Stop mode is possible through GPIO or asynchronous interrupt.

3.5.4 SMPS step-down converter

The built-in SMPS step-down converter is a highly power-efficient DC/DC non-linear switching regulator that provides lower power consumption than a conventional voltage regulator (LDO).

The step-down converter can be used to:

- Directly supply the V_{CORE} domain
  - the SMPS step-down converter operating modes follow the device system operating modes (Run, Stop, Standby).
  - the SMPS step-down converter output voltage are set according to the selected VOS and SVOS bits (voltage scaling)
- Provide intermediate voltage level to supply the internal voltage regulator (LDO)
  - The SMPS step-down converter operating modes follow the device system operating modes (Run, Stop, Standby).
  - The SMPS step-down converter output equals 1.8 V or 2.5 V according to the selected step-down level
- Provide an external supply
  - The SMPS step-down converter is forced to external operating mode
  - The SMPS step-down converter output equals 1.8 V or 2.5 V according to the selected step-down level

The 1.8 V or 2.5 V SMPS step-down converter output voltage imposes a minimum V_{DDSMPS} supply of 2.5 V or 3.3 V, respectively. It defines indirectly the minimum V_{DD} supply and I/O level.

3.6 Low-power modes

There are several ways to reduce power consumption on STM32H7B0xB:

- Decrease dynamic power consumption by slowing down the system clocks even in Run mode and individually clock gating the peripherals that are not used.
- Save power consumption when the CPU is idle, by selecting among the available low-power mode according to the user application needs. This allows achieving the best compromise between short startup time, low-power consumption, as well as available wakeup sources.

The devices feature several low-power modes:

- System Run with CSleep (CPU clock stopped)
- Autonomous with CD domain in DStop (CPU and CPU Domain bus matrix clocks stopped)
- Autonomous with CD domain in DStop2 (CPU and CPU Domain bus matrix clocks stopped, CPU domain in retention mode)
- System Stop (SRD domain clocks stopped) and CD domain in DStop (CPU and CPU Domain bus matrix clocks stopped)
- System Stop (SRD domain clocks stopped) and CD domain in DStop2 (CPU and CPU Domain bus matrix clocks stopped, CPU domain in retention mode)
- Standby (System, CD and SRD domains powered down)

CSleep and CStop low-power modes are entered by the MCU when executing the WFI (Wait for Interrupt) or WFE (Wait for Event) instructions, or when the SLEEPONEXIT bit of the Cortex®-M7 core is set after returning from an interrupt service routine.

The CPU domain can enter low-power mode (DStop or DStop2) when the processor, its subsystem and the peripherals allocated in the domain enter low-power mode.

If part of the domain is not in low-power mode, the domain remains in the current mode.

Finally the system can enter Stop or Standby when all EXTI wakeup sources are cleared and the power domains are in DStop or DStop2 mode.

Table 2. System vs domain low-power mode

System power mode	CD domain power mode	SRD domain power mode
Run	DRun/DStop/DStop2	DRun
Stop	DStop/DStop2	DStop
Standby	Standby	Standby

Some GPIO pins can be used to monitor CPU and domain power states:

Table 3. Overview of low-power mode monitoring pins

Power state monitoring pins	Description
PWR_CSLEEP	CPU clock OFF
PWR_CSTOP	CPU domain in low-power mode
PWR_NDSTOP2	CPU domain retention mode selection

3.7 Reset and clock controller (RCC)

The clock and reset controller is located in the SRD domain. The RCC manages the generation of all the clocks, as well as the clock gating and the control of the system and peripheral resets. It provides a high flexibility in the choice of clock sources and allows to apply clock ratios to improve the power consumption. In addition, on some communication peripherals that are capable to work with two different clock domains (either a bus interface clock or a kernel peripheral clock), the system frequency can be changed without modifying the baud rate.

3.7.1 Clock management

The devices embed four internal oscillators, two oscillators with external crystal or resonator, two internal oscillators with fast startup time and three PLLs.

The RCC receives the following clock source inputs:

- Internal oscillators:
  - 64 MHz HSI clock (1% accuracy)
  - 48 MHz RC oscillator
  - 4 MHz CSI clock
  - 32 kHz LSI clock
- External oscillators:
  - 4-50 MHz HSE clock
  - 32.768 kHz LSE clock

The RCC provides three PLLs: one for system clock, two for kernel clocks.

The system starts on the HSI clock. The user application can then select the clock configuration.

A high precision can be achieved for the 48 MHz clock by using the embedded clock recovery system (CRS). It uses the USB SOF signal, the LSE or an external signal (SYNC) to fine tune the oscillator frequency on-the- fly.

3.7.2 System reset sources

Power-on reset initializes all registers while system reset reinitializes the system except for the debug, part of the RCC and power controller status registers, as well as the backup power domain.

A system reset is generated in the following cases:

- Power-on reset (pwr_por_rst)
- Brownout reset
- Low level on NRST pin (external reset)
- Window watchdog
- Independent watchdog
- Software reset
- Low-power mode security reset
- Exit from Standby

3.8 General-purpose input/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain, with or without pull-up or pull-down), as input (floating, with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions. All GPIOs are high-current-capable and have speed selection to better manage internal noise, power consumption and electromagnetic emission.

After reset, all GPIOs are in Analog mode to reduce power consumption.

The I/O configuration can be locked if needed by following a specific sequence in order to avoid spurious writing to the I/Os registers.

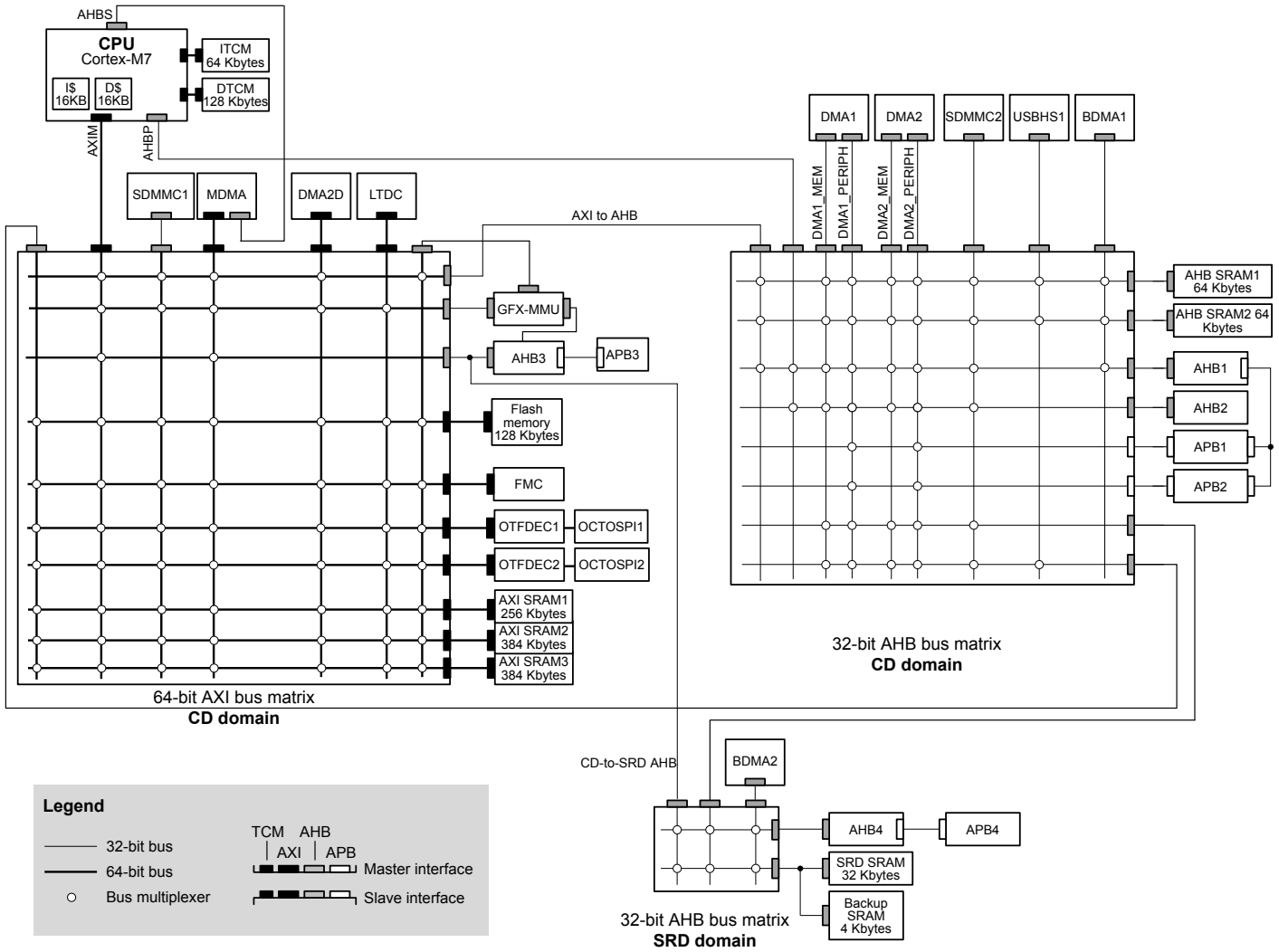
To maximize the performance, the I/O high-speed feature, HSLV, must be activated at low device supply voltage. This is needed to achieve the performance required for peripherals such as the SDMMC, FMC and OCTOSPI. The GPIOs are divided into four groups which can be optimized separately (refer to the description of HSLVx bits of SYSCFG_CCCSR register in RM0455).

The I/O high-speed feature must be used only when V_{DD} is lower than 2.7 V, and both the HSLV user option bits (VDDIO_HSLV and VDDMMC_HSLV) and HSLVx bits must be set to enable it (refer to RM0455 for details).

3.9 Bus-interconnect matrix

The devices feature an AXI bus matrix, two AHB bus matrices and bus bridges that allow interconnecting bus masters with bus slaves (see Figure 3. STM32H7B0xB bus matrix).

Figure 3. STM32H7B0xB bus matrix



3.10 DMA controllers

The devices feature five DMA instances to unload CPU activity:

- A master direct memory access (MDMA)

The MDMA is a high-speed DMA controller, which is in charge of all types of memory transfers (peripheral to memory, memory to memory, memory to peripheral), without any CPU action. It features a master AXI interface and a dedicated AHB interface to access Cortex®-M7 TCM memories.

The MDMA is located in the CD domain. It is able to interface with the other DMA controllers located in this domain to extend the standard DMA capabilities, or can manage peripheral DMA requests directly.

Each of the 16 channels can perform single block transfers, repeated block transfers and linked list transfers.

- Two dual-port DMAs (DMA1, DMA2) located in the CD domain and connected to the AHB matrix, with FIFO and request router capabilities.
- One basic DMA (BDMA1) located in the CD domain and connected to the AHB matrix. This DMA is dedicated to the DFSDM (see Section 3.26 Digital filter for sigma-delta modulators (DFSDM))
- One basic DMA (BDMA2) located in the SRD domain, with request router capabilities.

The DMA request router could be considered as an extension of the DMA controller. It routes the DMA peripheral requests to the DMA controller itself. This allowing managing the DMA requests with a high flexibility, maximizing the number of DMA requests that run concurrently, as well as generating DMA requests from peripheral output trigger or DMA event.

3.11 Chrom-ART Accelerator (DMA2D)

The Chrom-Art Accelerator (DMA2D) is a graphical accelerator which offers advanced bit blitting, row data copy and pixel format conversion. It supports the following functions:

- Rectangle filling with a fixed color
- Rectangle copy
- Rectangle copy with pixel format conversion
- Rectangle composition with blending and pixel format conversion

Various image format coding are supported, from indirect 4bpp color mode up to 32bpp direct color. It embeds dedicated memory to store color lookup tables. The DMA2D also supports block based YCbCr to handle JPEG decoder output.

An interrupt can be generated when an operation is complete or at a programmed watermark.

All the operations are fully automatized and are running independently from the CPU or the DMAs.

3.12 Chrom-GRC™ (GFXMMU)

The Chrom-GRC™ is a graphical oriented memory management unit aimed at:

- Optimizing memory usage according to the display shape
- Manage cache linear accesses to the frame buffer
- Prefetch data

The display shape is programmable to store only the visible image pixels.

A virtual memory space is provided which is seen by all system masters and can be physically mapped to any system memory.

An interrupt can be generated in case of buffer overflow or memory transfer error.

3.13 Nested vectored interrupt controller (NVIC)

The devices embed a nested vectored interrupt controller which is able to manage 16 priority levels, and handle up to 150 maskable interrupt channels plus the 16 interrupt lines of the Cortex®-M7 with FPU core.

- Closely coupled NVIC gives low-latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Allows early processing of interrupts
- Processing of late arriving, higher-priority interrupts
- Support tail chaining
- Processor context automatically saved

- Interrupt entry restored on interrupt exit with no instruction overhead

This hardware block provides flexible interrupt management features with minimum interrupt latency.

3.14 Extended interrupt and event controller (EXTI)

The EXTI controller performs interrupt and event management. In addition, it can wake up the processor, power domains and/or SRD domain from Stop mode.

The EXTI handles up to 89 independent event/interrupt lines split into 28 configurable events and 61 direct events.

Configurable events have dedicated pending flags, active edge selection, and software trigger capable.

Direct events provide interrupts or events from peripherals having a status flag.

3.15 Cyclic redundancy check calculation unit (CRC)

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code using a programmable polynomial.

Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the flash memory integrity. The CRC calculation unit helps compute a signature of the software during runtime, to be compared with a reference signature generated at link-time and stored at a given memory location.

3.16 Flexible memory controller (FMC)

The FMC controller main features are the following:

- Interface with static-memory mapped devices including:
  - Static random access memory (SRAM)
  - NOR flash memory/OneNAND flash memory
  - PSRAM (4 memory banks)
  - NAND flash memory with ECC hardware to check up to 8 Kbytes of data
- Interface with synchronous DRAM (SDRAM/Mobile LPSDR SDRAM) memories
- 8-,16-,32-bit data bus width
- Independent Chip Select control for each memory bank
- Independent configuration for each memory bank
- Write FIFO
- Read FIFO for SDRAM controller
- The maximum FMC_CLK/FMC_SDCLK frequency for synchronous accesses is the FMC kernel clock divided by 2.

3.17 Octo-SPI memory interface (OCTOSPI)

The OCTOSPI is a specialized communication interface targeting single, dual, quad or octal SPI memories.

The STM32H7B0xB embeds two separate Octo-SPI interfaces.

Each OCTOSPI instance supports single/dual/quad/octal SPI formats.

Multiplex of single/dual/quad/octal SPI over the same bus can be achieved using the integrated I/O manager.

The OCTOSPI can operate in any of the three following modes:

- Indirect mode: all the operations are performed using the OCTOSPI registers
- Status-polling mode: the external memory status register is periodically read and an interrupt can be generated in case of flag setting
- Memory-mapped mode: the external memory is memory mapped and it is seen by the system as if it was an internal memory supporting both read and write operations.

The OCTOSPI support two frame formats supported by most external serial memories such as serial PSRAMs, serial NOR flash memories, Hyper RAMs and Hyper flash memories:

- The classical frame format with the command, address, alternate byte, dummy cycles and data phase
- The HyperBus™ frame format.

Multichip package (MCP) combining any of the above mentioned memory types can also be supported.

3.18 Analog-to-digital converters (ADCs)

The STM32H7B0xB devices embed two analog-to-digital converters, whose resolution can be configured to 16, 14, 12, 10 or 8 bits. Each ADC shares up to 24 external channels, performing conversions in the single-shot or scan mode. In scan mode, automatic conversion is performed on a selected group of analog inputs.

Additional logic functions embedded in the ADC interface allow:

- Simultaneous sample and hold
- Interleaved sample and hold

The ADC can be served by the DMA controller, thus allowing to automatically transfer ADC converted values to a destination location without any software action.

In addition, an analog watchdog feature can accurately monitor the converted voltage of one, some or all selected channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

To synchronize A/D conversion and timers, the ADCs could be triggered by any of TIM1, TIM2, TIM3, TIM4, TIM6, TIM8, TIM15, and LPTIM1 timers.

3.19 Analog temperature sensor

The STM32H7B0xB embeds an analog temperature sensor that generates a voltage (V_{TS}) that varies linearly with the temperature. This temperature sensor is internally connected to ADC2_IN18. The conversion range is between 1.7 V and 3.6 V. It can measure the device junction temperature ranging from −40 to +125 °C.

The temperature sensor have a good linearity, but it has to be calibrated to obtain a good overall accuracy of the temperature measurement. As the temperature sensor offset varies from chip to chip due to process variation, the uncalibrated internal temperature sensor is suitable for applications that detect temperature changes only. To improve the accuracy of the temperature sensor measurement, each device is individually factory-calibrated by ST. The temperature sensor factory calibration data are stored by ST in the System memory area, which is accessible in read-only mode.

3.20 Digital temperature sensor (DTS)

The STM32H7B0xB embeds a sensor that converts the temperature into a square wave which frequency is proportional to the temperature. The PCLK or the LSE clock can be used as reference clock for the measurements. A formula given in the product reference manual (RM0455) allows to calculate the temperature according to the measured frequency stored in the DTS_DR register.

3.21 V_{BAT} operation

The V_{BAT} power domain contains the RTC, the backup registers and the backup SRAM.

To optimize battery duration, this power domain is supplied by V_{DD} when available or by the voltage applied on VBAT pin (when V_{DD} supply is not present). V_{BAT} power is switched when the PDR detects that V_{DD} dropped below the PDR level.

The voltage on the VBAT pin could be provided by an external battery, a supercapacitor or directly by V_{DD}, in which case, the V_{DD} mode is not functional.

V_{BAT} operation is activated when V_{DD} is not present.

The VBAT pin supplies the RTC, the backup registers and the backup SRAM.

The devices embed an internal V_{BAT} battery charging circuitry that can be activated when V_{DD} is present.

Note: When the microcontroller is supplied from V_{BAT}, external interrupts and RTC alarm/events do not exit it from V_{BAT} operation.

When PDR_ON pin is connected to V_{SS} (Internal Reset OFF), the V_{BAT} functionality is no more available and V_{BAT} pin should be connected to VDD.

3.22 Digital-to-analog converters (DAC)

The devices features one dual-channel DAC (DAC1), located in the CD domain, plus one single-channel DAC (DAC2), located in the SRD domain.

The three 12-bit buffered DAC channels can be used to convert three digital signals into three analog voltage signal outputs.

The following feature are supported:

- three DAC converters: one for each output channel
- 8-bit or 12-bit monotonic output
- left or right data alignment in 12-bit mode
- synchronized update capability
- noise-wave generation
- triangular-wave generation
- Triple DAC channel independent or simultaneous conversions
- DMA capability for each channel including DMA underrun error detection
- external triggers for conversion
- input voltage reference V_{REF+} or internal VREFBUF reference.

The DAC channels are triggered through the timer update outputs that are also connected to different DMA streams.

3.23 Voltage reference buffer (VREFBUF)

The built-in voltage reference buffer can be used as voltage reference for ADCs and DACs, as well as voltage reference for external components through the VREF+ pin.

Five different voltages are supported (refer to the reference manual for details).

3.24 Ultra-low-power comparators (COMP)

The STM32H7B0xB devices embed two rail-to-rail comparators (COMP1 and COMP2). They feature programmable reference voltage (internal or external), hysteresis and speed (low speed for low-power) as well as selectable output polarity.

The reference voltage can be one of the following:

- An external I/O
- A DAC output channel
- An internal reference voltage or submultiple (1/4, 1/2, 3/4)
- The analog temperature sensor
- The V_{BAT}/4 supply.

All comparators can wake up from Stop mode, generate interrupts and breaks for the timers, and be combined into a window comparator.

3.25 Operational amplifiers (OPAMP)

The STM32H7B0xB devices embed two rail-to-rail operational amplifiers (OPAMP1 and OPAMP2) with external or internal follower routing and PGA capability, and two inputs and one output each. These three I/Os can be connected to the external pins, thus enabling any type of external interconnections. The operational amplifiers can be configured internally as a follower, as an amplifier with a non-inverting gain ranging from 2 to 16 or with inverting gain ranging from -1 to -15.

The operational amplifier main features are:

- PGA with a non-inverting gain ranging of 2, 4, 8 or 16 or inverting gain ranging of -1, -3, -7 or -15
- Up to two positive inputs connected to DAC
- Output connected to internal ADC
- Low input bias current down to 1 nA
- Low input offset voltage down to 1.5 mV
- Gain bandwidth up to 8 MHz

The devices embed two operational amplifiers (OPMAP1 and OPAMP2) with two inputs and one output each. These three I/Os can be connected to the external pins, thus enabling any type of external interconnections. The operational amplifiers can be configured internally as a follower, as an amplifier with a non-inverting gain ranging from 2 to 16 or with inverting gain ranging from -1 to -15.

3.26 Digital filter for sigma-delta modulators (DFSDM)

The device embeds two DFSDM interfaces:

- DSFDM1

  It is located in the CD domain and features eight external digital serial interfaces (channels) and eight digital filters, or alternately eight internal parallel inputs.
- DSFDM2

  It is located in the SRD domain. DFSDM2 is a lite version including two external digital serial interfaces (channels) and one digital filters.

The DFSDM peripherals interface the external ΣΔ modulators to microcontroller and then perform digital filtering of the received data streams (which represent analog value on ΣΔ modulators inputs). DFSDMs can also interface PDM (Pulse Density Modulation) microphones and perform PDM to PCM conversion and filtering in hardware. The DFSDMs feature optional parallel data stream inputs from internal ADC peripherals or microcontroller memory (through DMA/CPU transfers into DFSDM).

DFSDM transceivers support several serial interface formats (to support various ΣΔ modulators). DFSDM digital filter modules perform digital processing according user selected filter parameters with up to 24-bit final ADC resolution.

The DFSDM peripherals support:

- Multiplexed input digital serial channels:
  - configurable SPI interface to connect various SD modulator(s)
  - configurable Manchester coded 1 wire interface support
  - PDM (Pulse Density Modulation) microphone input support
  - maximum input clock frequency up to 20 MHz (10 MHz for Manchester coding)
  - clock output for SD modulator(s): 0..20 MHz
- Alternative inputs from eight internal digital parallel channels (up to 16 bit input resolution):
  - internal sources: ADC data or memory data streams (DMA)
- Digital filter modules with adjustable digital signal processing:
  - Sinc^x filter: filter order/type (1..5), oversampling ratio (up to 1..1024)
  - integrator: oversampling ratio (1..256)
- Up to 24-bit output data resolution, signed output data format
- Automatic data offset correction (offset stored in register by user)
- Continuous or single conversion
- Start-of-conversion triggered by:
  - software trigger
  - internal timers
  - external events
  - start-of-conversion synchronously with first digital filter module (DFSDM0)
- Analog watchdog feature:
  - low value and high value data threshold registers
  - dedicated configurable Sincx digital filter (order = 1..3, oversampling ratio = 1..32)
  - input from final output data or from selected input digital serial channels
  - continuous monitoring independently from standard conversion
- Short circuit detector to detect saturated analog input values (bottom and top range):
  - up to 8-bit counter to detect 1..256 consecutive 0's or 1's on serial data stream
  - monitoring continuously each input serial channel
- Break signal generation on analog watchdog event or on short circuit detector event
- Extremes detector:
  - storage of minimum and maximum values of final conversion data
  - refreshed by software
- DMA capability to read the final conversion data
- Interrupts: end of conversion, overrun, analog watchdog, short circuit, input serial channel clock absence

- "Regular" or "injected" conversions:
  - "regular" conversions can be requested at any time or even in continuous mode without having any impact on the timing of "injected" conversions
  - "injected" conversions for precise timing and with high conversion priority

3.27 Digital camera interface (DCMI)

The devices embed a camera interface that can connect with camera modules and CMOS sensors through an 8-bit to 14-bit parallel interface, to receive video data. The camera interface can achieve a data transfer rate up to 140 Mbyte/s using a 80 MHz pixel clock. It features:

- Programmable polarity for the input pixel clock and synchronization signals
- Parallel data communication can be 8-, 10-, 12- or 14-bit
- Supports 8-bit progressive video monochrome or raw bayer format, YCbCr 4:2:2 progressive video, RGB 565 progressive video or compressed data (like JPEG)
- Supports continuous mode or snapshot (a single frame) mode
- Capability to automatically crop the image

3.28 Parallel synchronous slave interface (PSSI)

The PSSI is a generic synchronous 8-/16-bit parallel data input/output slave interface. It allows the transmitter to send a data valid signal to indicate when the data is valid, and the receiver to output a flow control signal to indicate when it is ready to sample the data.

The PSSI main features are:

- Slave mode operation
- 8- or 16-bit parallel data input or output
- 8-word (32-byte) FIFO
- Data enable (DE) alternate function input and Ready (RDY) alternate function output.

When enabled, these signals can either allow the transmitter to indicate when the data is valid or the receiver to indicate when it is ready to sample the data, or both.

The PSSI shares most of the circuitry with the digital camera interface (DCMI). It thus cannot be used simultaneously with the DCMI.

3.29 LCD-TFT display controller (LTDC)

The LCD-TFT display controller provides a 24-bit parallel digital RGB (Red, Green, Blue) and delivers all signals to interface directly to a broad range of LCD and TFT panels up to XGA (1024x768) resolution with the following features:

- 2 display layers with dedicated FIFO (64x32-bit)
- Color Look-Up table (CLUT) up to 256 colors (256x24-bit) per layer
- Up to 8 input color formats selectable per layer
- Flexible blending between two layers using alpha value (per pixel or constant)
- Flexible programmable parameters for each layer
- Color keying (transparency color)
- Up to 4 programmable interrupt events
- AXI master interface with burst of 16 words

3.30 JPEG codec (JPEG)

The JPEG codec can encode and decode a JPEG stream as defined in the

ISO/IEC10918-1 specification. It provides an fast and simple hardware compressor and decompressor of JPEG images with full management of JPEG headers.

The JPEG codec main features are as follows:

- 8-bit/channel pixel depths
- Single clock per pixel encoding and decoding
- Support for JPEG header generation and parsing

- Up to four programmable quantization tables
- Fully programmable Huffman tables (two AC and two DC)
- Fully programmable minimum coded unit (MCU)
- Encode/decode support (non simultaneous)
- Single clock Huffman coding and decoding
- Two-channel interface: Pixel/Compress In, Pixel/Compressed Out
- Stallable design
- Support for single greyscale component
- Ability to enable/disable header processing
- Internal register interface
- Fully synchronous design
- Configuration for high-speed decode mode

3.31 True random number generator (RNG)

All devices embed an RNG that delivers 32-bit random numbers generated by an integrated analog circuit. The RNG is a true random number generator that provides full entropy outputs to the application as 32-bit samples. It is composed of a live entropy source (analog) and an internal conditioning component.

3.32 Cryptographic acceleration (CRYP and HASH)

The devices embed a cryptographic processor that supports the advanced cryptographic algorithms usually required to ensure confidentiality, authentication, data integrity and non-repudiation when exchanging messages with a peer:

- Encryption/Decryption
  - DES/TDES (data encryption standard/triple data encryption standard): ECB (electronic codebook) and CBC (cipher block chaining) chaining algorithms, 64-, 128- or 192-bit key
  - AES (advanced encryption standard): ECB, CBC, GCM, CCM, and CTR (counter mode) chaining algorithms, 128, 192 or 256-bit key
- Universal HASH
  - SHA-1 and SHA-2 (secure HASH algorithms)
  - MD5
  - HMAC

The cryptographic accelerator supports DMA request generation.

3.33 On-the-fly decryption engine (OTFDEC)

The embedded OTFDEC decrypts in real-time the encrypted content stored in the external Octo-SPI memories used in Memory-mapped mode.

The OTFDEC uses the AES-128 algorithm in counter mode (CTR).

Code execution on external Octo-SPI memories can be protected against fault injection thanks to STMicroelectronics enhanced encryption mode (refer to RM0455 for details).

The OTFDEC main features are as follow:

- On-the-fly 128-bit decryption during STM32 Octo-SPI read operations (single or multiple).
  - AES-CTR algorithm with keystream FIFO (depth= 4)
  - Support for any read size
- Up to four independent encrypted regions
  - Region definition granularity: 4096 bytes
  - Region configuration write locking mechanism
  - Two optional decryption modes: execute-only and execute-never
- 128-bit key for each region, two-byte firmware version, and eight-byte application-defined nonce
- Encryption keys confidentiality and integrity protection
  - Write only registers with software locking mechanism
  - Availability of 8-bit CRC as public key information

- Support for STM32 Octo-SPI prefetching mechanism.
- Encryption mode

3.34 Timers and watchdogs

The devices include two advanced-control timers, ten general-purpose timers, two basic timers, three low-power timers, two watchdogs and a SysTick timer.

All timer counters can be frozen in Debug mode.

Table 4. Timer feature comparison compares the features of the advanced-control, general-purpose and basic timers.

Table 4. Timer feature comparison

Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/ compare channels	Comple-mentary output	Max interface clock (MHz)	Max timer clock (MHz) (1)
Advanced-control	TIM1, TIM8	16-bit	Up, Down, Up/down	Any integer between 1 and 65536	Yes	4	Yes	140	280
General purpose	TIM2, TIM5	32-bit	Up, Down, Up/down	Any integer between 1 and 65536	Yes	4	No	140	280
	TIM3, TIM4	16-bit	Up, Down, Up/down	Any integer between 1 and 65536	Yes	4	No	140	280
	TIM12	16-bit	Up	Any integer between 1 and 65536	No	2	No	140	280
	TIM13, TIM14	16-bit	Up	Any integer between 1 and 65536	No	1	No	140	280
	TIM15	16-bit	Up	Any integer between 1 and 65536	Yes	2	1	140	280
	TIM16, TIM17	16-bit	Up	Any integer between 1 and 65536	Yes	1	1	140	280
Basic	TIM6, TIM7	16-bit	Up	Any integer between 1 and 65536	Yes	0	No	140	280
Low-power timer	LPTIM1, LPTIM2, LPTIM3	16-bit	Up	1, 2, 4, 8, 16, 32, 64, 128	No	0	No	140	280

1. *The maximum timer clock is up to 280 MHz depending on TIMPRE bit in the RCC_CFGR register and CDPRE1/2 bits in RCC_CDCFGR register.*

3.34.1 Advanced-control timers (TIM1, TIM8)

The advanced-control timers (TIM1, TIM8) can be seen as three-phase PWM generators multiplexed on 6 channels. They have complementary PWM outputs with programmable inserted dead times. They can also be considered as complete general-purpose timers. Their 4 independent channels can be used for:

- Input capture
- Output compare
- PWM generation (edge- or center-aligned modes)
- One-pulse mode output

If configured as standard 16-bit timers, they have the same features as the general-purpose TIMx timers. If configured as 16-bit PWM generators, they have full modulation capability (0-100%).

The advanced-control timer can work together with the TIMx timers via the Timer Link feature for synchronization or event chaining.

The advanced-control timers support independent DMA request generation.

3.34.2 General-purpose timers (TIMx)

There are ten synchronizable general-purpose timers embedded in the STM32H7B0xB devices (see Table 4. Timer feature comparison for differences).

- **TIM2, TIM3, TIM4 and TIM5**

  The devices include 4 full-featured general-purpose timers: TIM2, TIM3, TIM4 and TIM5. TIM2 and TIM5 are based on a 32-bit auto-reload up/downcounter and a 16-bit prescaler while TIM3 and TIM4 are based on a 16-bit auto-reload up/downcounter and a 16-bit prescaler. All timers feature 4 independent channels for input capture/output compare, PWM or one-pulse mode output. This gives up to 16 input capture/output compare/PWMs on the largest packages.

  TIM2, TIM3, TIM4 and TIM5 general-purpose timers can work together, or with the other general-purpose timers and the advanced-control timers (TIM1, TIM8) via the Timer Link feature for synchronization or event chaining.

  Any of these general-purpose timers can be used to generate PWM outputs.

  TIM2, TIM3, TIM4 and TIM5 all have independent DMA request generation. They are capable of handling quadrature (incremental) encoder signals and the digital outputs from 1 to 4 hall-effect sensors.

- **TIM12, TIM13, TIM14, TIM15, TIM16 and TIM17**

  These timers are based on a 16-bit auto-reload upcounter and a 16-bit prescaler. TIM13, TIM14, TIM16 and TIM17 feature one independent channel, whereas TIM12 and TIM15 have two independent channels for input capture/output compare, PWM or one-pulse mode output. They can be synchronized with the TIM2, TIM3, TIM4 and TIM5 full-featured general-purpose timers or used as simple time bases.

3.34.3 Basic timers (TIM6 and TIM7)

These timers are mainly used for DAC trigger and waveform generation. They can also be used as a generic 16-bit time base.

TIM6 and TIM7 support independent DMA request generation.

3.34.4 Low-power timers (LPTIM1, LPTIM2, LPTIM3)

The low-power timers feature an independent clock and are running also in Stop mode if they are clocked by LSE, LSI or an external clock. The low-power timers are able to wakeup the devices from Stop mode.

The low-power timers support the following features:

- 16-bit up counter with 16-bit autoreload register
- 16-bit compare register
- Configurable output: pulse, PWM
- Continuous / one-shot mode
- Selectable software / hardware input trigger
- Selectable clock source:
- Internal clock source: LSE, LSI, HSI or APB clock
- External clock source over LPTIM input (working even with no internal clock source running, used by the Pulse Counter Application)
- Programmable digital glitch filter
- Encoder mode

3.34.5 Independent watchdog

The independent watchdog is based on a 12-bit downcounter and 8-bit prescaler. It is clocked from an independent 32 kHz internal RC and as it operates independently from the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free-running timer for application timeout management. It is hardware- or software-configurable through the option bytes.

3.34.6 Window watchdog

The window watchdog is based on a 7-bit downcounter that can be set as free-running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the main clock. It has an early warning interrupt capability and the counter can be frozen in debug mode.

3.34.7 SysTick timer

This timer is dedicated to real-time operating systems, but could also be used as a standard downcounter. It features:

- A 24-bit downcounter
- Autoreload capability
- Maskable system interrupt generation when the counter reaches 0
- Programmable clock source.

3.35 Real-time clock (RTC)

The RTC is an independent BCD timer/counter. It supports the following features:

- Calendar with subsecond, seconds, minutes, hours (12 or 24 format), week day, date, month, year, in BCD (binary-coded decimal) format.
- Automatic correction for 28, 29 (leap year), 30, and 31 days of the month.
- Two programmable alarms.
- On-the-fly correction from 1 to 32767 RTC clock pulses. This can be used to synchronize it with a master clock.
- Reference clock detection: a more precise second source clock (50 or 60 Hz) can be used to enhance the calendar precision.
- Digital calibration circuit with 0.95 ppm resolution, to compensate for quartz crystal inaccuracy.
- Timestamp feature which can be used to save the calendar content. This function can be triggered by an event on the timestamp pin, or by a tamper event, or by a switch to V_{BAT} mode.
- 17-bit auto-reload wakeup timer (WUT) for periodic events with programmable resolution and period.

The RTC is supplied through a switch that takes power either from the V_{DD} supply when present or from the V_{BAT} pin.

The RTC clock sources can be:

- A 32.768 kHz external crystal (LSE)
- An external resonator or oscillator (LSE)
- The internal low-power RC oscillator (LSI, with typical frequency of 32 kHz)
- The high-speed external clock (HSE) divided by 32.

The RTC is functional in V_{BAT} mode and in all low-power modes when it is clocked by the LSE. When clocked by the LSI, the RTC is not functional in V_{BAT} mode, but is functional in all low-power modes.

All RTC events (Alarm, Wakeup Timer, Timestamp or Tamper) can generate an interrupt and wakeup the device from the low-power modes.

3.36 Tamper and backup registers (TAMP)

The TAMP main features are the following:

- 32 backup registers:
  - The backup registers (TAMP_BKPxR) are implemented in the RTC domain that remains powered-on by V_{BAT} when the V_{DD} power is switched off.
- Three external tamper detection events
  - Each external event can be configured to be active or passive
  - External passive tampers with configurable filter and internal pull-up
- Seven internal tamper events
- Any tamper detection can generate an RTC timestamp event
- Any tamper detection can erase the RTC backup registers, the backup SRAM and the memory regions protected by the on-the-fly decryption engine (OTFDEC)

- Monotonic counter

3.37 Inter-integrated circuit interface (I^2C)

The STM32H7B0xB embed four I^2C interfaces.

The I^2C bus interface handles communications between the microcontroller and the serial I^2C bus. It controls all I^2C bus-specific sequencing, protocol, arbitration and timing.

The I2C peripheral supports:

- I^2C-bus specification and user manual rev. 5 compatibility:
  - Slave and master modes, multimaster capability
  - Standard-mode (Sm), with a bit rate up to 100 kbit/s
  - Fast-mode (Fm), with a bit rate up to 400 kbit/s
  - Fast-mode Plus (Fm+), with a bit rate up to 1 Mbit/s and 20 mA output drive I/Os
  - 7-bit and 10-bit addressing mode, multiple 7-bit slave addresses
  - Programmable setup and hold times
  - Optional clock stretching
- System management bus (SMBus) specification rev 2.0 compatibility:
  - Hardware PEC (packet error checking) generation and verification with ACK control
  - Address resolution protocol (ARP) support
  - SMBus alert
- Power system management protocol (PMBus®) specification rev 1.1 compatibility
- Independent clock: a choice of independent clock sources allowing the I2C communication speed to be independent from the PCLK reprogramming.
- Wakeup from Stop mode on address match
- Programmable analog and digital noise filters
- 1-byte buffer with DMA capability

3.38 Universal synchronous/asynchronous receiver transmitter (USART)

The STM32H7B0xB devices have five embedded universal synchronous receiver transmitters (USART1, USART2, USART3, USART6 and USART10) and five universal asynchronous receiver transmitters (UART4, UART5, UART7, UART8 and UART9). Refer to the table below for a summary of USARTx and UARTx features.

These interfaces provide asynchronous communication, IrDA SIR ENDEC support, multiprocessor communication mode, single-wire half-duplex communication mode and have LIN Master/Slave capability. They provide hardware management of the CTS and RTS signals, and RS485 Driver Enable. They are able to communicate at speeds of up to 10Mbit/s.

USART1, USART2, USART3, USART6 and USART10 also provide Smartcard mode (ISO 7816 compliant) and SPI-like communication capability.

The USARTs embed a Transmit FIFO (TXFIFO) and a Receive FIFO (RXFIFO). FIFO mode is enabled by software and is disabled by default.

All USART have a clock domain independent from the CPU clock, allowing the USARTx to wake up the MCU from Stop mode.The wakeup from Stop mode are programmable and can be done on:

- Start bit detection
- Any received data frame
- A specific programmed data frame
- Specific TXFIFO/RXFIFO status when FIFO mode is enabled.

All USART interfaces can be served by the DMA controller.

Table 5. USART features

X = supported.

USART modes/features	USART1/2/3/6/10	UART4/5/7/8/9
Hardware flow control for modem	X	X

USART modes/features	USART1/2/3/6/10	UART4/5/7/8/9
Continuous communication using DMA	X	X
Multiprocessor communication	X	X
Synchronous mode (Master/Slave)	X	-
Smartcard mode	X	-
Single-wire Half-duplex communication	X	X
IrDA SIR ENDEC block	X	X
LIN mode	X	X
Dual clock domain and wakeup from low power mode	X	X
Receiver timeout interrupt	X	X
Modbus communication	X	X
Auto baud rate detection	X	X
Driver Enable	X	X
USART data length	7, 8 and 9 bits	
Tx/Rx FIFO	X	X
Tx/Rx FIFO size	16	

3.39 Low-power universal asynchronous receiver transmitter (LPUART)

The device embeds one Low-power UART (LPUART1). The LPUART supports asynchronous serial communication with minimum power consumption. It supports half duplex single wire communication and modem operations (CTS/RTS). It allows multiprocessor communication.

The LPUART embeds a Transmit FIFO (TXFIFO) and a Receive FIFO (RXFIFO). FIFO mode is enabled by software and is disabled by default.

The LPUART has a clock domain independent from the CPU clock, and can wakeup the system from Stop mode. The wakeup from Stop mode are programmable and can be done on:

- Start bit detection
- Any received data frame
- A specific programmed data frame
- Specific TXFIFO/RXFIFO status when FIFO mode is enabled.

Only a 32.768 kHz clock (LSE) is needed to allow LPUART communication up to 9600 baud. Therefore, even in Stop mode, the LPUART can wait for an incoming frame while having an extremely low energy consumption. Higher speed clock can be used to reach higher baud rates.

LPUART interface can be served by the DMA controller.

3.40 Serial peripheral interfaces (SPI)/integrated interchip sound interfaces (I2S)

The devices feature up to six SPIs (SPI1/I2S1, SPI2/I2S2, SPI3/I2S3, SPI6/I2S6 and SPI4, SPI5) that allow communicating up to 125 Mbits/s in master and slave modes, in half-duplex, full-duplex and simplex modes. The 3-bit prescaler gives 8 master mode frequencies and the frame is configurable from 4 to 32 bits for SPI1/I2S1, SPI2/I2S2, SPI3/I2S3, from 4 to 16 bits for the others. All SPI interfaces support SS pulse mode, TI mode, Hardware CRC calculation, and 16x 8-bit embedded Rx and Tx FIFOs (SPI1/I2S1, SPI2/I2S2, SPI3/I2S3) or 8x 8-bit embedded Rx and Tx FIFOs (SPI4, SPI5, SPI6/I2S6), all with DMA capability. .

Four standard I^2S interfaces (multiplexed with SPI1, SPI2, SPI3, SPI6) are available. They can be operated in master or slave mode, in simplex communication modes, and can be configured to operate with a 16-/32-bit resolution as an input or output channel. Audio sampling frequencies from 8 kHz up to 192 kHz are supported. When one or all I^2S interfaces is/are configured in master mode, the master clock can be output to the external DAC/codec at 256 times the sampling frequency. All I^2S interfaces support 16x 8-bit embedded Rx and Tx FIFOs with DMA capability.

3.41 Serial audio interfaces (SAI)

The devices embed two SAIs (SAI1, SAI2) that allow designing many stereo or mono audio protocols such as I2S, LSB or MSB-justified, PCM/DSP, TDM or AC'97. An SPDIF output is available when the audio block is configured as a transmitter. To bring this level of flexibility and reconfigurability, the SAI contains two independent audio sub-blocks. Each block has it own clock generator and I/O line controller.

Audio sampling frequencies up to 192 kHz are supported.

One of the SAI supports up to 8 microphones thanks to an embedded PDM interface.

The SAI can work in master or slave configuration. The audio sub-blocks can be either receiver or transmitter and can work synchronously or asynchronously (with respect to the other one). The SAI can be connected with other SAIs to work synchronously.

3.42 SPDIFRX receiver interface (SPDIFRX)

The SPDIFRX peripheral is designed to receive an S/PDIF flow compliant with IEC-60958 and IEC-61937. These standards support simple stereo streams up to high sample rate, and compressed multi-channel surround sound, such as those defined by Dolby or DTS (up to 5.1).

The main SPDIFRX features are the following:

- Up to 4 inputs available
- Automatic symbol rate detection
- Maximum symbol rate: 12.288 MHz
- Stereo stream from 32 to 192 kHz supported
- Supports Audio IEC-60958 and IEC-61937, consumer applications
- Parity bit management
- Communication using DMA for audio samples
- Communication using DMA for control and user channel information
- Interrupt capabilities

The SPDIFRX receiver provides all the necessary features to detect the symbol rate, and decode the incoming data stream. The user can select the wanted SPDIF input, and when a valid signal will be available, the SPDIFRX will re-sample the incoming signal, decode the Manchester stream, recognize frames, sub-frames and blocks elements. It delivers to the CPU decoded data, and associated status flags.

The SPDIFRX also offers a signal named spdif_frame_sync, which toggles at the S/PDIF sub-frame rate that will be used to compute the exact sample rate for clock drift algorithms.

3.43 Single wire protocol master interface (SWPMI)

The single wire protocol master interface (SWPMI) is the master interface corresponding to the contactless frontend (CLF) defined in the ETSI TS 102 613 technical specification. The main features are:

- full-duplex communication mode
- automatic SWP bus state management (active, suspend, resume)
- configurable bit rate up to 2 Mbit/s
- automatic SOF, EOF and CRC handling

SWPMI can be served by the DMA controller.

3.44 Management data input/output (MDIO) slaves

The devices embed an MDIO slave interface it includes the following features:

- 32 MDIO register addresses, each of which is managed using separate input and output data registers:
  - 32 x 16-bit firmware read/write, MDIO read-only output data registers
  - 32 x 16-bit firmware read-only, MDIO write-only input data registers
- Configurable slave (port) address
- Independently maskable interrupts/events:
  - MDIO register write
  - MDIO register read
  - MDIO protocol error

- Able to operate in and wake up from STOP mode

3.45 SD/SDIO/MMC card host interfaces (SDMMC)

Two SDMMC host interfaces are available. They support *MultiMediaCard System Specification* version 4.51 in three different databus modes: 1 bit (default), 4 bits and 8 bits.

One of the SDMMC interface can be supplied through a separate V_{DDMMC} supply. If required, it can thus operate at a different voltage level than all other I/Os.

Both interfaces support the *SD memory card specifications* version 4.1. and the *SDIO card specification* version 4.0. in two different databus modes: 1 bit (default) and 4 bits.

Each SDMMC host interface supports only one SD/SDIO/MMC card at any one time and a stack of MMC Version 4.51 or previous.

The SDMMC host interface embeds a dedicated DMA controller allowing high-speed transfers between the interface and the SRAM.

3.46 Controller area network (FDCAN1, FDCAN2)

The controller area network (CAN) subsystem consists of two CAN modules, a shared message RAM memory and a clock calibration unit.

Both CAN modules (FDCAN1 and FDCAN2) are compliant with ISO 11898-1 (CAN protocol specification version 2.0 part A, B) and CAN FD protocol specification version 1.0.

FDCAN1 supports time triggered CAN (TTCAN) specified in ISO 11898-4, including event synchronized time-triggered communication, global system time, and clock drift compensation. FDCAN1 contains additional registers, specific to the time triggered feature. The CAN FD option can be used together with event-triggered and time-triggered CAN communication.

A 10 Kbyte message RAM memory implements filters, receive FIFOs, receive buffers, transmit event FIFOs, transmit buffers (and triggers for TTCAN). This message RAM is shared between the two FDCAN1 and FDCAN2 modules.

The common clock calibration unit is optional. It can be used to generate a calibrated clock for both FDCAN1 and FDCAN2 from the HSI internal RC oscillator and the PLL, by evaluating CAN messages received by the FDCAN1.

3.47 Universal serial bus on-the-go high-speed (OTG_HS)

The devices embed an USB OTG high-speed (up to 480 Mbit/s) device/host/OTG peripheral that supports both full-speed and high-speed operations. It integrates the transceivers for full-speed operation (12 Mbit/s) and a UTMI low-pin interface (ULPI) for high-speed operation (480 Mbit/s). When using the USB OTG_HS interface in HS mode, an external PHY device connected to the ULPI is required.

The USB OTG_HS peripheral is compliant with the USB 2.0 specification and with the OTG 2.0 specification. It features software-configurable endpoint setting and supports suspend/resume. The USB OTG_HS controller requires a dedicated 48 MHz clock that is generated by a PLL connected to the HSE oscillator.

The main features are:

- Combined Rx and Tx FIFO size of 4 Kbytes with dynamic FIFO sizing
- Supports the session request protocol (SRP) and host negotiation protocol (HNP)
- 8 bidirectional endpoints
- 16 host channels with periodic OUT support
- Software configurable to OTG1.3 and OTG2.0 modes of operation
- USB 2.0 LPM (Link Power Management) support
- Battery Charging Specification Revision 1.2 support
- Internal FS OTG PHY support
- External HS or HS OTG operation supporting ULPI in SDR mode

The OTG PHY is connected to the microcontroller ULPI port through 12 signals. It can be clocked using the 60 MHz output.

- Internal USB DMA
- HNP/SNP/IP inside (no need for any external resistor)
- For OTG/Host modes, a power switch is needed in case bus-powered devices are connected

3.48 High-definition multimedia interface (HDMI) - consumer electronics control (CEC)

The device embeds a HDMI-CEC controller that provides hardware support for the consumer electronics control (CEC) protocol (supplement 1 to the HDMI standard).

This protocol provides high-level control functions between all audiovisual products in an environment. It is specified to operate at low speeds with minimum processing and memory overhead. It has a clock domain independent from the CPU clock, allowing the HDMI-CEC controller to wake up the MCU from Stop mode on data reception.

3.49 Debug infrastructure

The devices offer a comprehensive set of debug and trace features to support software development and system integration.

- Breakpoint debugging
- Code execution tracing
- Software instrumentation
- JTAG debug port
- Serial-wire debug port
- Trigger input and output
- Serial-wire trace port
- Trace port
- Arm® CoreSight™ debug and trace components

The debug can be controlled via a JTAG/Serial-wire debug access port, using industry standard debugging tools.

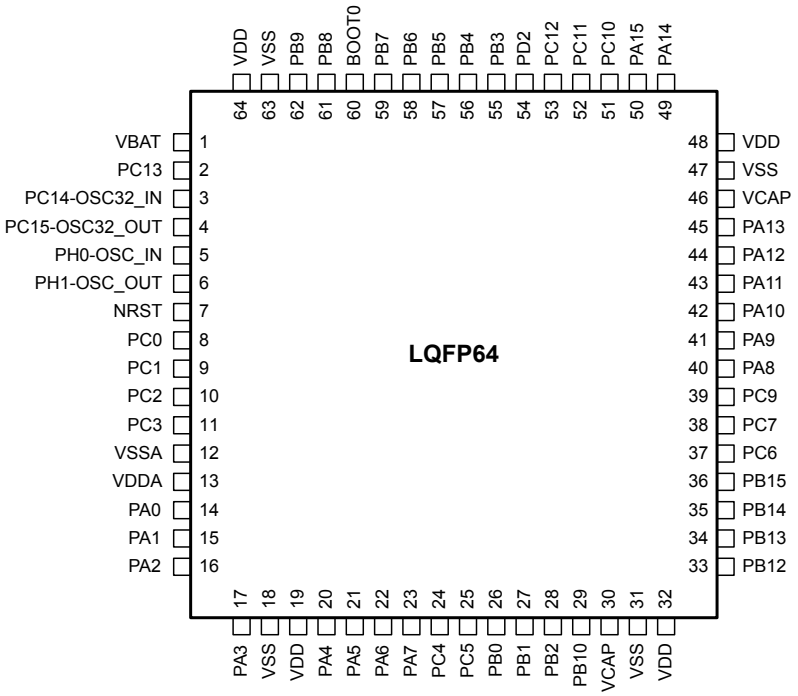
The trace port performs data capture for logging and analysis.

4 Memory mapping

Refer to the product line reference manual (RM0455) for details on the memory mapping as well as the boundary addresses for all peripherals.

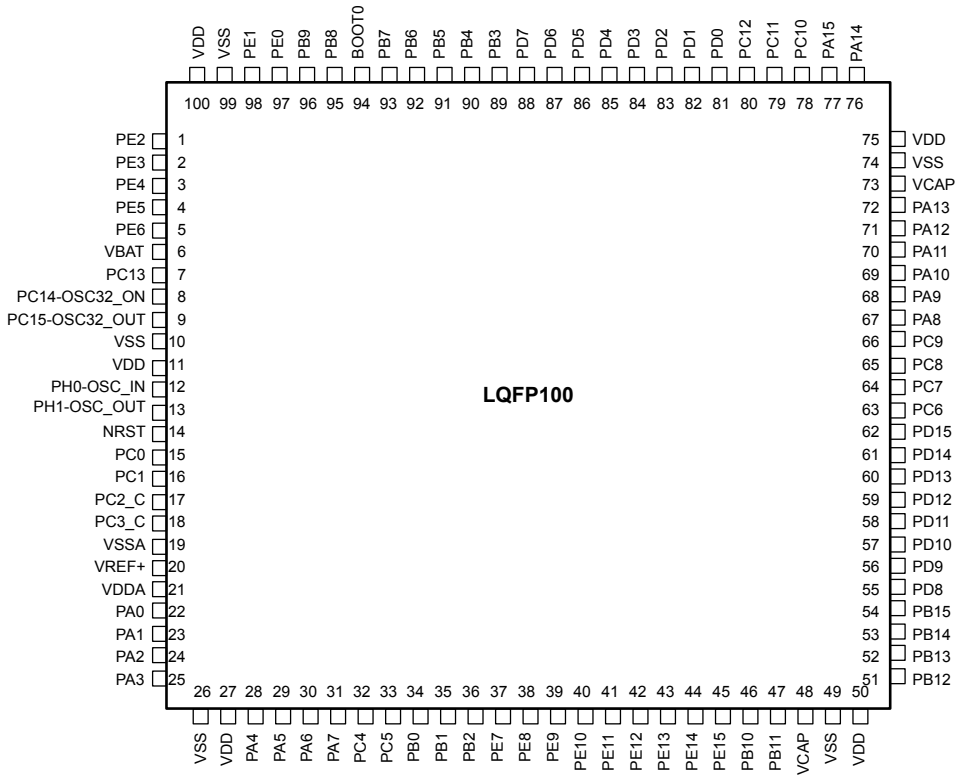
5 Pin descriptions

Figure 4. **LQFP64 (STM32H7B0xB without SMPS) pinout**



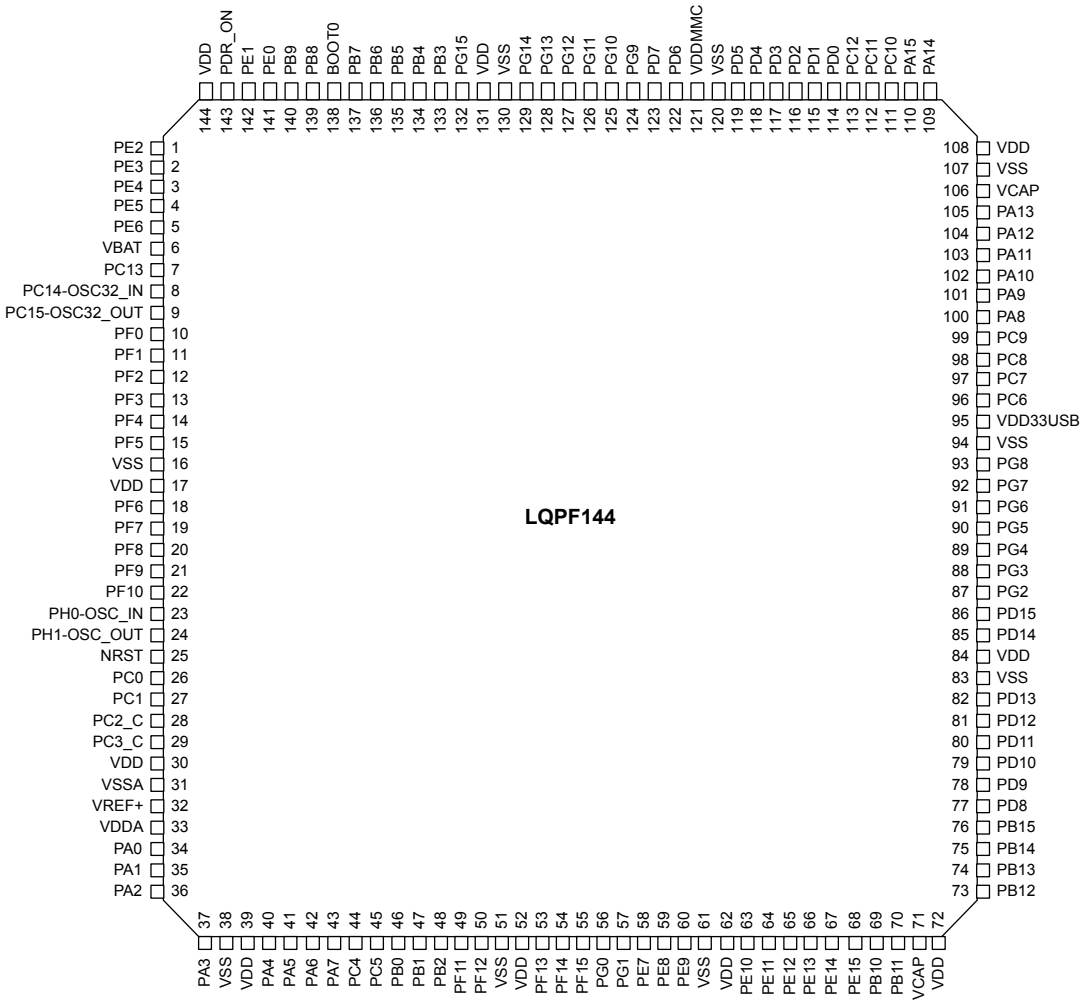
1. The above figure shows the package top view.

Figure 5. LQFP100 (STM32H7B0xB without SMPS) pinout



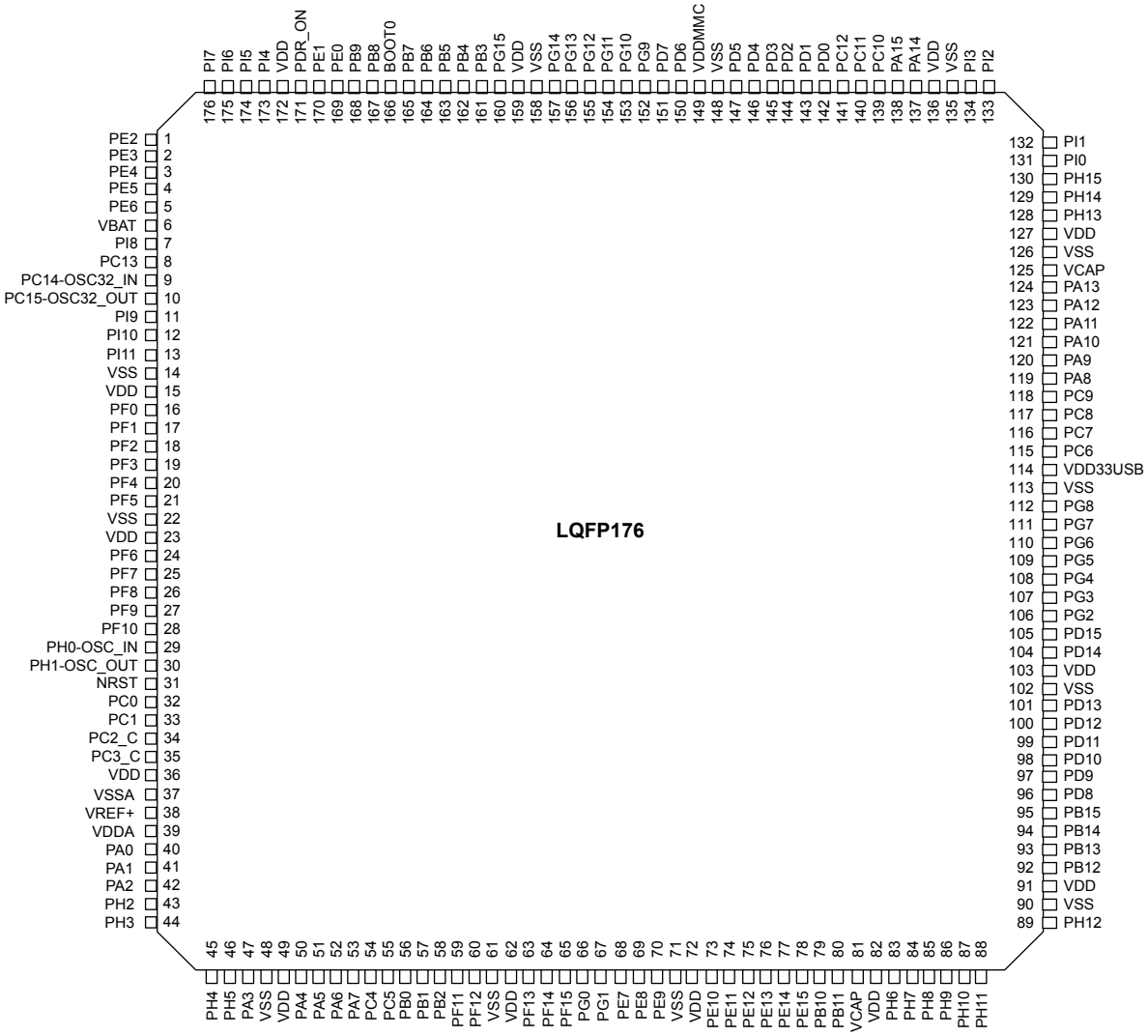
1. The above figure shows the package top view.

Figure 6. **LQFP144 (STM32H7B0xB without SMPS) pinout**



1. The above figure shows the package top view.

Figure 7. LQFP176 (STM32H7B0xB without SMPS) pinout



1. The above figure shows the package top view.

Figure 8. **UFBGA169 (STM32H7B0xB with SMPS) ballout**

	1	2	3	4	5	6	7	8	9	10	11	12	13
A	PE4	PE2	VDD	VCAP	PB6	VDDMMC	VDD	PG10	PD5	VDD	PC12	PC10	PH14
B	PC15-OSC32_OUT	PE3	VSS	VDDLDO	PB8	PB4	VSS	PG11	PD6	VSS	PC11	PA14	PH13
C	PC14-OSC32_IN	PE6	PE5	PDR_ON	PB9	PB5	PG14	PG9	PD4	PD1	PA15	VSS	VDD
D	VDD	VSS	PC13	PE1	PE0	PB7	PG13	PD7	PD3	PD0	PA13	VDDLDO	VCAP
E	VLXSMPS	VSSSMPS	VBAT	PF1	PF3	BOOT0	PG15	PG12	PD2	PA10	PA9	PA8	PA12
F	VDDSMPS	VFBSMPS	PF0	PF2	PF5	PF7	PB3	PG4	PC6	PC7	PC9	PC8	PA11
G	VDD	VSS	PF4	PF6	PF9	NRST	PF13	PE7	PG6	PG7	PG8	VDD50USB	VDD33USB
H	PH0-OSC_IN	PH1-OSC_OUT	PF10	PF8	PC2	PA4	PF14	PE8	PG2	PG3	PG5	VSS	VDD
J	PC0	PC1	VSSA	PC3	PA0	PA7	PF15	PE9	PE14	PD11	PD13	PD15	PD14
K	PC3_C	PC2_C	PA0_C	PA1	PA6	PC4	PG0	PE13	PH10	PH12	PD9	PD10	PD12
L	VDDA	VREF+	PA1_C	PA5	PB1	PB2	PG1	PE12	PB10	PH11	PB13	VSS	VDD
M	VDD	VSS	PH3	VSS	PB0	PF11	VSS	PE10	PB11	VDDLDO	VSS	PD8	PB15
N	PA2	PH2	PA3	VDD	PC5	PF12	VDD	PE11	PE15	VCAP	VDD	PB12	PB14

1. The above figure shows the package top view.

Figure 9. UFBGA176+25 (STM32H7B0xB with SMPS) ballout

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
A	VSS	PB8	VDDLDO	VCAP	PB6	PB3	PG11	PG9	PD3	PD1	PA15	PA14	VDDLDO	VCAP	VSS
B	PE4	PE3	PB9	PE0	PB7	PB4	PG13	PD7	PD5	PD2	PC12	PH14	PA13	PA8	PA12
C	PC13	VSS	PE2	PE1	BOOT0	PB5	PG14	PG10	PD4	PD0	PC11	PC10	PH13	PA10	PA11
D	PC15-OSC32_OUT	PC14-OSC32_IN	PE5	PDR_ON	VDD MMC	VSS	PG15	PG12	PD6	VSS	VDD	PH15	PA9	PC8	PC7
E	VSS	VBAT	PE6	VDD								VDD	PC9	PC6	VDD50 USB
F	VLX SMPS	VSS SMPS	PF1	PF0		VSS	VSS	VSS	VSS	VSS		VSS	VDD33 USB	PG6	PG5
G	VDD SMPS	VFB SMPS	PF2	VDD		VSS	VSS	VSS	VSS	VSS		PG8	PG7	PG4	PG2
H	PF6	PF4	PF5	PF3		VSS	VSS	VSS	VSS	VSS		VDD	PG3	PD14	PD13
J	PH0-OSC_IN	PF8	PF7	PF9		VSS	VSS	VSS	VSS	VSS		PD15	PD11	VSS	PD12
K	PH1-OSC_OUT	VSS	PF10	VDD		VSS	VSS	VSS	VSS	VSS		VSS	PD9	PB15	PB14
L	NRST	PC0	PC1	VREF−								VDD	PD10	PD8	PB13
M	PC2	PC3	VREF+	VDDA	VDD	VSS	PC5	PB1	VDD	VSS	PH7	PE14	PH11	PH9	PB12
N	PC2_C	PC3_C	VSSA	PH2	PA3	PA7	PF11	PE8	PG1	PF15	PF13	PB10	PH8	PH10	PH12
P	PA0	PA1	PA1_C	PH4	PA4	PA5	PB2	PG0	PE7	PB11	PF12	PE12	PE13	PE15	PH6
R	VSS	PA2	PA0_C	PH3	PH5	PC4	PA6	PB0	PE10	PF14	PE9	PE11	VCAP	VDDLDO	VSS

1. The above figure shows the package top view.
2. The devices with SMPS correspond to commercial code STM32H7B0IIK6Q.

Table 6. Legend/abbreviations used in the pinout table

Name	Abbreviation	Definition
Pin name	Unless otherwise specified in brackets below the pin name, the pin function during and after reset is the same as the actual pin name	
Pin type	S	Supply pin
	I	Input only pin
	I/O	Input / output pin
	ANA	Analog-only Input
I/O structure	FT	5 V tolerant I/O
	TT	3.3 V tolerant I/O
	B	Dedicated BOOT0 pin
	RST	Bidirectional reset pin with embedded weak pull-up resistor
	Option for TT and FT I/Os	
	_f	I2C FM+ option

Name		Abbreviation	Definition
I/O structure		_a	analog option (supplied by V_{DDA})
		_u	USB option (supplied by V_{DD33}USB)
		_h0 (1)	High-speed low voltage (mainly SDMMC2 on V_{DDMMC} power rail)
		_h1(1)	High-speed low voltage (mainly for OCTOSPI)
		_h2(1)	High-speed low voltage (mainly for FMC)
		_h3(1)	High-speed low voltage
		_s	Secondary supply (supplied by V_{DDMMC}) (2)
Notes		Unless otherwise specified by a note, all I/Os are set as floating inputs during and after reset.	
Pin functions	Alternate functions	Functions selected through GPIOx_AFR registers	
	Additional functions	Functions directly selected/enabled through peripheral registers	

1. *Refer to SYSCFG_CCCSR register in the device reference manual for how to set a group of I/Os in High-speed low-voltage mode. Depending on the chosen I/Os (for example OCTOSPI), it can belong to several groups of I/Os and several HSLVx bits need to be set (refer to Table Pin/ball definition). Take care that the VDDIO_HSLV and/or VDDMMC_HSLV option bits must also be set.*
2. *Refer to the table Features and peripheral counts for the list of packages featuring a V_{DDMMC} separate supply pad.*

Table 7. STM32H7B0xB pin/ball definition

Pin/ball name(1) (2)						Pin name (function after reset)	Pin type	I/O structure	Alternate functions	Additional functions
UFBGA169 with SMPS	UFBGA176+25 with SMPS	LQFP64	LQFP100	LQFP144	LQFP176					
A2	C3	-	1	1	1	PE2	I/O	FT_h2	TRACECLK, SAI1_CK1, SPI4_SCK, SAI1_MCLK_A, OCTOSPIM_P1_IO2, USART10_RX, FMC_A23, EVENTOUT	-
B2	B2	-	2	2	2	PE3	I/O	FT_h2	TRACED0, TIM15_BKIN, SAI1_SD_B, USART10_TX, FMC_A19, EVENTOUT	-
A1	B1	-	3	3	3	PE4	I/O	FT_h2	TRACED1, SAI1_D2, DFSDM1_DATIN3, TIM15_CH1N, SPI4_SS, SAI1_FS_A, FMC_A20, DCMI_D4/PSSI_D4, LCD_B0, EVENTOUT	-
C3	D3	-	4	4	4	PE5	I/O	FT_h2	TRACED2, SAI1_CK2, DFSDM1_CKIN3, TIM15_CH1, SPI4_MISO, SAI1_SCK_A, FMC_A21, DCMI_D6/PSSI_D6, LCD_G0, EVENTOUT	-
C2	E3	-	5	5	5	PE6	I/O	FT_h2	TRACED3, TIM1_BKIN2, SAI1_D1, TIM15_CH2, SPI4_MOSI, SAI1_SD_A, SAI2_MCK_B, TIM1_BKIN2_COMP12, FMC_A22, DCMI_D7/PSSI_D7, LCD_G1, EVENTOUT	-
B3	A1	-	-	-	-	VSS	S	-	-	-
A3	-	-	-	-	-	VDD	S	-	-	-
E3	E2	1	6	6	6	VBAT	S	-	-	-
D2	A15	-	-	-	-	VSS	S	-	-	-

Pin/ball name[1] [2]						Pin name (function after reset)	Pin type	I/O structure	Alternate functions	Additional functions
UFBGA169 with SMPS	UFBGA176+25 with SMPS	LQFP64	LQFP100	LQFP144	LQFP176					
-	-	-	-	-	7	PI8	I/O	FT	EVENTOUT	TAMP_IN2/ TAMP_OUT3, RTC_OUT2, WKUP4
D3	C1	2	7	7	8	PC13	I/O	FT	EVENTOUT	TAMP_IN1/ TAMP_OUT2/ TAMP_OUT3, RTC_OUT1/RTC_TS, WKUP3
-	C2	-	-	-	-	VSS	S	-	-	-
C1	D2	3	8	8	9	PC14-OSC32_IN (OSC32_IN)	I/O	FT	EVENTOUT	OSC32_IN
B1	D1	4	9	9	10	PC15-OSC32_OUT (OSC32_OUT)	I/O	FT	EVENTOUT	OSC32_OUT
-	-	-	-	-	11	PI9	I/O	FT_h2	OCTOSPIM_P2_IO0, UART4_RX, FDCAN1_RX, FMC_D30, LCD_VSYNC, EVENTOUT	-
-	-	-	-	-	12	PI10	I/O	FT_h2	OCTOSPIM_P2_IO1, FMC_D31, PSSI_D14, LCD_HSYNC, EVENTOUT	-
-	-	-	-	-	13	PI11	I/O	FT	OCTOSPIM_P2_IO2, LCD_G6, OTG_HS_ULPI_DIR, PSSI_D15, EVENTOUT	WKUP5
-	D10	-	-	-	14	VSS	S	-	-	-
D1	D11	-	-	-	15	VDD	S	-	-	-
E2	F2	-	-	-	-	VSSSMPS	S	-	-	-
E1	F1	-	-	-	-	VLXSMPS	S	-	-	-
F1	G1	-	-	-	-	VDDSMPS	S	-	-	-
F2	G2	-	-	-	-	VFBSMPS	S	-	-	-
F3	F4	-	-	10	16	PF0	I/O	FT_f	I2C2_SDA, OCTOSPIM_P2_IO0, FMC_A0, EVENTOUT	-
E4	F3	-	-	11	17	PF1	I/O	FT_f	I2C2_SCL, OCTOSPIM_P2_IO1, FMC_A1, EVENTOUT	-
F4	G3	-	-	12	18	PF2	I/O	FT_h2	I2C2_SMBA, OCTOSPIM_P2_IO2, FMC_A2, EVENTOUT	-
E5	H4	-	-	13	19	PF3	I/O	FT_h2	OCTOSPIM_P2_IO3, FMC_A3, EVENTOUT	-
G3	H2	-	-	14	20	PF4	I/O	FT_h2	OCTOSPIM_P2_CLK, FMC_A4, EVENTOUT	-
F5	H3	-	-	15	21	PF5	I/O	FT_h2	OCTOSPIM_P2_NCLK, FMC_A5, EVENTOUT	-
B7	E1	-	10	16	22	VSS	S	-	-	-
A7	E4	-	11	17	23	VDD	S	-	-	-
G4	H1	-	-	18	24	PF6	I/O	FT_h1	TIM16_CH1, SPI5_SS, SAI1_SD_B, UART7_Rx, OCTOSPIM_P1_IO3, EVENTOUT	-
F6	J3	-	-	19	25	PF7	I/O	FT_h1	TIM17_CH1, SPI5_SCK, SAI1_MCLK_B, UART7_Tx, OCTOSPIM_P1_IO2, EVENTOUT	-
H4	J2	-	-	20	26	PF8	I/O	FT_h1	TIM16_CH1N, SPI5_MISO, SAI1_SCK_B, UART7_RTS, TIM13_CH1, OCTOSPIM_P1_IO0, EVENTOUT	-

Pin/ball name[1] [2]						Pin name (function after reset)	Pin type	I/O structure	Alternate functions	Additional functions
UFBGA169 with SMPS	UFBGA176+25 with SMPS	LQFP64	LQFP100	LQFP144	LQFP176					
G5	J4	-	-	21	27	PF9	I/O	FT_h1	TIM17_CH1N, SPI5_MOSI, SAI1_FS_B, UART7_CTS, TIM14_CH1, OCTOSPIM_P1_IO1, EVENTOUT	-
H3	K3	-	-	22	28	PF10	I/O	FT_h1	TIM16_BKIN, SAI1_D3, PSSI_D15, OCTOSPIM_P1_CLK, DCMI_D11/ PSSI_D11, LCD_DE, EVENTOUT	-
H1	J1	5	12	23	29	PH0-OSC_IN(PH0)	I/O	FT	EVENTOUT	OSC_IN
H2	K1	6	13	24	30	PH1-OSC_OUT (PH1)	I/O	FT	EVENTOUT	OSC_OUT
G6	L1	7	14	25	31	NRST	I/O	RST	-	-
J1	L2	8	15	26	32	PC0	I/O	FT_a	DFSDM1_CKIN0, DFSDM1_DATIN4, SAI2_FS_B, FMC_A25, OTG_HS_ULPI_STP, LCD_G2, FMC_SDNWE, LCD_R5, EVENTOUT	ADC12_INP10
J2	L3	9	16	27	33	PC1	I/O	FT_ah0	TRACED0, SAI1_D1, DFSDM1_DATIN0, DFSDM1_CKIN4, SPI2_MOSI/I2S2_SDO, SAI1_SD_A, SDMMC2_CK, OCTOSPIM_P1_IO4, MDIOS_MDC, LCD_G5, EVENTOUT	ADC12_INP11, ADC12_INN10, TAMP_IN3, WKUP6
H5 [3]	M1 [3]	10	-	-	-	PC2	I/O	FT_a	PWR_CSTOP, DFSDM1_CKIN1, SPI2_MISO/I2S2_SDI, DFSDM1_CKOUT, OCTOSPIM_P1_IO2, OTG_HS_ULPI_DIR, OCTOSPIM_P1_IO5, FMC_SDNE0, EVENTOUT	ADC12_INP12, ADC12_INN11
K2 [3]	N1 [3]	-	17[4]	28[4]	34[4]	PC2_C	ANA	TT_a	-	ADC2_INP0, ADC2_INN1
J4[3]	M2[3]	11	-	-	-	PC3	I/O	FT_a	PWR_CSLEEP, DFSDM1_DATIN1, SPI2_MOSI/I2S2_SDO, OCTOSPIM_P1_IO0, OTG_HS_ULPI_NXT, OCTOSPIM_P1_IO6, FMC_SDCKE0, EVENTOUT	ADC12_INP13, ADC12_INN12
K1[3]	N2[3]	-	18[4]	29[4]	35[4]	PC3_C	ANA	TT_a	-	ADC2_INP1
G1	E12	-	-	30	36	VDD	S		-	-
G2	F6	-	-	-	-	VSS	S		-	-
J3	N3	12	19	31	37	VSSA	S		-	-
-	L4	-	-	-	-	VREF-	S		-	-
L2	M3	-	20	32	38	VREF+	S		-	-
L1	M4	13	21	33	39	VDDA	S		-	-
J5[3]	P1[3]	14	22	34	40	PA0	I/O	FT_a	TIM2_CH1/TIM2_ETR, TIM5_CH1, TIM8_ETR, TIM15_BKIN, SPI6_SS/I2S6_WS, USART2_CTS/ USART2_NSS, UART4_TX, SDMMC2_CMD, SAI2_SD_B, EVENTOUT	ADC1_INP16, WKUP1
K3[3]	R3[3]	-	-	-	-	PA0_C	ANA	TT_a	-	ADC1_INP0, ADC1_INN1
K4[3]	P2[3]	15	23	35	41	PA1	I/O	FT_ah1	TIM2_CH2, TIM5_CH2, LPTIM3_OUT, TIM15_CH1N, USART2_RTS, UART4_RX, OCTOSPIM_P1_IO3, SAI2_MCK_B, OCTOSPIM_P1_DQS, LCD_R2, EVENTOUT	ADC1_INP17, ADC1_INN16
L3[3]	P3[3]	-	-	-	-	PA1_C	ANA	TT_a	-	ADC1_INP1

Pin/ball name[1] [2]						Pin name (function after reset)	Pin type	I/O structure	Alternate functions	Additional functions
UFBGA169 with SMPS	UFBGA176+25 with SMPS	LQFP64	LQFP100	LQFP144	LQFP176					
N1	R2	16	24	36	42	PA2	I/O	FT_a	TIM2_CH3, TIM5_CH3, TIM15_CH1, DFSDM2_CKIN1, USART2_TX, SAI2_SCK_B, MDIOS_MDIO, LCD_R1, EVENTOUT	ADC1_INP14, WKUP2
N2	N4	-	-	-	43	PH2	I/O	FT_h2	LPTIM1_IN2, OCTOSPIM_P1_IO4, SAI2_SCK_B, FMC_SDCKE0, LCD_R0, EVENTOUT	-
M1	G4	-	-	-	-	VDD	S	-	-	-
M2	F7	-	-	-	-	VSS	S	-	-	-
M3	R4	-	-	-	44	PH3	I/O	FT_ah2	OCTOSPIM_P1_IO5, SAI2_MCK_B, FMC_SDNE0, LCD_R1, EVENTOUT	-
-	P4	-	-	-	45	PH4	I/O	FT_fa	I2C2_SCL, LCD_G5, OTG_HS_ULPI_NXT, PSSI_D14, LCD_G4, EVENTOUT	-
-	R5	-	-	-	46	PH5	I/O	FT_fa	I2C2_SDA, SPI5_SS, FMC_SDNWE, EVENTOUT	-
N3	N5	17	25	37	47	PA3	I/O	FT_ah1	TIM2_CH4, TIM5_CH4, OCTOSPIM_P1_CLK, TIM15_CH2, I2S6_MCK, USART2_RX, LCD_B2, OTG_HS_ULPI_D0, LCD_B5, EVENTOUT	ADC1_INP15
M4	F8	18	26	38	48	VSS	S	-	-	-
N4	H12	19	27	39	49	VDD	S	-	-	-
H6	P5	20	28	40	50	PA4	I/O	TT_a	TIM5_ETR, SPI1_SS/I2S1_WS, SPI3_SS/I2S3_WS, USART2_CK, SPI6_SS/I2S6_WS, DCMI_HSYNC/ PSSI_DE, LCD_VSYNC, EVENTOUT	ADC1_INP18, DAC1_OUT1
L4	P6	21	29	41	51	PA5	I/O	TT_ah0	PWR_NDSTOP2, TIM2_CH1/ TIM2_ETR, TIM8_CH1N, SPI1_SCK/ I2S1_CK, SPI6_SCK/I2S6_CK, OTG_HS_ULPI_CK, PSSI_D14, LCD_R4, EVENTOUT	ADC1_INP19, ADC1_INN18, DAC1_OUT2
K5	R7	22	30	42	52	PA6	I/O	TT_ah1	TIM1_BKIN, TIM3_CH1, TIM8_BKIN, SPI1_MISO/I2S1_SDI, OCTOSPIM_P1_IO3, SPI6_MISO/ I2S6_SDI, TIM13_CH1, TIM8_BKIN_COMP12, MDIOS_MDC, TIM1_BKIN_COMP12, DCMI_PIXCLK/ PSSI_PDCK, LCD_G2, EVENTOUT	ADC12_INP3, DAC2_OUT1
J6	N6	23	31	43	53	PA7	I/O	FT_ah1	TIM1_CH1N, TIM3_CH2, TIM8_CH1N, DFSDM2_DATIN1, SPI1_MOSI/ I2S1_SDO, SPI6_MOSI/I2S6_SDO, TIM14_CH1, OCTOSPIM_P1_IO2, FMC_SDNWE, LCD_VSYNC, EVENTOUT	ADC12_INP7, ADC12_INN3, OPAMP1_VINM
K6	R6	24	32	44	54	PC4	I/O	FT_a	DFSDM1_CKIN2, I2S1_MCK, SPDIFRX1_IN2, FMC_SDNE0, LCD_R7, EVENTOUT	ADC12_INP4, OPAMP1_VOUT, COMP1_INM
N5	M7	25	33	45	55	PC5	I/O	FT_ah1	SAI1_D3, DFSDM1_DATIN2, PSSI_D15, SPDIFRX1_IN3, OCTOSPIM_P1_DQS, FMC_SDCKE0, COMP1_OUT, LCD_DE, EVENTOUT	ADC12_INP8, ADC12_INN4, OPAMP1_VINM
N7	K4	-	-	-	-	VDD	S	-	-	-
M7	F9	-	-	-	-	VSS	S	-	-	-

Pin/ball name[1] [2]						Pin name (function after reset)	Pin type	I/O structure	Alternate functions	Additional functions
UFBGA169 with SMPS	UFBGA176+25 with SMPS	LQFP64	LQFP100	LQFP144	LQFP176					
M5	R8	26	34	46	56	PB0	I/O	FT_ah0	TIM1_CH2N, TIM3_CH3, TIM8_CH2N, DFSDM2_CKOUT, DFSDM1_CKOUT, UART4_CTS, LCD_R3, OTG_HS_ULPI_D1, OCTOSPIM_P1_IO1, LCD_G1, EVENTOUT	ADC12_INP9, ADC12_INN5, OPAMP1_VINP, COMP1_INP
L5	M8	27	35	47	57	PB1	I/O	FT_ah0	TIM1_CH3N, TIM3_CH4, TIM8_CH3N, DFSDM1_DATIN1, LCD_R6, OTG_HS_ULPI_D2, OCTOSPIM_P1_IO0, LCD_G0, EVENTOUT	ADC12_INP5, COMP1_INM
L6	P7	28	36	48	58	PB2	I/O	FT_ah1	RTC_OUT2, SAI1_D1, DFSDM1_CKIN1, SAI1_SD_A, SPI3_MOSI/I2S3_SDO, OCTOSPIM_P1_CLK, OCTOSPIM_P1_DQS, EVENTOUT	COMP1_INP
M6	N7	-	-	49	59	PF11	I/O	FT_ah1	SPI5_MOSI, OCTOSPIM_P1_NCLK, SAI2_SD_B, FMC_SDNRAS, DCMI_D12/PSSI_D12, EVENTOUT	ADC1_INP2
N6	P11	-	-	50	60	PF12	I/O	FT_ah2	OCTOSPIM_P2_DQS, FMC_A6, EVENTOUT	ADC1_INP6, ADC1_INN2
-	F10	-	-	51	61	VSS	S	-	-	-
-	L12	-	-	52	62	VDD	S	-	-	-
G7	N11	-	-	53	63	PF13	I/O	FT_ah2	DFSDM1_DATIN6, I2C4_SMBA, FMC_A7, EVENTOUT	ADC2_INP2
H7	R10	-	-	54	64	PF14	I/O	FT_fah2	DFSDM1_CKIN6, I2C4_SCL, FMC_A8, EVENTOUT	ADC2_INP6, ADC2_INN2
J7	N10	-	-	55	65	PF15	I/O	FT_fh2	I2C4_SDA, FMC_A9, EVENTOUT	-
K7	P8	-	-	56	66	PG0	I/O	FT_h2	OCTOSPIM_P2_IO4, UART9_RX, FMC_A10, EVENTOUT	-
-	F12	-	-	-	-	VSS	S	-	-	-
-	M5	-	-	-	-	VDD	S	-	-	-
L7	N9	-	-	57	67	PG1	I/O	FT_h2	OCTOSPIM_P2_IO5, UART9_TX, FMC_A11, EVENTOUT	OPAMP2_VINM
G8	P9	-	37	58	68	PE7	I/O	FT_ah2	TIM1_ETR, DFSDM1_DATIN2, UART7_Rx, OCTOSPIM_P1_IO4, FMC_D4/FMC_DA4, EVENTOUT	OPAMP2_VOUT, COMP2_INM
H8	N8	-	38	59	69	PE8	I/O	FT_ah2	TIM1_CH1N, DFSDM1_CKIN2, UART7_Tx, OCTOSPIM_P1_IO5, FMC_D5/FMC_DA5, COMP2_OUT, EVENTOUT	OPAMP2_VINM
J8	R11	-	39	60	70	PE9	I/O	FT_ah2	TIM1_CH1, DFSDM1_CKOUT, UART7_RTS, OCTOSPIM_P1_IO6, FMC_D6/FMC_DA6, EVENTOUT	OPAMP2_VINP, COMP2_INP
M11	G6	-	-	61	71	VSS	S	-	-	-
N11	M9	-	-	62	72	VDD	S	-	-	-
M8	R9	-	40	63	73	PE10	I/O	FT_ah2	TIM1_CH2N, DFSDM1_DATIN4, UART7_CTS, OCTOSPIM_P1_IO7, FMC_D7/FMC_DA7, EVENTOUT	COMP2_INM
N8	R12	-	41	64	74	PE11	I/O	FT_ah2	TIM1_CH2, DFSDM1_CKIN4, SPI4_SS, SAI2_SD_B, OCTOSPIM_P1_NCS, FMC_D8/FMC_DA8, LCD_G3, EVENTOUT	COMP2_INP

Pin/ball name[1] [2]						Pin name (function after reset)	Pin type	I/O structure	Alternate functions	Additional functions
UFBGA169 with SMPS	UFBGA176+25 with SMPS	LQFP64	LQFP100	LQFP144	LQFP176					
L8	P12	-	42	65	75	PE12	I/O	FT_h2	TIM1_CH3N, DFSDM1_DATIN5, SPI4_SCK, SAI2_SCK_B, FMC_D9/ FMC_DA9, COMP1_OUT, LCD_B4, EVENTOUT	-
K8	P13	-	43	66	76	PE13	I/O	FT_h2	TIM1_CH3, DFSDM1_CKIN5, SPI4_MISO, SAI2_FS_B, FMC_D10/ FMC_DA10, COMP2_OUT, LCD_DE, EVENTOUT	-
J9	M12	-	44	67	77	PE14	I/O	FT_h2	TIM1_CH4, SPI4_MOSI, SAI2_MCK_B, FMC_D11/FMC_DA11, LCD_CLK, EVENTOUT	-
N9	P14	-	45	68	78	PE15	I/O	FT_h2	TIM1_BKIN, USART10_CK, FMC_D12/ FMC_DA12, TIM1_BKIN_COMP12, LCD_R7, EVENTOUT	-
L9	N12	29	46	69	79	PB10	I/O	FT_f	TIM2_CH3, LPTIM2_IN1, I2C2_SCL, SPI2_SCK/I2S2_CK, DFSDM1_DATIN7, USART3_TX, OCTOSPIM_P1_NCS, OTG_HS_ULPI_D3, LCD_G4, EVENTOUT	-
M9	P10	-	47	70	80	PB11	I/O	FT_f	TIM2_CH4, LPTIM2_ETR, I2C2_SDA, DFSDM1_CKIN7, USART3_RX, OTG_HS_ULPI_D4, LCD_G5, EVENTOUT	-
N10	R13	30	48	71	81	VCAP	S	-	-	-
-	M10	31	49	-	-	VSS	S	-	-	-
M10	R14	-	-	-	-	VDDLDO	S	-	-	-
-	-	32	50	72	82	VDD	S	-	-	-
-	P15	-	-	-	83	PH6	I/O	FT	TIM12_CH1, I2C2_SMBA, SPI5_SCK, FMC_SDNE1, DCMI_D8/PSSI_D8, EVENTOUT	-
-	M11	-	-	-	84	PH7	I/O	FT_f	I2C3_SCL, SPI5_MISO, FMC_SDCKE1, DCMI_D9/PSSI_D9, EVENTOUT	-
-	N13	-	-	-	85	PH8	I/O	FT_fh2	TIM5_ETR, I2C3_SDA, FMC_D16, DCMI_HSYNC/PSSI_DE, LCD_R2, EVENTOUT	-
-	M14	-	-	-	86	PH9	I/O	FT_h2	TIM12_CH2, I2C3_SMBA, FMC_D17, DCMI_D0/PSSI_D0, LCD_R3, EVENTOUT	-
K9	N14	-	-	-	87	PH10	I/O	FT_h2	TIM5_CH1, I2C4_SMBA, FMC_D18, DCMI_D1/PSSI_D1, LCD_R4, EVENTOUT	-
L10	M13	-	-	-	88	PH11	I/O	FT_fh2	TIM5_CH2, I2C4_SCL, FMC_D19, DCMI_D2/PSSI_D2, LCD_R5, EVENTOUT	-
K10	N15	-	-	-	89	PH12	I/O	FT_fh2	TIM5_CH3, I2C4_SDA, FMC_D20, DCMI_D3/PSSI_D3, LCD_R6, EVENTOUT	-
L12	G10	-	-	-	90	VSS	S	-	-	-
L13	-	-	-	-	91	VDD	S	-	-	-
N12	M15	33	51	73	92	PB12	I/O	FT_h1	TIM1_BKIN, OCTOSPIM_P1_NCLK, I2C2_SMBA, SPI2_SS/ I2S2_WS, DFSDM1_DATIN1, USART3_CK, FDCAN2_RX, OTG_HS_ULPI_D5, DFSDM2_DATIN1, TIM1_BKIN_COMP12, UART5_RX, EVENTOUT	-

Pin/ball name[1] [2]						Pin name (function after reset)	Pin type	I/O structure	Alternate functions	Additional functions
UFBGA169 with SMPS	UFBGA176+25 with SMPS	LQFP64	LQFP100	LQFP144	LQFP176					
L11	L15	34	52	74	93	PB13	I/O	FT_h0	TIM1_CH1N, LPTIM2_OUT, DFSDM2_CKIN1, SPI2_SCK/I2S2_CK, DFSDM1_CKIN1, USART3_CTS/ USART3_NSS, FDCAN2_TX, OTG_HS_ULPI_D6, SDMMC1_D0, DCMI_D2/PSSI_D2, UART5_TX, EVENTOUT	-
N13	K15	35	53	75	94	PB14	I/O	FT_h0	TIM1_CH2N, TIM12_CH1, TIM8_CH2N, USART1_TX, SPI2_MISO/I2S2_SDI, DFSDM1_DATIN2, USART3_RTS, UART4_RTS, SDMMC2_D0, LCD_CLK, EVENTOUT	-
M13	K14	36	54	76	95	PB15	I/O	FT_h0	RTC_REFIN, TIM1_CH3N, TIM12_CH2, TIM8_CH3N, USART1_RX, SPI2_MOSI/ I2S2_SDO, DFSDM1_CKIN2, UART4_CTS, SDMMC2_D1, LCD_G7, EVENTOUT	-
M12	L14	-	55	77	96	PD8	I/O	FT_h2	DFSDM1_CKIN3, USART3_TX, SPDIFRX1_IN1, FMC_D13/FMC_DA13, EVENTOUT	-
K11	K13	-	56	78	97	PD9	I/O	FT_h2	DFSDM1_DATIN3, USART3_RX, FMC_D14/FMC_DA14, EVENTOUT	-
K12	L13	-	57	79	98	PD10	I/O	FT_h2	DFSDM1_CKOUT, DFSDM2_CKOUT, USART3_CK, FMC_D15/FMC_DA15, LCD_B3, EVENTOUT	-
-	H6	-	-	-	-	VSS	S	-	-	-
J10	J13	-	58	80	99	PD11	I/O	FT_h2	LPTIM2_IN2, I2C4_SMBA, USART3_CTS/USART3_NSS, OCTOSPIM_P1_IO0, SAI2_SD_A, FMC_A16/FMC_CLE, EVENTOUT	-
K13	J15	-	59	81	100	PD12	I/O	FT_fh2	LPTIM1_IN1, TIM4_CH1, LPTIM2_IN1, I2C4_SCL, USART3_RTS, OCTOSPIM_P1_IO1, SAI2_FS_A, FMC_A17/FMC_ALE, DCMI_D12/ PSSI_D12, EVENTOUT	-
J11	H15	-	60	82	101	PD13	I/O	FT_fh2	LPTIM1_OUT, TIM4_CH2, I2C4_SDA, OCTOSPIM_P1_IO3, SAI2_SCK_A, UART9_RTS, FMC_A18, DCMI_D13/ PSSI_D13, EVENTOUT	-
H12	R1	-	-	83	102	VSS	S	-	-	-
H13	-	-	-	84	103	VDD	S	-	-	-
J13	H14	-	61	85	104	PD14	I/O	FT_h2	TIM4_CH3, UART8_CTS, UART9_RX, FMC_D0/FMC_DA0, EVENTOUT	-
J12	J12	-	62	86	105	PD15	I/O	FT_h2	TIM4_CH4, UART8_RTS, UART9_TX, FMC_D1/FMC_DA1, EVENTOUT	-
-	D6	-	-	-	-	VSS	S	-	-	-
-	G7	-	-	-	-	VSS	S	-	-	-
H9	G15	-	-	87	106	PG2	I/O	FT_h2	TIM8_BKIN, TIM8_BKIN_COMP12, FMC_A12, EVENTOUT	-
H10	H13	-	-	88	107	PG3	I/O	FT_h2	TIM8_BKIN2, TIM8_BKIN2_COMP12, FMC_A13, EVENTOUT	-
C12	H10	-	-	-	-	VSS	S	-	-	-
C13	-	-	-	-	-	VDD	S	-	-	-
F8	G14	-	-	89	108	PG4	I/O	FT_h2	TIM1_BKIN2, TIM1_BKIN2_COMP12, FMC_A14/FMC_BA0, EVENTOUT	-

Pin/ball name[1] [2]						Pin name (function after reset)	Pin type	I/O structure	Alternate functions	Additional functions
UFBGA169 with SMPS	UFBGA176+25 with SMPS	LQFP64	LQFP100	LQFP144	LQFP176					
H11	F15	-	-	90	109	PG5	I/O	FT_h2	TIM1_ETR, FMC_A15/FMC_BA1, EVENTOUT	-
G9	F14	-	-	91	110	PG6	I/O	FT_h2	TIM17_BKIN, OCTOSPIM_P1_NCS, FMC_NE3, DCMI_D12/PSSI_D12, LCD_R7, EVENTOUT	-
G10	G13	-	-	92	111	PG7	I/O	FT_h2	SAI1_MCLK_A, USART6_CK, OCTOSPIM_P2_DQS, FMC_INT, DCMI_D13/PSSI_D13, LCD_CLK, EVENTOUT	-
G11	G12	-	-	93	112	PG8	I/O	FT_h2	TIM8_ETR, SPI6_SS/I2S6_WS, USART6_RTS, SPDIFRX1_IN2, FMC_SDCLK, LCD_G7, EVENTOUT	-
-	J6	-	-	94	113	VSS	S	-	-	-
G12	E15	-	-	-	-	VDD50USB	S	-	-	-
G13	F13	-	-	95	114	VDD33USB	S	-	-	-
F9	E14	37	63	96	115	PC6	I/O	FT_h0	TIM3_CH1, TIM8_CH1, DFSDM1_CKIN3, I2S2_MCK, USART6_TX, SDMMC1_D0DIR, FMC_NWAIT, SDMMC2_D6, SDMMC1_D6, DCMI_D0/PSSI_D0, LCD_HSYNC, EVENTOUT	SWPMI_IO
F10	D15	38	64	97	116	PC7	I/O	FT_h0	TRGIO, TIM3_CH2, TIM8_CH2, DFSDM1_DATIN3, I2S3_MCK, USART6_RX, SDMMC1_D123DIR, FMC_NE1, SDMMC2_D7, SWPMI_TX, SDMMC1_D7, DCMI_D1/PSSI_D1, LCD_G6, EVENTOUT	-
F12	D14	-	65	98	117	PC8	I/O	FT_h0	TRACED1, TIM3_CH3, TIM8_CH3, USART6_CK, UART5_RTS, FMC_NE2/ FMC_NCE, FMC_INT, SWPMI_RX, SDMMC1_D0, DCMI_D2/PSSI_D2, EVENTOUT	-
F11	E13	39	66	99	118	PC9	I/O	FT_fh0	MCO2, TIM3_CH4, TIM8_CH4, I2C3_SDA, I2S_CKIN, UART5_CTS, OCTOSPIM_P1_IO0, LCD_G3, SWPMI_SUSPEND, SDMMC1_D1, DCMI_D3/PSSI_D3, LCD_B2, EVENTOUT	-
-	J7	-	-	-	-	VSS	S	-	-	-
E12	B14	40	67	100	119	PA8	I/O	FT_fh0	MCO1, TIM1_CH1, TIM8_BKIN2, I2C3_SCL, USART1_CK, OTG_HS_SOF, UART7_RX, TIM8_BKIN2_COMP12, LCD_B3, LCD_R6, EVENTOUT	-
E11	D13	41	68	101	120	PA9	I/O	FT_u	TIM1_CH2, LPUART1_TX, I2C3_SMBA, SPI2_SCK/I2S2_CK, USART1_TX, DCMI_D0/PSSI_D0, LCD_R5, EVENTOUT	OTG_HS_VBUS
E10	C14	42	69	102	121	PA10	I/O	FT_u	TIM1_CH3, LPUART1_RX, USART1_RX, OTG_HS_ID, MDIOS_MDIO, LCD_B4, DCMI_D1/ PSSI_D1, LCD_B1, EVENTOUT	-
F13	C15	43	70	103	122	PA11	I/O	FT_u	TIM1_CH4, LPUART1_CTS, SPI2_SS/ I2S2_WS, UART4_RX, USART1_CTS/ USART1_NSS, FDCAN1_RX, LCD_R4, EVENTOUT	OTG_HS_DM

Pin/ball name[1] [2]						Pin name (function after reset)	Pin type	I/O structure	Alternate functions	Additional functions
UFBGA169 with SMPS	UFBGA176+25 with SMPS	LQFP64	LQFP100	LQFP144	LQFP176					
E13	B15	44	71	104	123	PA12	I/O	FT_u	TIM1_ETR, LPUART1_RTS, SPI2_SCK/I2S2_CK, UART4_TX, USART1_RTS, SAI2_FS_B, FDCAN1_TX, LCD_R5, EVENTOUT	OTG_HS_DP
D11	B13	45	72	105	124	PA13(JTMS/SWDIO)	I/O	FT	JTMS/SWDIO, EVENTOUT	-
D13	A14	46	73	106	125	VCAP	S	-	-	-
B10	M6	47	74	107	126	VSS	S	-	-	-
D12	A13	-	-	-	-	VDDLDO	S	-	-	-
A10	-	48	75	108	127	VDD	S	-	-	-
B13	C13	-	-	-	128	PH13	I/O	FT_h2	TIM8_CH1N, UART4_TX, FDCAN1_TX, FMC_D21, LCD_G2, EVENTOUT	-
A13	B12	-	-	-	129	PH14	I/O	FT_h2	TIM8_CH2N, UART4_RX, FDCAN1_RX, FMC_D22, DCMI_D4/PSSI_D4, LCD_G3, EVENTOUT	-
-	D12	-	-	-	130	PH15	I/O	FT_h2	TIM8_CH3N, FMC_D23, DCMI_D11/PSSI_D11, LCD_G4, EVENTOUT	-
-	-	-	-	-	131	PI0	I/O	FT_h2	TIM5_CH4, SPI2_SS/I2S2_WS, FMC_D24, DCMI_D13/PSSI_D13, LCD_G5, EVENTOUT	-
-	J9	-	-	-	-	VSS	S	-	-	-
-	-	-	-	-	132	PI1	I/O	FT_h2	TIM8_BKIN2, SPI2_SCK/I2S2_CK, TIM8_BKIN2_COMP12, FMC_D25, DCMI_D8/PSSI_D8, LCD_G6, EVENTOUT	-
-	-	-	-	-	133	PI2	I/O	FT_h2	TIM8_CH4, SPI2_MISO/I2S2_SDI, FMC_D26, DCMI_D9/PSSI_D9, LCD_G7, EVENTOUT	-
-	-	-	-	-	134	PI3	I/O	FT_h2	TIM8_ETR, SPI2_MOSI/I2S2_SDO, FMC_D27, DCMI_D10/PSSI_D10, EVENTOUT	-
-	J10	-	-	-	135	VSS	S	-	-	-
-	-	-	-	-	136	VDD	S	-	-	-
B12	A12	49	76	109	137	PA14(JTCK/SWCLK)	I/O	FT	JTCK/SWCLK, EVENTOUT	-
C11	A11	50	77	110	138	PA15(JTDI)	I/O	FT	JTDI, TIM2_CH1/TIM2_ETR, HDMI_CEC, SPI1_SS/I2S1_WS, SPI3_SS/I2S3_WS, SPI6_SS/I2S6_WS, UART4_RTS, LCD_R3, UART7_TX, LCD_B6, EVENTOUT	-
A12	C12	51	78	111	139	PC10	I/O	FT_h0	DFSDM1_CKIN5, DFSDM2_CKIN0, SPI3_SCK/I2S3_CK, USART3_TX, UART4_TX, OCTOSPIM_P1_IO1, LCD_B1, SWPMI_RX, SDMMC1_D2, DCMI_D8/PSSI_D8, LCD_R2, EVENTOUT	-
B11	C11	52	79	112	140	PC11	I/O	FT_h0	DFSDM1_DATIN5, DFSDM2_DATIN0, SPI3_MISO/I2S3_SDI, USART3_RX, UART4_RX, OCTOSPIM_P1_NCS, SDMMC1_D3, DCMI_D4/PSSI_D4, LCD_B4, EVENTOUT	-
A11	B11	53	80	113	141	PC12	I/O	FT_h0	TRACED3, TIM15_CH1, DFSDM2_CKOUT, SPI6_SCK/I2S6_CK, SPI3_MOSI/I2S3_SDO, USART3_CK, UART5_TX, SDMMC1_CK, DCMI_D9/PSSI_D9, LCD_R6, EVENTOUT	-

Pin/ball name[1] [2]						Pin name (function after reset)	Pin type	I/O structure	Alternate functions	Additional functions
UFBGA169 with SMPS	UFBGA176+25 with SMPS	LQFP64	LQFP100	LQFP144	LQFP176					
-	J14	-	-	-	-	VSS	S	-	-	-
D10	C10	-	81	114	142	PD0	I/O	FT_h2	DFSDM1_CKIN6, UART4_RX, FDCAN1_RX, UART9_CTS, FMC_D2/ FMC_DA2, LCD_B1, EVENTOUT	-
C10	A10	-	82	115	143	PD1	I/O	FT_h2	DFSDM1_DATIN6, UART4_TX, FDCAN1_TX, FMC_D3/FMC_DA3, EVENTOUT	-
E9	B10	54	83	116	144	PD2	I/O	FT_h0	TRACED2, TIM3_ETR, TIM15_BKIN, UART5_RX, LCD_B7, SDMMC1_CMD, DCMI_D11/PSSI_D11, LCD_B2, EVENTOUT	-
D9	A9	-	84	117	145	PD3	I/O	FT_h2	DFSDM1_CKOUT, SPI2_SCK/ I2S2_CK, USART2_CTS/USART2_NSS, FMC_CLK, DCMI_D5/PSSI_D5, LCD_G7, EVENTOUT	-
C9	C9	-	85	118	146	PD4	I/O	FT_h1	USART2_RTS, OCTOSPIM_P1_IO4, FMC_NOE, EVENTOUT	-
A9	B9	-	86	119	147	PD5	I/O	FT_h1	USART2_TX, OCTOSPIM_P1_IO5, FMC_NWE, EVENTOUT	-
-	K2	-	-	120	148	VSS	S	-	-	-
-	-	-	-	121	149	VDDMMC	S	-	-	-
B9	D9	-	87	122	150	PD6	I/O	FT_sh3	SAI1_D1, DFSDM1_CKIN4, DFSDM1_DATIN1, SPI3_MOSI/ I2S3_SDO, SAI1_SD_A, USART2_RX, OCTOSPIM_P1_IO6, SDMMC2_CK, FMC_NWAIT, DCMI_D10/PSSI_D10, LCD_B2, EVENTOUT	-
D8	B8	-	88	123	151	PD7	I/O	FT_sh3	DFSDM1_DATIN4, SPI1_MOSI/ I2S1_SDO, DFSDM1_CKIN1, USART2_CK, SPDIFRX1_IN0, OCTOSPIM_P1_IO7, SDMMC2_CMD, FMC_NE1, EVENTOUT	-
-	K6	-	-	-	-	VSS	S	-	-	-
A6	D5	-	-	-	-	VDDMMC	S	-	-	-
C8	A8	-	-	124	152	PG9	I/O	FT_sh3	SPI1_MISO/I2S1_SDI, USART6_RX, SPDIFRX1_IN3, OCTOSPIM_P1_IO6, SAI2_FS_B, SDMMC2_D0, FMC_NE2/ FMC_NCE, DCMI_VSYNC/PSSI_RDY, EVENTOUT	-
A8	C8	-	-	125	153	PG10	I/O	FT_sh3	OCTOSPIM_P2_IO6, SPI1_SS/ I2S1_WS, LCD_G3, SAI2_SD_B, SDMMC2_D1, FMC_NE3, DCMI_D2/ PSSI_D2, LCD_B2, EVENTOUT	-
B8	A7	-	-	126	154	PG11	I/O	FT_sh3	LPTIM1_IN2, SPI1_SCK/I2S1_CK, SPDIFRX1_IN0, OCTOSPIM_P2_IO7, SDMMC2_D2, USART10_RX, DCMI_D3/PSSI_D3, LCD_B3, EVENTOUT	-
E8	D8	-	-	127	155	PG12	I/O	FT_sh3	LPTIM1_IN1, OCTOSPIM_P2_NCS, SPI6_MISO/I2S6_SDI, USART6_RTS, SPDIFRX1_IN1, LCD_B4, SDMMC2_D3, USART10_TX, FMC_NE4, LCD_B1, EVENTOUT	-
D7	B7	-	-	128	156	PG13	I/O	FT_sh3	TRACED0, LPTIM1_OUT, SPI6_SCK/I2S6_CK, USART6_CTS/ USART6_NSS, SDMMC2_D6, USART10_CTS/USART10_NSS, FMC_A24, LCD_R0, EVENTOUT	-

Pin/ball name[1] [2]						Pin name (function after reset)	Pin type	I/O structure	Alternate functions	Additional functions
UFBGA169 with SMPS	UFBGA176+25 with SMPS	LQFP64	LQFP100	LQFP144	LQFP176					
C7	C7	-	-	129	157	PG14	I/O	FT_sh3	TRACED1, LPTIM1_ETR, SPI6_MOSI/I2S6_SDO, USART6_TX, OCTOSPIM_P1_IO7, SDMMC2_D7, USART10_RTS, FMC_A25, LCD_B0, EVENTOUT	-
-	K7	-	-	130	158	VSS	S	-	-	-
-	-	-	-	131	159	VDD	S	-	-	-
-	K8	-	-	-	-	VSS	S	-	-	-
E7	D7	-	-	132	160	PG15	I/O	FT_h1	USART6_CTS/USART6_NSS, OCTOSPIM_P2_DQS, USART10_CK, FMC_SDNCAS, DCMI_D13/PSSI_D13, EVENTOUT	-
F7	A6	55	89	133	161	PB3(JTDO/ TRACESWO)	I/O	FT_h0	JTDO/TRACESWO, TIM2_CH2, SPI1_SCK/I2S1_CK, SPI3_SCK/ I2S3_CK, SPI6_SCK/I2S6_CK, SDMMC2_D2, CRS_SYNC, UART7_RX, EVENTOUT	-
B6	B6	56	90	134	162	PB4(NJTRST)	I/O	FT_h0	NJTRST, TIM16_BKIN, TIM3_CH1, SPI1_MISO/I2S1_SDI, SPI3_MISO/ I2S3_SDI, SPI2_SS/I2S2_WS, SPI6_MISO/I2S6_SDI, SDMMC2_D3, UART7_TX, EVENTOUT	-
C6	C6	57	91	135	163	PB5	I/O	FT_h0	TIM17_BKIN, TIM3_CH2, I2C1_SMBA, SPI1_MOSI/I2S1_SDO, I2C4_SMBA, SPI3_MOSI/I2S3_SDO, SPI6_MOSI/ I2S6_SDO, FDCAN2_RX, OTG_HS_ULPI_D7, LCD_B5, FMC_SDCKE1, DCMI_D10/PSSI_D10, UART5_RX, EVENTOUT	-
A5	A5	58	92	136	164	PB6	I/O	FT_f	TIM16_CH1N, TIM4_CH1, I2C1_SCL, HDMI_CEC, I2C4_SCL, USART1_TX, LPUART1_TX, FDCAN2_TX, OCTOSPIM_P1_NCS, DFSDM1_DATIN5, FMC_SDNE1, DCMI_D5/PSSI_D5, UART5_TX, EVENTOUT	-
D6	B5	59	93	137	165	PB7	I/O	FT_fa	TIM17_CH1N, TIM4_CH2, I2C1_SDA, I2C4_SDA, USART1_RX, LPUART1_RX, DFSDM1_CKIN5, FMC_NL, DCMI_VSYNC/PSSI_RDY, EVENTOUT	PVD_IN
E6	C5	60	94	138	166	BOOT0	I	B	-	VPP
B5	A2	61	95	139	167	PB8	I/O	FT_fsh3	TIM16_CH1, TIM4_CH3, DFSDM1_CKIN7, I2C1_SCL, I2C4_SCL, SDMMC1_CKIN, UART4_RX, FDCAN1_RX, SDMMC2_D4, SDMMC1_D4, DCMI_D6/ PSSI_D6, LCD_B6, EVENTOUT	-
C5	B3	62	96	140	168	PB9	I/O	FT_fsh3	TIM17_CH1, TIM4_CH4, DFSDM1_DATIN7, I2C1_SDA, SPI2_SS/I2S2_WS, I2C4_SDA, SDMMC1_CDIR, UART4_TX, FDCAN1_TX, SDMMC2_D5, I2C4_SMBA, SDMMC1_D5, DCMI_D7/ PSSI_D7, LCD_B7, EVENTOUT	-
D5	B4	-	97	141	169	PE0	I/O	FT_h2	LPTIM1_ETR, TIM4_ETR, LPTIM2_ETR, UART8_RX, SAI2_MCK_A, FMC_NBL0, DCMI_D2/ PSSI_D2, LCD_R0, EVENTOUT	-

Pin/ball name[1] [2]						Pin name (function after reset)	Pin type	I/O structure	Alternate functions	Additional functions
UFBGA169 with SMPS	UFBGA176+25 with SMPS	LQFP64	LQFP100	LQFP144	LQFP176					
D4	C4	-	98	142	170	PE1	I/O	FT_h2	LPTIM1_IN2, UART8_TX, FMC_NBL1, DCMI_D3/PSSI_D3, LCD_R6, EVENTOUT	-
A4	A4	-	-	-	-	VCAP	S	-	-	-
-	K10	63	99	-	-	VSS	S	-	-	-
C4	D4	-	-	143	171	PDR_ON	S	-	-	-
B4	A3	-	-	-	-	VDDLDO	S	-	-	-
-	-	64	100	144	172	VDD	S	-	-	-
-	-	-	-	-	173	PI4	I/O	FT_h2	TIM8_BKIN, SAI2_MCK_A, TIM8_BKIN_COMP12, FMC_NBL2, DCMI_D5/PSSI_D5, LCD_B4, EVENTOUT	-
-	-	-	-	-	174	PI5	I/O	FT_h2	TIM8_CH1, SAI2_SCK_A, FMC_NBL3, DCMI_VSYNC/PSSI_RDY, LCD_B5, EVENTOUT	-
-	-	-	-	-	175	PI6	I/O	FT_h2	TIM8_CH2, SAI2_SD_A, FMC_D28, DCMI_D6/PSSI_D6, LCD_B6, EVENTOUT	-
-	-	-	-	-	176	PI7	I/O	FT_h2	TIM8_CH3, SAI2_FS_A, FMC_D29, DCMI_D7/PSSI_D7, LCD_B7, EVENTOUT	-
-	K12	-	-	-	-	VSS	S	-	-	-
-	G8	-	-	-	-	VSS	S	-	-	-
-	G9	-	-	-	-	VSS	S	-	-	-
-	H7	-	-	-	-	VSS	S	-	-	-
-	H8	-	-	-	-	VSS	S	-	-	-
-	H9	-	-	-	-	VSS	S	-	-	-
-	J8	-	-	-	-	VSS	S	-	-	-
-	K9	-	-	-	-	VSS	S	-	-	-
-	R15	-	-	-	-	VSS	S	-	-	-

1. *The devices with SMPS correspond to commercial code STM32H7B0xIxxQ.*
2. *A non-connected I/O in a given package is configured as an output tied to V_{SS}. Any analog peripheral connected to such a pad (such as OPAMP, VREF+) must be disabled.*
3. *Pxy_C and Pxy pins/balls are two separate pads (analog switch open). The analog switch is configured through a SYSCFG register. Refer to the product reference manual for a detailed description of the switch configuration bits.*
4. *There is a direct path between Pxy_C and Pxy pins/balls, through an analog switch. Pxy alternate functions are available on Pxy_C when the analog switch is closed. The analog switch is configured through a SYSCFG register. Refer to the product reference manual for a detailed description of the switch configuration bits.*

Table 8. Port A alternate functions

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	LPTIM1/ TIM1/2/16/17	PDM_SAI1/ TIM3/4/5/12/15	DFSDM1/ LPTIM2/3/ LPUART1/ OCTOSPIM_P1/2/ TIM8	CEC/DCMI/ PSSI/ DFSDM1/2/ I2C1/2/3/4/ LPTIM2/ TIM15/ USART1	CEC/SPI1/ I2S1/SPI2/ I2S2/SPI3/ I2S3/ SPI4/5/ SPI6/I2S6	DFSDM1/2/ I2C4/ OCTOSPIM_P1/ SAI1/SPI3/I2S3/ UART4	SDMMC1/ SPI2/I2S2/ SPI3/I2S3/ SPI6/I2S6/ UART7/ USART1/2/3/6	LPUART1/ SAI2/ SDMMC1/ SPDIFRX1/ SPI6/I2S6/ UART4/5/8	FDCAN1/2/FMC/ LCD/ OCTOSPIM_P1/2/ SDMMC2/ SPDIFRX1/ TIM13/14	CRS/FMC/LCD/ OCTOSPIM_P1/ OTG1_FS/OTG1_HS/ SAI2/SDMMC2/TIM8	DFSDM1/2/ I2C4/LCD/ MDIOS/ OCTOSPIM_P1/ SDMMC2/ SWPMI1/ TIM1/8/ UART7/9/ USART10	FMC/LCD/ MDIOS/ SDMMC1/ TIM1/8	COMP/DCMI/ PSSI/LCD/ TIM1	LCD/UART5	SYS
Port A	PA0	-	TIM2_CH1/ TIM2_ETR	TIM5_CH1	TIM8_ETR	TIM15_BKIN	SPI6_SS/ I2S6_WS	-	USART2_ CTS/ USART2_ NSS	UART4_TX	SDMMC2_CMD	SAI2_SD_B	-	-	-	-	EVENTOUT
	PA1	-	TIM2_CH2	TIM5_CH2	LPTIM3_OUT	TIM15_CH1N	-	-	USART2_ RTS	UART4_RX	OCTOSPIM_ P1_IO3	SAI2_MCK_B	OCTOSPIM_ P1_DQS	-	-	LCD_R2	EVENTOUT
	PA2	-	TIM2_CH3	TIM5_CH3	-	TIM15_CH1	-	DFSDM2_ CKIN1	USART2_ TX	SAI2_SCK_B	-	-	-	MDIOS_MDIO	-	LCD_R1	EVENTOUT
	PA3	-	TIM2_CH4	TIM5_CH4	OCTOSPIM_ P1_CLK	TIM15_CH2	I2S6_MCK	-	USART2_ RX	-	LCD_B2	OTG_HS_ ULPI_D0	-	-	-	LCD_B5	EVENTOUT
	PA4	-	-	TIM5_ETR	-	-	SPI1_SS/ I2S1_WS	SPI3_SS/ I2S3_WS	USART2_ CK	SPI6_SS/ I2S6_WS	-	-	-	-	DCMI_HSYNC/ PSSI_DE	LCD_ VSYNC	EVENTOUT
	PA5	PWR_NDSTOP2	TIM2_CH1/ TIM2_ETR	-	TIM8_CH1N	-	SPI1_SCK/ I2S1_CK	-	-	SPI6_SCK/ I2S6_CK	-	OTG_HS_ ULPI_CK	-	-	PSSI_D14	LCD_R4	EVENTOUT
	PA6	-	TIM1_BKIN	TIM3_CH1	TIM8_BKIN	-	SPI1_MISO/ I2S1_SDI	OCTOSPIM_ P1_IO3	-	SPI6_MISO/ I2S6_SDI	TIM13_CH1	TIM8_BKIN_COMP12	MDIOS_MDC	TIM1_BKIN_ COMP12	DCMI_PIXCLK/ PSSI_PDCK	LCD_G2	EVENTOUT
	PA7	-	TIM1_CH1N	TIM3_CH2	TIM8_CH1N	DFSDM2_ DATIN1	SPI1_MOSI/ I2S1_SDO	-	-	SPI6_MOSI/ I2S6_SDO	TIM14_CH1	OCTOSPIM_P1_IO2	-	FMC_SDNWE	-	LCD_VSYNC	EVENTOUT
	PA8	MCO1	TIM1_CH1	-	TIM8_BKIN2	I2C3_SCL	-	-	USART1_ CK	-	-	OTG_HS_ SOF	UART7_RX	TIM8_BKIN2_ COMP12	LCD_B3	LCD_R6	EVENTOUT
	PA9	-	TIM1_CH2	-	LPUART1_TX	I2C3_SMBA	SPI2_SCK/ I2S2_CK	-	USART1_ TX	-	-	-	-	-	DCMI_D0/ PSSI_D0	LCD_R5	EVENTOUT
	PA10	-	TIM1_CH3	-	LPUART1_RX	-	-	-	USART1_ RX	-	-	OTG_HS_ ID	MDIOS_MDIO	LCD_B4	DCMI_D1/ PSSI_D1	LCD_B1	EVENTOUT
	PA11	-	TIM1_CH4	-	LPUART1_CTS	-	SPI2_SS/ I2S2_WS	UART4_RX	USART1_ CTS/ USART1_NSS	-	FDCAN1_ RX	-	-	-	-	LCD_R4	EVENTOUT
	PA12	-	TIM1_ETR	-	LPUART1_RTS	-	SPI2_SCK/ I2S2_CK	UART4_TX	USART1_ RTS	SAI2_FS_B	FDCAN1_ TX	-	-	-	-	LCD_R5	EVENTOUT
	PA13	JTMS/ SWDIO	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENTOUT
	PA14	JTCK/ SWCLK	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENTOUT
	PA15	JTDI	TIM2_CH1/ TIM2_ETR	-	-	HDMI_CEC	SPI1_SS/ I2S1_WS	SPI3_SS/ I2S3_WS	SPI6_SS/ I2S6_WS	UART4_ RTS	LCD_R3	-	UART7_TX	-	-	LCD_B6	EVENTOUT

Table 9. Port B alternate functions

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	LPTIM1/ TIM1/2/16/17	PDM_SAI1/ TIM3/4/5/12/15	DFSDM1/ LPTIM2/3/ LPUART1/ OCTOSPIM_P1/2/ TIM8	CEC/DCMI/PSSI/ DFSDM1/2/ I2C1/2/3/4/ LPTIM2/TIM15/ USART1	CEC/SPI1/ I2S1/SPI2/ I2S2/SPI3/ I2S3/ SPI4/5/ SPI6/I2S6	DFSDM1/2/I2C4/ OCTOSPIM_P1/ SAI1/SPI3/I2S3/ UART4	SDMMC1/SPI2/ I2S2/SPI3/I2S3/ SPI6/I2S6/ UART7/ USART1/2/3/6	LPUART1/ SAI2/ SDMMC1/ SPDIFRX1/ SPI6/I2S6/ UART4/5/8	FDCAN1/2/FMC/LCD/OCTOSPIM_P1/2/ SDMMC2/ SPDIFRX1/TIM13/14	CRS/FMC/LCD/ OCTOSPIM_P1/ OTG1_FS/ OTG1_HS/ SAI2/SDMMC2/ TIM8	DFSDM1/2/ I2C4/LCD/MDIOS/ OCTOSPIM_P1/ SDMMC2/SWPMI1/ TIM1/8/UART7/9/ USART10	FMC/LCD/ MDIOS/ SDMMC1/ TIM1/8	COMP/DCMI/ PSSI/LCD/TIM1	LCD/ UART5	SYS
Port B	PB0	-	TIM1_CH2N	TIM3_CH3	TIM8_CH2N	DFSDM2_CKOUT	-	DFSDM1_CKOUT	-	UART4_CTS	LCD_R3	OTG_HS_ ULPI_D1	OCTOSPIM_P1_IO1	-	-	LCD_G1	EVENTOUT
	PB1	-	TIM1_CH3N	TIM3_CH4	TIM8_CH3N	-	-	DFSDM1_DATIN1	-	-	LCD_R6	OTG_HS_ ULPI_D2	OCTOSPIM_ P1_IO0	-	-	LCD_G0	EVENTOUT
	PB2	RTC_OUT2	-	SAI1_D1	-	DFSDM1_CKIN1	-	SAI1_SD_A	SPI3_MOSI/ I2S3_SDO	-	OCTOSPIM_P1_CLK	OCTOSPIM_ P1_DQS	-	-	-	-	EVENTOUT
	PB3	JTDO/ TRACESWO	TIM2_CH2	-	-	-	SPI1_SCK/ I2S1_CK	SPI3_SCK/ I2S3_CK	-	SPI6_SCK/ I2S6_CK	SDMMC2_D2	CRS_SYNC	UART7_RX	-	-	-	EVENTOUT
	PB4	NJTRST	TIM16_BKIN	TIM3_CH1	-	-	SPI1_MISO/ I2S1_SDI	SPI3_MISO/ I2S3_SDI	SPI2_SS/ I2S2_WS	SPI6_MISO/ I2S6_SDI	SDMMC2_D3	-	UART7_TX	-	-	-	EVENTOUT
	PB5	-	TIM17_BKIN	TIM3_CH2	-	I2C1_SMBA	SPI1_MOSI/ I2S1_SDO	I2C4_SMBA	SPI3_MOSI/ I2S3_SDO	SPI6_MOSI/ I2S6_SDO	FDCAN2_RX	OTG_HS_ ULPI_D7	LCD_B5	FMC_SDCKE1	DCMI_D10/ PSSI_D10	UART5_RX	EVENTOUT
	PB6	-	TIM16_CH1N	TIM4_CH1	-	I2C1_SCL	HDMI_CEC	I2C4_SCL	USART1_TX	LPUART1_TX	FDCAN2_TX	OCTOSPIM_ P1_NCS	DFSDM1_DATIN5	FMC_SDNE1	DCMI_D5/PSSI_D5	UART5_TX	EVENTOUT
	PB7	-	TIM17_CH1N	TIM4_CH2	-	I2C1_SDA	-	I2C4_SDA	USART1_RX	LPUART1_RX	-	-	DFSDM1_CKIN5	FMC_NL	DCMI_VSYNC/ PSSI_RDY	-	EVENTOUT
	PB8	-	TIM16_CH1	TIM4_CH3	DFSDM1_CKIN7	I2C1_SCL	-	I2C4_SCL	SDMMC1_CKIN	UART4_RX	FDCAN1_RX	SDMMC2_D4	-	SDMMC1_D4	DCMI_D6/PSSI_D6	LCD_B6	EVENTOUT
	PB9	-	TIM17_CH1	TIM4_CH4	DFSDM1_DATIN7	I2C1_SDA	SPI2_SS/ I2S2_WS	I2C4_SDA	SDMMC1_CDIR	UART4_TX	FDCAN1_TX	SDMMC2_D5	I2C4_SMBA	SDMMC1_D5	DCMI_D7/PSSI_D7	LCD_B7	EVENTOUT
	PB10	-	TIM2_CH3	-	LPTIM2_IN1	I2C2_SCL	SPI2_SCK/ I2S2_CK	DFSDM1_DATIN7	USART3_TX	-	OCTOSPIM_ P1_NCS	OTG_HS_ ULPI_D3	-	-	-	LCD_G4	EVENTOUT
	PB11	-	TIM2_CH4	-	LPTIM2_ETR	I2C2_SDA	-	DFSDM1_CKIN7	USART3_RX	-	-	OTG_HS_ ULPI_D4	-	-	-	LCD_G5	EVENTOUT
	PB12	-	TIM1_BKIN	-	OCTOSPIM_ P1_NCLK	I2C2_SMBA	SPI2_SS/ I2S2_WS	DFSDM1_DATIN1	USART3_CK	-	FDCAN2_RX	OTG_HS_ ULPI_D5	DFSDM2_DATIN1	-	TIM1_BKIN_COMP12	UART5_RX	EVENTOUT
	PB13	-	TIM1_CH1N	-	LPTIM2_OUT	DFSDM2_CKIN1	SPI2_SCK/ I2S2_CK	DFSDM1_CKIN1	USART3_CTS/ USART3_NSS	-	FDCAN2_TX	OTG_HS_ ULPI_D6	-	SDMMC1_D0	DCMI_D2/PSSI_D2	UART5_TX	EVENTOUT
	PB14	-	TIM1_CH2N	TIM12_CH1	TIM8_CH2N	USART1_TX	SPI2_MISO/ I2S2_SDI	DFSDM1_DATIN2	USART3_RTS	UART4_RTS	SDMMC2_D0	-	-	-	-	LCD_CLK	EVENTOUT
	PB15	RTC_REFIN	TIM1_CH3N	TIM12_CH2	TIM8_CH3N	USART1_RX	SPI2_MOSI/ I2S2_SDO	DFSDM1_CKIN2	-	UART4_CTS	SDMMC2_D1	-	-	-	-	LCD_G7	EVENTOUT

Table 10. Port C alternate functions

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	LPTIM1/ TIM1/2/16/17	PDM_SAI1/ TIM3/4/5/12/15	DFSDM1/ LPTIM2/3/ LPUART1/ OCTOSPIM_P1/2/ TIM8	CEC/DCMI/PSSI/ DFSDM1/2/ I2C1/2/3/4/ LPTIM2/TIM15/ USART1	CEC/SPI1/ I2S1/SPI2/ I2S2/SPI3/ I2S3/ SPI4/5/ SPI6/I2S6	DFSDM1/2/I2C4/ OCTOSPIM_P1/ SAI1/SPI3/I2S3/ UART4	SDMMC1/ SPI2/I2S2/ SPI3/I2S3/ SPI6/I2S6/ UART7/ USART1/2/3/6	LPUART1/ SAI2/SDMMC1/ SPDIFRX1/SPI6/ I2S6/UART4/5/8	FDCAN1/2/FMC/LCD /OCTOSPIM_P1/2/ SDMMC2/ SPDIFRX1/TIM13/14	CRS/FMC/LCD/ OCTOSPIM_P1/ OTG1_FS/ OTG1_HS/ SAI2/SDMMC2/ TIM8	DFSDM1/2/ I2C4/LCD/ MDIOS/ OCTOSPIM_P1/ SDMMC2/ SWPMI1/ TIM1/8/ UART7/9/ USART10	FMC/LCD/ MDIOS/ SDMMC1/ TIM1/8	COMP/DCMI/ PSSI/LCD/ TIM1	LCD/UART5	SYS
Port C	PC0	-	-	-	DFSDM1_CKIN0	-	-	DFSDM1_DATIN4	-	SAI2_FS_B	FMC_A25	OTG_HS_ ULPI_STP	LCD_G2	FMC_SDNWE	-	LCD_R5	EVENTOUT
	PC1	TRACED0	-	SAI1_D1	DFSDM1_DATIN0	DFSDM1_CKIN4	SPI2_MOSI/ I2S2_SDO	SAI1_SD_A	-	-	SDMMC2_CK	OCTOSPIM_ P1_IO4	-	MDIOS_MDC	-	LCD_G5	EVENTOUT
	PC2	PWR_CSTOP	-	-	DFSDM1_CKIN1	-	SPI2_MISO/ I2S2_SDI	DFSDM1_CKOUT	-	-	OCTOSPIM_P1_IO2	OTG_HS_ ULPI_DIR	OCTOSPIM_ P1_IO5	FMC_SDNE0	-	-	EVENTOUT
	PC3	PWR_CSLEEP	-	-	DFSDM1_DATIN1	-	SPI2_MOSI/ I2S2_SDO	-	-	-	OCTOSPIM_P1_IO0	OTG_HS_ ULPI_NXT	OCTOSPIM_ P1_IO6	FMC_SDCKE0	-	-	EVENTOUT
	PC4	-	-	-	DFSDM1_CKIN2	-	I2S1_MCK	-	-	-	SPDIFRX1_IN2	-	-	FMC_SDNE0	-	LCD_R7	EVENTOUT
	PC5	-	-	SAI1_D3	DFSDM1_DATIN2	PSSI_D15	-	-	-	-	SPDIFRX1_IN3	OCTOSPIM_ P1_DQS	-	FMC_SDCKE0	COMP1_OUT	LCD_DE	EVENTOUT
	PC6	-	-	TIM3_CH1	TIM8_CH1	DFSDM1_CKIN3	I2S2_MCK	-	USART6_TX	SDMMC1_D0DIR	FMC_NWAIT	SDMMC2_D6	-	SDMMC1_D6	DCMI_D0/ PSSI_D0	LCD_HSYNC	EVENTOUT
	PC7	TRGIO	-	TIM3_CH2	TIM8_CH2	DFSDM1_DATIN3	-	I2S3_MCK	USART6_RX	SDMMC1_D123DIR	FMC_NE1	SDMMC2_D7	SWPMI_TX	SDMMC1_D7	DCMI_D1/ PSSI_D1	LCD_G6	EVENTOUT
	PC8	TRACED1	-	TIM3_CH3	TIM8_CH3	-	-	-	USART6_CK	UART5_RTS	FMC_NE2/ FMC_NCE	FMC_INT	SWPMI_RX	SDMMC1_D0	DCMI_D2/ PSSI_D2	-	EVENTOUT
	PC9	MCO2	-	TIM3_CH4	TIM8_CH4	I2C3_SDA	I2S_CKIN	-	-	UART5_CTS	OCTOSPIM_P1_IO0	LCD_G3	SWPMI_ SUSPEND	SDMMC1_D1	DCMI_D3/ PSSI_D3	LCD_B2	EVENTOUT
	PC10	-	-	-	DFSDM1_CKIN5	DFSDM2_CKIN0	-	SPI3_SCK/ I2S3_CK	USART3_TX	UART4_TX	OCTOSPIM_P1_IO1	LCD_B1	SWPMI_RX	SDMMC1_D2	DCMI_D8/ PSSI_D8	LCD_R2	EVENTOUT
	PC11	-	-	-	DFSDM1_DATIN5	DFSDM2_DATIN0	-	SPI3_MISO/ I2S3_SDI	USART3_RX	UART4_RX	OCTOSPIM_P1_NCS	-	-	SDMMC1_D3	DCMI_D4/ PSSI_D4	LCD_B4	EVENTOUT
	PC12	TRACED3	-	TIM15_CH1	-	DFSDM2_CKOUT	SPI6_SCK/ I2S6_CK	SPI3_MOSI/ I2S3_SDO	USART3_CK	UART5_TX	-	-	-	SDMMC1_CK	DCMI_D9/ PSSI_D9	LCD_R6	EVENTOUT
	PC13	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENTOUT
	PC14	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENTOUT
	PC15	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENTOUT

Table 11. Port D alternate functions

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	LPTIM1/ TIM1/2/16/17	PDM_SAI1/ TIM3/4/5/12/15	DFSDM1/ LPTIM2/3/ LPUART1/ OCTOSPIM_P1/2/ TIM8	CEC/DCMI/PSSI/ DFSDM1/2/ I2C1/2/3/4/ LPTIM2/TIM15/ USART1	CEC/SPI1/ I2S1/SPI2/ I2S2/SPI3/ I2S3/ SPI4/5/ SPI6/I2S6	DFSDM1/2/I2C4/ OCTOSPIM_P1/ SAI1/SPI3/I2S3/ UART4	SDMMC1/ SPI2/I2S2/ SPI3/I2S3/ SPI6/I2S6/ UART7/ USART1/2/3/6	LPUART1/ SAI2/ SDMMC1/ SPDIFRX1/ SPI6/I2S6/ UART4/5/8	FDCAN1/2/FMC/LCD/OCTOSPIM_P1/2/ SDMMC2/ SPDIFRX1/ TIM13/14	CRS/FMC/LCD/ OCTOSPIM_P1/ OTG1_FS/ OTG1_HS/SAI2/ SDMMC2/TIM8	DFSDM1/2/ I2C4/LCD/ MDIOS/ OCTOSPIM_P1/ SDMMC2/ SWPMI1/ TIM1/8/ UART7/9/ USART10	FMC/LCD/ MDIOS/ SDMMC1/ TIM1/8	COMP/ DCMI/ PSSI/LCD/ TIM1	LCDUART5	SYS
Port D	PD0	-	-	-	DFSDM1_CKIN6	-	-	-	-	UART4_RX	FDCAN1_RX	-	UART9_CTS	FMC_D2/ FMC_DA2	-	LCD_B1	EVENTOUT
	PD1	-	-	-	DFSDM1_DATIN6	-	-	-	-	UART4_TX	FDCAN1_TX	-	-	FMC_D3/ FMC_DA3	-	-	EVENTOUT
	PD2	TRACED2	-	TIM3_ETR	-	TIM15_BKIN	-	-	-	UART5_RX	LCD_B7	-	-	SDMMC1_CMD	DCMI_D11/ PSSI_D11	LCD_B2	EVENTOUT
	PD3	-	-	-	DFSDM1_CKOUT	-	SPI2_SCK/ I2S2_CK	-	USART2_CTS/ USART2_NSS	-	-	-	-	FMC_CLK	DCMI_D5/ PSSI_D5	LCD_G7	EVENTOUT
	PD4	-	-	-	-	-	-	-	USART2_RTS	-	-	OCTOSPIM_P1_IO4	-	FMC_NOE	-	-	EVENTOUT
	PD5	-	-	-	-	-	-	-	USART2_TX	-	-	OCTOSPIM_P1_IO5	-	FMC_NWE	-	-	EVENTOUT
	PD6	-	-	SAI1_D1	DFSDM1_CKIN4	DFSDM1_DATIN1	SPI3_MOSI/ I2S3_SDO	SAI1_SD_A	USART2_RX	-	-	OCTOSPIM_P1_IO6	SDMMC2_CK	FMC_NWAIT	DCMI_D10/ PSSI_D10	LCD_B2	EVENTOUT
	PD7	-	-	-	DFSDM1_DATIN4	-	SPI1_MOSI/ I2S1_SDO	DFSDM1_CKIN1	USART2_CK	-	SPDIFRX1_IN0	OCTOSPIM_P1_IO7	SDMMC2_CMD	FMC_NE1	-	-	EVENTOUT
	PD8	-	-	-	DFSDM1_CKIN3	-	-	-	USART3_TX	-	SPDIFRX1_IN1	-	-	FMC_D13/ FMC_DA13	-	-	EVENTOUT
	PD9	-	-	-	DFSDM1_DATIN3	-	-	-	USART3_RX	-	-	-	-	FMC_D14/ FMC_DA14	-	-	EVENTOUT
	PD10	-	-	-	DFSDM1_CKOUT	DFSDM2_CKOUT	-	-	USART3_CK	-	-	-	-	FMC_D15/ FMC_DA15	-	LCD_B3	EVENTOUT
	PD11	-	-	-	LPTIM2_IN2	I2C4_SMBA	-	-	USART3_CTS/ USART3_NSS	-	OCTOSPIM_P1_IO0	SAI2_SD_A	-	FMC_A16/ FMC_CLE	-	-	EVENTOUT
	PD12	-	LPTIM1_IN1	TIM4_CH1	LPTIM2_IN1	I2C4_SCL	-	-	USART3_RTS	-	OCTOSPIM_P1_IO1	SAI2_FS_A	-	FMC_A17/ FMC_ALE	DCMI_D12/ PSSI_D12	-	EVENTOUT
	PD13	-	LPTIM1_OUT	TIM4_CH2	-	I2C4_SDA	-	-	-	-	OCTOSPIM_P1_IO3	SAI2_SCK_A	UART9_RTS	FMC_A18	DCMI_D13/ PSSI_D13	-	EVENTOUT
	PD14	-	-	TIM4_CH3	-	-	-	-	-	UART8_CTS	-	-	UART9_RX	FMC_D0/ FMC_DA0	-	-	EVENTOUT
	PD15	-	-	TIM4_CH4	-	-	-	-	-	UART8_RTS	-	-	UART9_TX	FMC_D1/ FMC_DA1	-	-	EVENTOUT

Table 12. Port E alternate functions

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	LPTIM1/ TIM1/2/16/17	PDM_SAI1/ TIM3/4/5/12/15	DFSDM1/ LPTIM2/3/ LPUART1/ OCTOSPIM_P1/2/ TIM8	CEC/DCMI/ PSSI/ DFSDM1/2/ I2C1/2/3/4/ LPTIM2/ TIM15/ USART1	CEC/SPI1/ I2S1/SPI2/ I2S2/SPI3/ I2S3/ SPI4/5/ SPI6/I2S6	DFSDM1/2/ I2C4/ OCTOSPIM_P1/ SAI1/SPI3/I2S3/ UART4	SDMMC1/ SPI2/I2S2/ SPI3/I2S3/ SPI6/I2S6/ UART7/ USART1/2/3/6	LPUART1/ SAI2/ SDMMC1/ SPDIFRX1/ SPI6/I2S6/ UART4/5/8	FDCAN1/2/FMC/LCD/OCTOSPIM_P1/2/ SDMMC2/ SPDIFRX1/ TIM13/14	CRS/FMC/LCD/ OCTOSPIM_P1/ OTG1_FS/ OTG1_HS/SAI2/ SDMMC2/TIM8	DFSDM1/2/ I2C4/LCD/MDIOS/ OCTOSPIM_P1/ SDMMC2/SWPMI1/ TIM1/8/UART7/9/ USART10	FMC/LCD/ MDIOS/ SDMMC1/ TIM1/8	COMP/DCMI/ PSSI/LCD/TIM1	LCD/ UART5	SYS
Port E	PE0	-	LPTIM1_ETR	TIM4_ETR	-	LPTIM2_ETR	-	-	-	UART8_Rx	-	SAI2_MCK_A	-	FMC_NBL0	DCMI_D2/ PSSI_D2	LCD_R0	EVENTOUT
	PE1	-	LPTIM1_IN2	-	-	-	-	-	-	UART8_Tx	-	-	-	FMC_NBL1	DCMI_D3/ PSSI_D3	LCD_R6	EVENTOUT
	PE2	TRACECLK	-	SAI1_CK1	-	-	SPI4_SCK	SAI1_MCLK_A	-	-	OCTOSPIM_P1_IO2	-	USART10_RX	FMC_A23	-	-	EVENTOUT
	PE3	TRACED0	-	-	-	TIM15_BKIN	-	SAI1_SD_B	-	-	-	-	USART10_TX	FMC_A19	-	-	EVENTOUT
	PE4	TRACED1	-	SAI1_D2	DFSDM1_DATIN3	TIM15_CH1N	SPI4_SS	SAI1_FS_A	-	-	-	-	-	FMC_A20	DCMI_D4/ PSSI_D4	LCD_B0	EVENTOUT
	PE5	TRACED2	-	SAI1_CK2	DFSDM1_CKIN3	TIM15_CH1	SPI4_MISO	SAI1_SCK_A	-	-	-	-	-	FMC_A21	DCMI_D6/ PSSI_D6	LCD_G0	EVENTOUT
	PE6	TRACED3	TIM1_BKIN2	SAI1_D1	-	TIM15_CH2	SPI4_MOSI	SAI1_SD_A	-	-	-	SAI2_MCK_B	TIM1_BKIN2_ COMP12	FMC_A22	DCMI_D7/ PSSI_D7	LCD_G1	EVENTOUT
	PE7	-	TIM1_ETR	-	DFSDM1_DATIN2	-	-	-	UART7_RX	-	-	OCTOSPIM_P1_IO4	-	FMC_D4/ FMC_DA4	-	-	EVENTOUT
	PE8	-	TIM1_CH1N	-	DFSDM1_CKIN2	-	-	-	UART7_TX	-	-	OCTOSPIM_P1_IO5	-	FMC_D5/ FMC_DA5	COMP2_OUT	-	EVENTOUT
	PE9	-	TIM1_CH1	-	DFSDM1_CKOUT	-	-	-	UART7_RTS	-	-	OCTOSPIM_P1_IO6	-	FMC_D6/ FMC_DA6	-	-	EVENTOUT
	PE10	-	TIM1_CH2N	-	DFSDM1_DATIN4	-	-	-	UART7_CTS	-	-	OCTOSPIM_P1_IO7	-	FMC_D7/ FMC_DA7	-	-	EVENTOUT
	PE11	-	TIM1_CH2	-	DFSDM1_CKIN4	-	SPI4_SS	-	-	-	-	SAI2_SD_B	OCTOSPIM_P1_NCS	FMC_D8/ FMC_DA8	-	LCD_G3	EVENTOUT
	PE12	-	TIM1_CH3N	-	DFSDM1_DATIN5	-	SPI4_SCK	-	-	-	-	SAI2_SCK_B	-	FMC_D9/ FMC_DA9	COMP1_ OUT	LCD_B4	EVENTOUT
	PE13	-	TIM1_CH3	-	DFSDM1_CKIN5	-	SPI4_MISO	-	-	-	-	SAI2_FS_B	-	FMC_D10/ FMC_DA10	COMP2_OUT	LCD_DE	EVENTOUT
	PE14	-	TIM1_CH4	-		-	SPI4_MOSI	-	-	-	-	SAI2_MCK_B	-	FMC_D11/ FMC_DA11	-	LCD_CLK	EVENTOUT
	PE15	-	TIM1_BKIN	-		-	-	-	-	-	-	-	USART10_CK	FMC_D12/ FMC_DA12	TIM1_BKIN_COMP12	LCD_R7	EVENTOUT

Table 13. Port F alternate functions

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	LPTIM1/ TIM1/2/16/17	PDM_SAI1/ TIM3/4/5/12/15	DFSDM1/ LPTIM2/3/ LPUART1/ OCTOSPIM_P1/2/ TIM8	CEC/DCMI/ PSSI/ DFSDM1/2/ I2C1/2/3/4/ LPTIM2/ TIM15/ USART1	CEC/SPI1/ I2S1/SPI2/ I2S2/SPI3/ I2S3/SPI4/5/ SPI6/I2S6	DFSDM1/2/I2C4/ OCTOSPIM_P1/ SAI1/SPI3/I2S3/ UART4	SDMMC1/SPI2/ I2S2/SPI3/ I2S3/SPI6/ I2S6/UART7/ USART1/2/3/6	LPUART1/ SAI2/ SDMMC1/ SPDIFRX1/ SPI6/I2S6/ UART4/5/8	FDCAN1/2/FMC/LCD/ OCTOSPIM_P1/2/ SDMMC2/SPDIFRX1/ TIM13/14	CRS/FMC/LCD/ OCTOSPIM_P1/ OTG1_FS/OTG1_HS/ SAI2/SDMMC2/TIM8	DFSDM1/2/ I2C4/LCD/ MDIOS/ OCTOSPIM_P1/ SDMMC2/ SWPMI1/TIM1/8/ UART7/9/ USART10	FMC/LCD/ MDIOS/ SDMMC1/ TIM1/8	COMP/ DCMI/ PSSI/LCD/ TIM1	LCD/ UART5	SYS
Port F	PF0	-	-	-	-	I2C2_SDA	-	-	-	-	OCTOSPIM_P2_IO0	-	-	FMC_A0	-	-	EVENTOUT
	PF1	-	-	-	-	I2C2_SCL	-	-	-	-	OCTOSPIM_P2_IO1	-	-	FMC_A1	-	-	EVENTOUT
	PF2	-	-	-	-	I2C2_SMBA	-	-	-	-	OCTOSPIM_P2_IO2	-	-	FMC_A2	-	-	EVENTOUT
	PF3	-	-	-	-	-	-	-	-	-	OCTOSPIM_P2_IO3	-	-	FMC_A3	-	-	EVENTOUT
	PF4	-	-	-	-	-	-	-	-	-	OCTOSPIM_P2_CLK	-	-	FMC_A4	-	-	EVENTOUT
	PF5	-	-	-	-	-	-	-	-	-	OCTOSPIM_ P2_NCLK	-	-	FMC_A5	-	-	EVENTOUT
	PF6	-	TIM16_CH1	-	-	-	SPI5_SS	SAI1_SD_B	UART7_Rx	-	-	OCTOSPIM_P1_IO3	-	-	-	-	EVENTOUT
	PF7	-	TIM17_CH1	-	-	-	SPI5_SCK	SAI1_MCLK_B	UART7_Tx	-	-	OCTOSPIM_P1_IO2	-	-	-	-	EVENTOUT
	PF8	-	TIM16_CH1N	-	-	-	SPI5_MISO	SAI1_SCK_B	UART7_RTS	-	TIM13_CH1	OCTOSPIM_P1_IO0	-	-	-	-	EVENTOUT
	PF9	-	TIM17_CH1N	-	-	-	SPI5_MOSI	SAI1_FS_B	UART7_CTS	-	TIM14_CH1	OCTOSPIM_P1_IO1	-	-	-	-	EVENTOUT
	PF10	-	TIM16_BKIN	SAI1_D3	-	PSSI_D15	-	-	-	-	OCTOSPIM_P1_CLK	-	-	-	DCMI_D11/ PSSI_D11	LCD_DE	EVENTOUT
	PF11	-	-	-	-	-	SPI5_MOSI	-	-	-	OCTOSPIM_ P1_NCLK	SAI2_SD_B	-	FMC_SDNRAS	DCMI_D12/ PSSI_D12	-	EVENTOUT
	PF12	-	-	-	-	-	-	-	-	-	OCTOSPIM_ P2_DQS	-	-	FMC_A6	-	-	EVENTOUT
	PF13	-	-	-	DFSDM1_DATIN6	I2C4_SMBA	-	-	-	-	-	-	-	FMC_A7	-	-	EVENTOUT
	PF14	-	-	-	DFSDM1_CKIN6	I2C4_SCL	-	-	-	-	-	-	-	FMC_A8	-	-	EVENTOUT
	PF15	-	-	-	-	I2C4_SDA	-	-	-	-	-	-	-	FMC_A9	-	-	EVENTOUT

Table 14. Port G alternate functions

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	LPTIM1/ TIM1/2/16/17	PDM_SAI1/ TIM3/4/5/12/15	DFSDM1/LPTIM2/3/ LPUART1/ OCTOSPIM_P1/2/ TIM8	CEC/DCMI/ PSSI/ DFSDM1/2/ I2C1/2/3/4/ LPTIM2/ TIM15/ USART1	CEC/SPI1/ I2S1/SPI2/ I2S2/SPI3/ I2S3/ SPI4/5/ SPI6/I2S6	DFSDM1/2/ I2C4/ OCTOSPIM_P1/ SAI1/SPI3/I2S3/ UART4	SDMMC1/ SPI2/I2S2/ SPI3/I2S3/ SPI6/I2S6/ UART7/ USART1/2/3/6	LPUART1/ SAI2/ SDMMC1/ SPDIFRX1/ SPI6/I2S6/ UART4/5/8	FDCAN1/2/FMC/LCD /OCTOSPIM_P1/2/ SDMMC2/SPDIFRX1/ TIM13/14	CRS/FMC/LCD/ OCTOSPIM_P1/ OTG1_FS/ OTG1_HS/SAI2/ SDMMC2/TIM8	DFSDM1/2/ I2C4/LCD/ MDIOS/ OCTOSPIM_P1/ SDMMC2/ SWPMI1/ TIM1/8/ UART7/9/ USART10	FMC/LCD/ MDIOS/ SDMMC1/ TIM1/8	COMP/DCMI/ PSSI/LCD/ TIM1	LCD/ UART5	SYS
Port G	PG0	-	-	-	-	-	-	-	-	-	OCTOSPIM_P2_IO4	-	UART9_RX	FMC_A10	-	-	EVENTOUT
	PG1	-	-	-	-	-	-	-	-	-	OCTOSPIM_P2_IO5	-	UART9_TX	FMC_A11	-	-	EVENTOUT
	PG2	-	-	-	TIM8_BKIN	-	-	-	-	-	-	-	TIM8_BKIN_ COMP12	FMC_A12	-	-	EVENTOUT
	PG3	-	-	-	TIM8_BKIN2	-	-	-	-	-	-	-	TIM8_BKIN2_ COMP12	FMC_A13	-	-	EVENTOUT
	PG4	-	TIM1_BKIN2	-	-	-	-	-	-	-	-	-	TIM1_BKIN2_ COMP12	FMC_A14/ FMC_BA0	-	-	EVENTOUT
	PG5	-	TIM1_ETR	-	-	-	-	-	-	-	-	-	-	FMC_A15/ FMC_BA1	-	-	EVENTOUT
	PG6	-	TIM17_BKIN	-	-	-	-	-	-	-	-	OCTOSPIM_P1_NCS	-	FMC_NE3	DCMI_D12/ PSSI_D12	LCD_R7	EVENTOUT
	PG7	-	-	-	-	-	-	SAI1_MCLK_A	USART6_CK	-	OCTOSPIM_P2_DQS	-	-	FMC_INT	DCMI_D13/ PSSI_D13	LCD_CLK	EVENTOUT
	PG8	-	-	-	TIM8_ETR	-	SPI6_SS/ I2S6_WS	-	USART6_RTS	SPDIFRX1_IN2	-	-	-	FMC_SDCLK	-	LCD_G7	EVENTOUT
	PG9	-	-	-	-	-	SPI1_MISO/ I2S1_SDI	-	USART6_RX	SPDIFRX1_IN3	OCTOSPIM_P1_IO6	SAI2_FS_B	SDMMC2_D0	FMC_NE2/ FMC_NCE	DCMI_VSYNC/ PSSI_RDY	-	EVENTOUT
	PG10	-	-	-	OCTOSPIM_P2_IO6	-	SPI1_SS/ I2S1_WS	-	-	-	LCD_G3	SAI2_SD_B	SDMMC2_D1	FMC_NE3	DCMI_D2/ PSSI_D2	LCD_B2	EVENTOUT
	PG11	-	LPTIM1_IN2	-	-	-	SPI1_SCK/ I2S1_CK	-	-	SPDIFRX1_IN0	OCTOSPIM_P2_IO7	SDMMC2_D2	USART10_RX	-	DCMI_D3/ PSSI_D3	LCD_B3	EVENTOUT
	PG12	-	LPTIM1_IN1	-	OCTOSPIM_P2_NCS	-	SPI6_MISO/ I2S6_SDI	-	USART6_RTS	SPDIFRX1_IN1	LCD_B4	SDMMC2_D3	USART10_TX	-	-	LCD_B1	EVENTOUT
	PG13	TRACED0	LPTIM1_OUT	-	-	-	SPI6_SCK/ I2S6_CK	-	USART6_CTS/ USART6_NSS	-	-	SDMMC2_D6	USART10_CTS/ USART10_NSS	-	-	LCD_R0	EVENTOUT
	PG14	TRACED1	LPTIM1_ETR	-	-	-	SPI6_MOSI/ I2S6_SDO	-	USART6_TX	-	OCTOSPIM_P1_IO7	SDMMC2_D7	USART10_RTS	-	-	LCD_B0	EVENTOUT
	PG15	-	-	-	-	-	-	-	USART6_CTS/ USART6_NSS	-	OCTOSPIM_P2_DQS	-	-	-	DCMI_D13/ PSSI_D13	-	EVENTOUT

Table 15. Port H alternate functions

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	LPTIM1/ TIM1/2/16/17	PDM_SAI1/ TIM3/4/5/12/15	DFSDM1/ LPTIM2/3/ LPUART1/ OCTOSPIM_P1/2/ TIM8	CEC/DCMI/ PSSI/ DFSDM1/2/ I2C1/2/3/4/ LPTIM2/ TIM15/ USART1	CEC/SPI1/ I2S1/SPI2/ I2S2/SPI3/ I2S3/SPI4/5/ SPI6/I2S6	DFSDM1/2/I2C4/ OCTOSPIM_P1/ SAI1/SPI3/I2S3/ UART4	SDMMC1/SPI2/ I2S2/SPI3/I2S3/ SPI6/I2S6/ UART7/ USART1/2/3/6	LPUART1/ SAI2/ SDMMC1/ SPDIFRX1/ SPI6/I2S6/ UART4/5/8	FDCAN1/2/FMC/LCD /OCTOSPIM_P1/2/ SDMMC2/SPDIFRX1/ TIM13/14	CRS/FMC/LCD/ OCTOSPIM_P1/ OTG1_FS/ OTG1_HS/SAI2/ SDMMC2/TIM8	DFSDM1/2/ I2C4/LCD/ MDIOS/ OCTOSPIM_P1/ SDMMC2/ SWPMI1/TIM1/8/ UART7/9/ USART10	FMC/LCD/ MDIOS/ SDMMC1/ TIM1/8	COMP/DCMI/ PSSI/LCD/TIM1	LCD/ UART5	SYS
Port H	PH0	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENTOUT
	PH1	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENTOUT
	PH2	-	LPTIM1_IN2	-	-	-	-	-	-	-	OCTOSPIM_P1_IO4	SAI2_SCK_B	-	FMC_SDCKE0	-	LCD_R0	EVENTOUT
	PH3	-	-	-	-	-	-	-	-	-	OCTOSPIM_P1_IO5	SAI2_MCK_B	-	FMC_SDNE0	-	LCD_R1	EVENTOUT
	PH4	-	-	-	-	I2C2_SCL	-	-	-	-	LCD_G5	OTG_HS_ ULPI_NXT	-	-	PSSI_D14	LCD_G4	EVENTOUT
	PH5	-	-	-	-	I2C2_SDA	SPI5_SS	-	-	-	-	-	-	FMC_SDNWE	-	-	EVENTOUT
	PH6	-	-	TIM12_CH1	-	I2C2_SMBA	SPI5_SCK	-	-	-	-	-	-	FMC_SDNE1	DCMI_D8/ PSSI_D8	-	EVENTOUT
	PH7	-	-	-	-	I2C3_SCL	SPI5_MISO	-	-	-	-	-	-	FMC_SDCKE1	DCMI_D9/ PSSI_D9	-	EVENTOUT
	PH8	-	-	TIM5_ETR	-	I2C3_SDA	-	-	-	-	-	-	-	FMC_D16	DCMI_HSYNC/ PSSI_DE	LCD_R2	EVENTOUT
	PH9	-	-	TIM12_CH2	-	I2C3_SMBA	-	-	-	-	-	-	-	FMC_D17	DCMI_D0/ PSSI_D0	LCD_R3	EVENTOUT
	PH10	-	-	TIM5_CH1	-	I2C4_SMBA	-	-	-	-	-	-	-	FMC_D18	DCMI_D1/ PSSI_D1	LCD_R4	EVENTOUT
	PH11	-	-	TIM5_CH2	-	I2C4_SCL	-	-	-	-	-	-	-	FMC_D19	DCMI_D2/ PSSI_D2	LCD_R5	EVENTOUT
	PH12	-	-	TIM5_CH3	-	I2C4_SDA	-	-	-	-	-	-	-	FMC_D20	DCMI_D3/ PSSI_D3	LCD_R6	EVENTOUT
	PH13	-	-	-	TIM8_CH1N	-	-	-	-	UART4_TX	FDCAN1_TX	-	-	FMC_D21	-	LCD_G2	EVENTOUT
	PH14	-	-	-	TIM8_CH2N	-	-	-	-	UART4_RX	FDCAN1_RX	-	-	FMC_D22	DCMI_D4/ PSSI_D4	LCD_G3	EVENTOUT
	PH15	-	-	-	TIM8_CH3N	-	-	-	-	-	-	-	-	FMC_D23	DCMI_D11/ PSSI_D11	LCD_G4	EVENTOUT

Table 16. Port I alternate functions

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	LPTIM1/ TIM1/2/16/17	PDM_SAI1/ TIM3/4/5/12/15	DFSDM1/LPTIM2/3/ LPUART1/ OCTOSPIM_P1/2/ TIM8	CEC/DCMI/ PSSI/ DFSDM1/2/ I2C1/2/3/4/ LPTIM2/ TIM15/ USART1	CEC/SPI1/ I2S1/SPI2/ I2S2/SPI3/ I2S3/SPI4/5/ SPI6/I2S6	DFSDM1/2/I2C4/ OCTOSPIM_P1/ SAI1/SPI3/I2S3/ UART4	SDMMC1/SPI2/ I2S2/SPI3/ I2S3/SPI6/ I2S6/UART7/ USART1/2/3/6	LPUART1/ SAI2/ SDMMC1/ SPDIFRX1/ SPI6/I2S6/ UART4/5/8	FDCAN1/2/FMC/LCD/ OCTOSPIM_P1/2/ SDMMC2/ SPDIFRX1/ TIM13/14	CRS/FMC/LCD/ OCTOSPIM_P1/ OTG1_FS/ OTG1_HS/SAI2/ SDMMC2/TIM8	DFSDM1/2/I2C4/LCD/ MDIOS/OCTOSPIM_P1/ SDMMC2/SWPMI1/ TIM1/8/UART7/9/ USART10	FMC/LCD/ MDIOS/ SDMMC1/ TIM1/8	COMP/DCMI/ PSSI/LCD/TIM1	LCD/ UART5	SYS
Port I	PI0	-	-	TIM5_CH4	-	-	SPI2_SS/ I2S2_WS	-	-	-	-	-	-	FMC_D24	DCMI_D13/ PSSI_D13	LCD_G5	EVENTOUT
	PI1	-	-	-	TIM8_BKIN2	-	SPI2_SCK/ I2S2_CK	-	-	-	-	-	TIM8_BKIN2_COMP12	FMC_D25	DCMI_D8/ PSSI_D8	LCD_G6	EVENTOUT
	PI2	-	-	-	TIM8_CH4	-	SPI2_MISO/ I2S2_SDI	-	-	-	-	-	-	FMC_D26	DCMI_D9/ PSSI_D9	LCD_G7	EVENTOUT
	PI3	-	-	-	TIM8_ETR	-	SPI2_MOSI/ I2S2_SDO	-	-	-	-	-	-	FMC_D27	DCMI_D10/ PSSI_D10	-	EVENTOUT
	PI4	-	-	-	TIM8_BKIN	-	-	-	-	-	-	SAI2_MCK_A	TIM8_BKIN_COMP12	FMC_NBL2	DCMI_D5/ PSSI_D5	LCD_B4	EVENTOUT
	PI5	-	-	-	TIM8_CH1	-	-	-	-	-	-	SAI2_SCK_A	-	FMC_NBL3	DCMI_VSYNC/ PSSI_ RDY	LCD_B5	EVENTOUT
	PI6	-	-	-	TIM8_CH2	-	-	-	-	-	-	SAI2_SD_A	-	FMC_D28	DCMI_D6/ PSSI_D6	LCD_B6	EVENTOUT
	PI7	-	-	-	TIM8_CH3	-	-	-	-	-	-	SAI2_FS_A	-	FMC_D29	DCMI_D7/ PSSI_D7	LCD_B7	EVENTOUT
	PI8	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVENTOUT
	PI9	-	-	-	OCTOSPIM_P2_IO0	-	-	-	-	UART4_RX	FDCAN1_RX	-	-	FMC_D30	-	LCD_ VSYNC	EVENTOUT
	PI10	-	-	-	OCTOSPIM_P2_IO1	-	-	-	-	-	-	-	-	FMC_D31	PSSI_D14	LCD_ HSYNC	EVENTOUT
	PI11	-	-	-	OCTOSPIM_P2_IO2	-	-	-	-	-	LCD_G6	OTG_HS_ ULPI_DIR	-	-	PSSI_D15	-	EVENTOUT

6 Electrical characteristics

6.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V_{SS}.

6.1.1 Minimum and maximum values

Unless otherwise specified the minimum and maximum values are guaranteed in the worst conditions of junction temperature, supply voltage and frequencies by tests in production on 100% of the devices with an junction temperature at T_J = 25 °C and T_J = T_{Jmax} (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean±3σ).

6.1.2 Typical values

Unless otherwise specified, typical data are based on T_J = 25 °C, V_{DD} = 3.3 V (for the 1.62 V ≤ V_{DD} ≤ 3.6 V voltage range). They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95% of the devices have an error less than or equal to the value indicated (mean±2σ).

6.1.3 Typical curves

Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

6.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in Figure 10. Pin loading conditions.

6.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in Figure 11. Pin input voltage.

Figure 10. Pin loading conditions

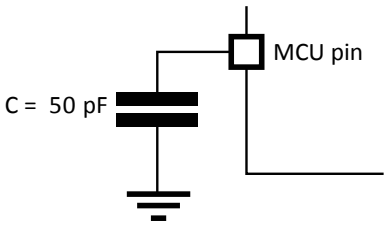
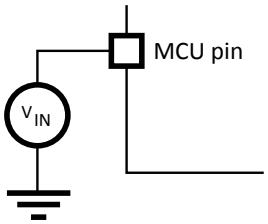
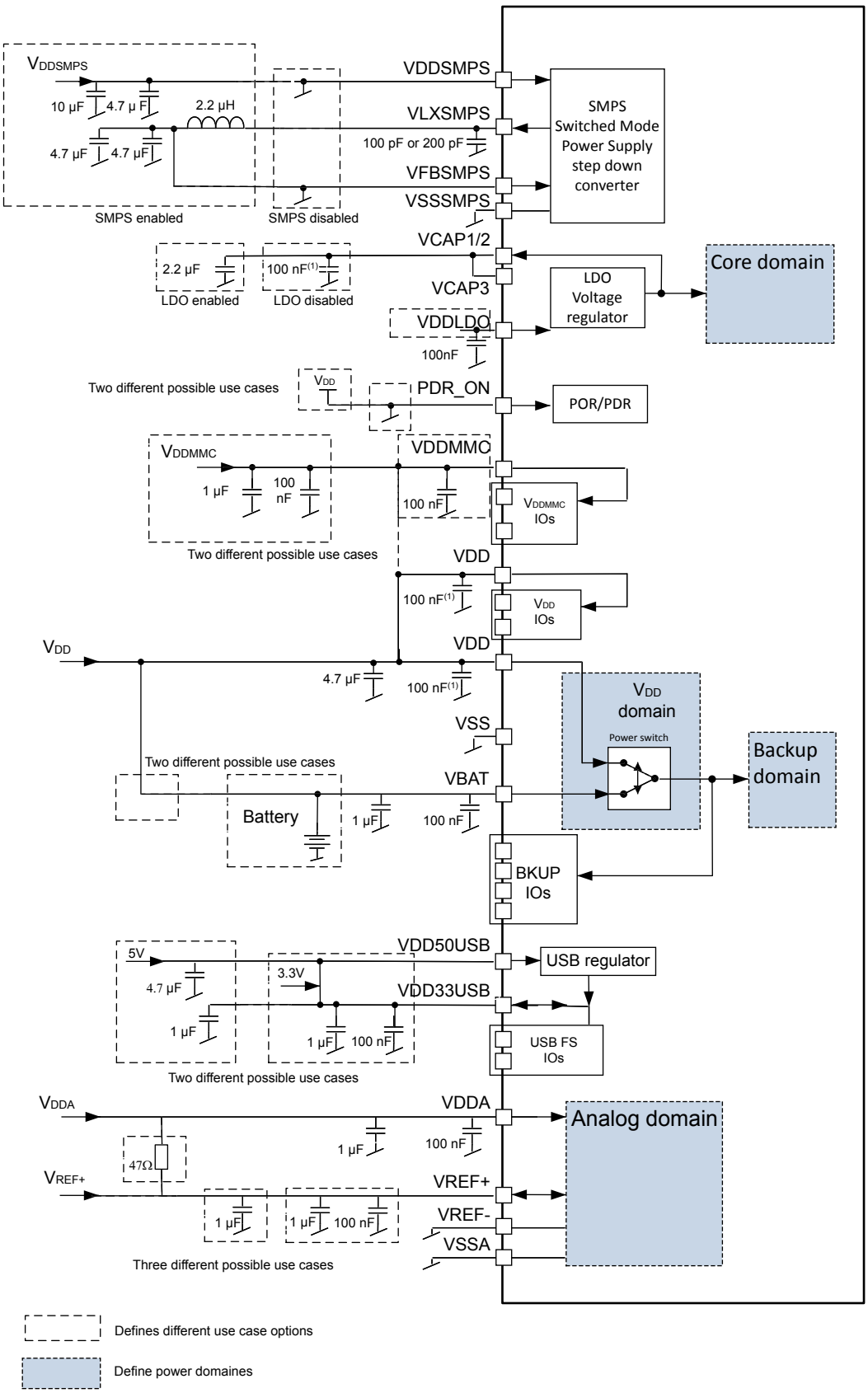


Figure 11. Pin input voltage



6.1.6 Power supply scheme

Figure 12. **Power supply scheme**



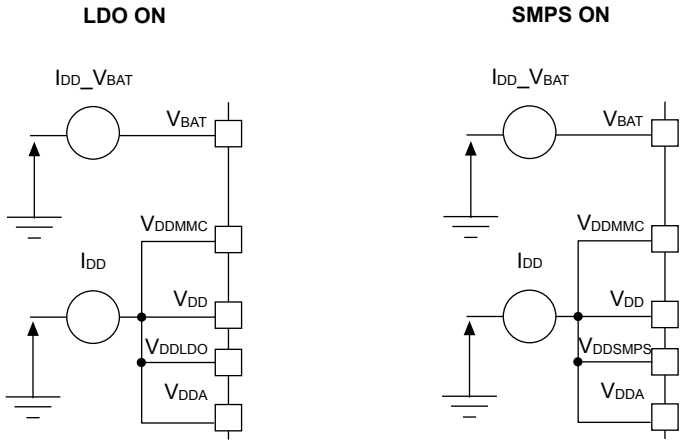
1. 100 nF filtering capacitor on each package pin.
2. A tolerance of +/- 20% is acceptable on decoupling capacitors.

Note: Refer to Getting started with STM32H7A3/7B3 and STM32H7B0 hardware development(AN5307) for more details.

Caution: Each power supply pair (V_{DD}/V_{SS}, V_{DDA}/V_{SSA} ...) must be decoupled with filtering ceramic capacitors as shown above. These capacitors must be placed as close as possible to, or below, the appropriate pins on the underside of the PCB to ensure good operation of the device. It is not recommended to remove filtering capacitors to reduce PCB size or cost. This might cause incorrect operation of the device.

6.1.7 Current consumption measurement

Figure 13. Current consumption measurement scheme



6.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in Table 17. Voltage characteristics, Table 18. Current characteristics, and Table 19. Thermal characteristics may cause permanent damage to the device. These are stress ratings only and the functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability. Device mission profile (application conditions) is compliant with JEDEC JESD47 Qualification Standard, extended mission profiles are available on demand.

Table 17. Voltage characteristics

All main power (V_{DD}, V_{DDA}, $V_{DD33USB}$, V_{DDMMC}, V_{DDSMPS}, V_{BAT}) and ground (V_{SS}, V_{SSA}) pins must always be connected to the external power supply, in the permitted range.

Symbols	Ratings	Min	Max	Unit
$V_{DDX} - V_{SS}$	External main supply voltage (including V_{DD}, V_{DDLDO}, V_{DDSMPS}, V_{DDA}, $V_{DD33USB}$, V_{DDMMC}, V_{BAT}, V_{REF+})	−0.3	4.0	V
V_{IN}(1)	Input voltage on FT_xxx pins	V_{SS}−0.3	Min(V_{DD}, V_{DDA}, $V_{DD33USB}$, V_{DDMMC}, V_{BAT}) +4.0(2)(3)	V
	Input voltage on TT_xx pins	V_{SS}−0.3	4.0	V
	Input voltage on BOOT0 pin	V_{SS}	9.0	V
	Input voltage on any other pins	V_{SS}−0.3	4.0	V
$\lvert\Delta V_{DDX}\rvert$	Variations between different V_{DDX} power pins of the same domain	-	50	mV

Symbols	Ratings	Min	Max	Unit
$\lvert V_{SSx}-V_{SS}\rvert$	Variations between all the different ground pins	-	50	mV

1. *V_{IN} maximum value must always be respected. Refer to Table 62. I/O current injection susceptibility for the maximum allowed injected current values.*
2. *To sustain a voltage higher than 4 V the internal pull-up/pull-down resistors must be disabled.*
3. *This formula has to be applied on power supplies related to the I/O structure described by the pin definition table.*

Table 18. Current characteristics

Symbols	Ratings	Max	Unit
ΣI_{VDD}	Total current into sum of all V_{DD} power lines (source)(1)	620	mA
ΣI_{VSS}	Total current out of sum of all V_{SS} ground lines (sink)(1)	620	
I_{VDD}	Maximum current into each V_{DD} power pin (source)(1)	100	
I_{VSS}	Maximum current out of each V_{SS} ground pin (sink)(1)	100	
I_{IO}	Output current sunk or sourced by any I/O and control pin	20	
	Output current sunk or sourced by Pxy_C pin	1	
$\Sigma I_{(PIN)}$	Total output current sunk by sum of all I/Os and control pins(2)	140	
	Total output current sourced by sum of all I/Os and control pins(2)	140	
$I_{INJ(PIN)}$(3)(4)	Injected current on FT_xxx, TT_xx, RST and B pins except PA4, PA5	−5/+0	
	Injected current on PA4, PA5	−0/0	
$\Sigma I_{INJ(PIN)}$	Total injected current (sum of all I/Os and control pins)(5)	±25	

1. *All main power (V_{DD}, V_{DDA}, V_{DDSMPS}, V_{DDLDO}, $V_{DD33USB}$, V_{DDMMC}) and ground (V_{SS}, V_{SSA}) pins must always be connected to the external power supplies, in the permitted range.*
2. *This current consumption must be correctly distributed over all I/Os and control pins. The total output current must not be sunk/sourced between two consecutive power supply pins referring to high pin count QFP packages.*
3. *A positive injection is induced by $V_{IN}>V_{DD}$ while a negative injection is induced by $V_{IN}<V_{SS}$. $I_{INJ(PIN)}$ must never be exceeded. Refer also to Table 17. Voltage characteristics for the maximum allowed input voltage values.*
4. *Positive injection is not possible on these I/Os and does not occur for input voltages lower than the specified maximum value.*
5. *When several inputs are submitted to a current injection, the maximum $\sum I_{INJ(PIN)}$ is the absolute sum of the positive and negative injected currents (instantaneous values).*

Table 19. Thermal characteristics

Symbol	Ratings	Value	Unit
T_{STG}	Storage temperature range	−65 to +150	°C
T_{Jmax}	Maximum junction temperature	130(1)	

1. *The junction temperature is limited to 105 °C in the VOS0 voltage range.*

6.3 Operating conditions

6.3.1 General operating conditions

Table 20. General operating conditions

Symbol	Parameter	Operating conditions	Min	Typ	Max	Unit
V_{DD}	Standard operating voltage	-	1.62(1)	-	3.6	V
V_{DDLDO}	Supply voltage for the internal regulator	VDDLDO ≤ VDD	1.62(1)	-	3.6	
			1.2(2)	-	3.6	
V_{DDSMPS}	Supply voltage for the internal SMPS Step-down converter	VDDSMPS = VDD	1.62(1)	-	3.6	
V_{DDMMC}	Standard operating voltage for independent MMC I/Os	Indenpent MMC I/Os used	1.62(1)	-	3.6	
		Independent MMC I/Os not used V_{DDMMC} = V_{DD}	1.62(1)	-	3.6	
$V_{DD33USB}$	Standard operating voltage, USB domain	USB used	3.0	-	3.6	
		USB not used	0	-	3.6	
V_{DDA}	Analog operating voltage	ADC or COMP used	1.62	-	3.6	
		DAC used	1.8	-		
		OPAMP used	2.0	-		
		VREFBUF used	1.8	-		
		ADC, DAC, OPAMP, COMP, VREFBUF not used	0	-		
V_{BAT}	Backup operating voltage	-	1.2	-	3.6	
V_{IN}	I/O Input voltage	TT_xx I/O	−0.3	-	V_{DD}+0.3	
		BOOT0	0	-	9	
		All I/O except BOOT0 and TT_xx	−0.3	-	Min(V_{DD}, V_{DDA}, $V_{DD33USB}$, V_{DDMMC}) +3.6 V < 5.5 V(3)	
V_{CORE}	Internal regulator ON (LDO or SMPS)(4)	VOS3 (max frequency 88 MHz)	0.95	1.0	1.05	
		VOS2 (max frequency 160 MHz)	1.05	1.10	1.15	
		VOS1 (max frequency 225 MHz)	1.15	1.20	1.25	
		VOS0 (max frequency 280 MHz)	1.25	1.30	1.35	
	Regulator OFF: external V_{CORE} voltage must be supplied from external regulator on VCAP pins	VOS3 (max frquency 88 MHz)	0.97	1.0	1.05	
		VOS2 (max frequency 160 MHz)	1.07	1.10	1.15	
		VOS1 (max frequency 225 MHz)	1.17	1.20	1.25	

Symbol	Parameter	Operating conditions	Min	Typ	Max	Unit
V_{CORE}	Regulator OFF: external V_{CORE} voltage must be supplied from external regulator on VCAP pins	VOS0 (max frequency 280 MHz)	1.27	1.30	1.33	V
T_A	Ambient temperature for the suffix 6 version	Maximum power dissipation	–40	-	85	°C
		Low-power dissipation(5)	–40	-	105	
T_J	Junction temperature range	VOS0	–40	-	105	°C
		VOS3, VOS2, VOS1	–40	-	130	

1. *When a reset occurs, the functionality is guaranteed down to V_{PDRmax} or to the specified V_{DDmin} when the PDR is OFF. The PDR can only be switched OFF though the PDR_ON pin that is not available in all packages (refer to Table 7. STM32H7B0xB pin/ball definition)*
2. *Only for power-up sequence when the SMPS step-down converter is configured to supply the LDO.*
3. *This formula has to be applied on power supplies related to the I/O structures described by the pin definition table.*
4. *At startup, the external V_{CORE} voltage must remain higher or equal to 1.10 V before disabling the internal regulator (LDO).*
5. *In low-power dissipation state, T_A can be extended to this range as long as T_J does not exceed T_{Jmax} (see Section 7.7 Thermal characteristics).*

Table 21. Maximum allowed clock frequencies

Symbol (1)(2)	Parameter	VOS0	VOS1	VOS2	VOS3	Unit
f_{CPU}	CPU	280	225	160	88	MHz
f_{ACLK}	AXI	280	225	160	88	
f_{HCLK}	AHB	280	225	160	88	
f_{PCLK}	APB	140	112.5	80	44	
$f_{ltdc_ker_ck}$	LTDC	140	112.5	80	44	
$f_{fmc_ker_ck}$	FMC	280	225	160	88	
$f_{octospi_ker_clk}$	OCTOSPI1/2	280	225	160	88	
$f_{sdmmc_ker_ck}$	SDMMC1/2	280	225	160	88	
f_{DFSDM1_Aclk}	DFSDM1	140	112.5	80	44	
f_{DFSDM1_Clk}		140	112.5	80	44	
f_{DFSDM2_Aclk}	DFSDM2	140	112.5	80	44	
f_{DFSDM2_Clk}		140	112.5	80	44	
$f_{fdcan_ker_ck}$	FDCAN	140	112.5	80	44	
$f_{cec_ker_ck}$	HDMI_CEC	66	66	66	44	
$f_{I2C_ker_ck}$	I2C[1:4]	140	112.5	80	44	
$f_{lptim_ker_ck}$	LPTIM[1:3]	140	112.5	80	44	
$f_{rcc_tim_ker_ck}$	TIM[2:7],TIM[12:14]	280	225	160	88	
$f_{rcc_tim_ker_ck}$	PWM1,PWM8,TIM[15:17]	280	225	160	88	
f_{rng_clk}	RNG	140	112.5	80	44	
$f_{sai_a_ker_ck}$	SAI1	150	150	80	80	
$f_{sai_b_ker_ck}$						
$f_{sai_a_ker_ck}$	SAI2	150	150	80	80	
$f_{sai_b_ker_ck}$						

Symbol (1)(2)	Parameter	VOS0	VOS1	VOS2	VOS3	Unit
$f_{spdifrx_ker_ck}$	SPDIFRX1	280	225	160	88	MHz
$f_{spi_ker_ck}$	SPI[1:6]	280	225	160	88	
$f_{lpuart_ker_ck}$	LPUART1	140	112.5	80	44	
$f_{usart_ker_ck}$	USART1/2/3/6/10	280	225	160	88	
$f_{uart_ker_ck}$	UART4/5/7/8/9	280	225	160	88	
f_{adp_clk}	USBOTG	48	48	48	48	
f_{ulpi_ck}	USB1ULPI	66	66	66	66	
$f_{adc_ker_ck}$	ADC1/2	50	50	50	50	
f_{dac_pclk}	DAC1/2	140	112.5	80	44	
$f_{rtc_ker_ck}$	RTC	1	1	1	1	

1. *Guaranteed by design.*
2. *The maximum kernel clock frequencies can be limited by the maximum peripheral clock frequency (refer each peripheral electrical characteristics).*

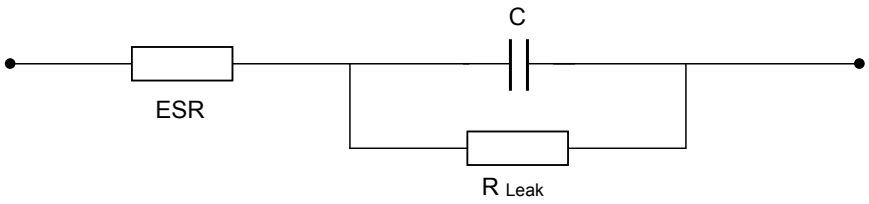
Table 22. Supply voltage and maximum frequency configuration

Power scale	V_{CORE} source	Max T_J (°C)	Max frequency (MHz)	Min V_{DD} (V)
VOS0	LDO/SMPS	105	280	1.71
VOS1	LDO/SMPS	130	225	1.62
VOS2	LDO/SMPS	130	160	1.62
VOS3	LDO/SMPS	130	88	1.62
SVOS4	LDO/SMPS	130	N/A	1.62
SVOS5	LDO/SMPS	130	N/A	1.62

6.3.2 VCAP external capacitor

Stabilization for the embedded LDO regulator is achieved by connecting an external capacitor C_{EXT} to the VCAPx pin. C_{EXT} is specified in Table 23. VCAP operating conditions. Two external capacitors must be connected to VCAP pins (refer to *Getting started with STM32H7A3/7B3 and STM32H7B0 hardware development* (AN5307).

Figure 14. External capacitor C_{EXT}



1. Legend: ESR is the equivalent series resistance.

Table 23. VCAP operating conditions

When the internal LDO voltage regulator is switched OFF, the two 2.2 µF VCAP capacitors are not required. However all VCAPx package pins must be connected together and it is recommended to add a ceramic filtering capacitor of 100 nF as close as possible to each VCAPx pin.

Symbol	Parameter	Conditions
CEXT	External capacitor for LDO enabled	2.2 µF(1)(2)
ESR	ESR of external capacitor	< 100 mΩ

1. *This value corresponds to CEXT typical value. A variation of ±20% is tolerated.*
2. *If the VCAP3 pin is available (depending on the package), it must be connected to the other VCAP pins. No additional capacitor is required.*

6.3.3 SMPS step-down converter

The devices embed a high power efficiency SMPS step-down converter requiring external components. Refer to *Getting started with STM32H7A3/7B3 and STM32H7B0 hardware development* (AN5307) for the required components and tradeoffs.

Table 24. Characteristics of SMPS step-down converter external components

Symbol	Parameter	Conditions
C_{IN}	Capacitance of external capacitor on VDDSMPS	4.7 µF
	ESR of external capacitor	100 mΩ
C_{filt}	Capacitance of external capacitor on VLXSMPS pin	220 pF
C_{OUT}	Capacitance of external capacitor on VFBSMPS pin	10 µF
	ESR of external capacitor	20 mΩ
L	Inductance of external Inductor on VLXSMPS pin	2.2 µH
-	Serial DC resistor	150 mΩ
I_{SAT}	DC current at which the inductance drops 30% from its value without current.	1.7 A
I_{RMS}	Average current for a 40 °C rise: rated current for which the temperature of the inductor is raised 40°C by DC current	1.4 A

Table 25. SMPS step-down converter characteristics for external usage

Symbol	Conditions	Min	Typ	Max	Unit
V_{DDSMPS}(1)	V_{OUT} = 1.8 V	2.3	-	3.6	V
	V_{OUT} = 2.5 V	3	-	3.6	
V_{OUT}(2)	I_{OUT}=600 mA	2.25	2.5	2.75	V
		1.62	1.8	1.98	
I_{OUT}	internal and external usage	-	-	600	mA
	External usage only(3)	-	-	600	
RDS_{ON}	-	-	100	120	mΩ
IDD_{SMPS_Q}	Quiescent current	-	220	-	µA
T_{SMPS_START}	V_{OUT} = 1.8 V	-	270	405	µs
	V_{OUT} = 2.5 V	-	360	540	

1. *The switching frequency is 2.4 MHz±10%*
2. *Including line transient and load transient.*
3. *These characteristics are given for SMPSEXTHP bit is set in the PWR_CR3 register.*

The SMPS current consumption can be determined using the following formula based on the maximum LDO current consumption provided in Section 6.3.7 Supply current characteristics:

$$I_{DDSMPS} = I_{DDLDO} \times (V_{CORE} \div (V_{DD} \times efficency))$$

where

I_{DDLDO} is the current in LDO configuration given in the following tables

V_{CORE} is the digital core supply (V_{CAP})

Efficiency is defined in the following curves.

Figure 15. SMPS efficicency vs load current in Run, Sleep and Stop mode with SVOS3 MR mode, T_J = 30 °C

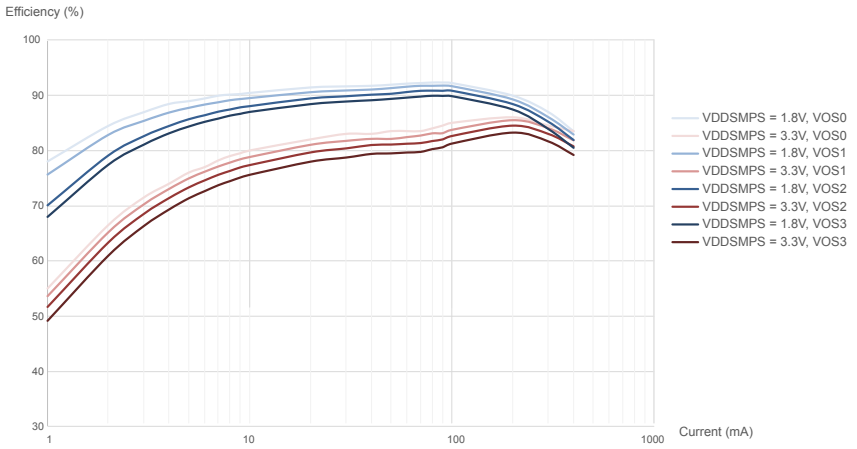


Figure 16. SMPS efficiency vs load current in Run, Sleep and Stop mode with SVOS3 MR mode, T_J = 130 °C

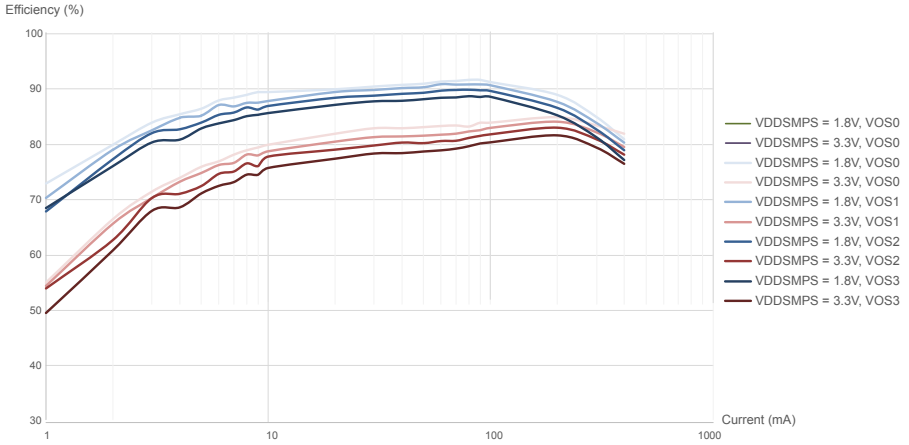


Figure 17. SMPS efficiency vs load current in Stop and DStop modes (SVOS3 LP mode, SVOS4, SVOS5), T_J = 30 °C

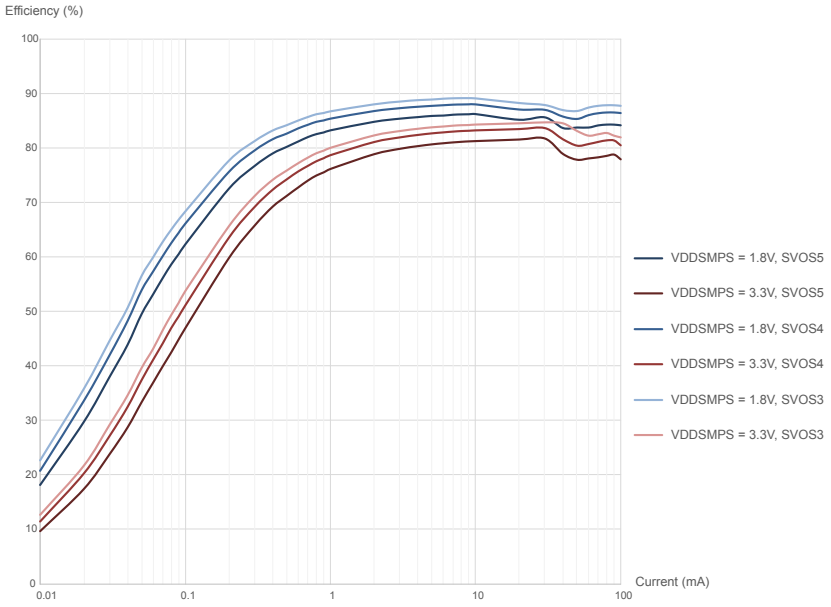


Figure 18. SMPS efficiency vs load current in Stop and DStop modes (SVOS3 LP mode, SVOS4, SVOS5), T_J = 130 °C

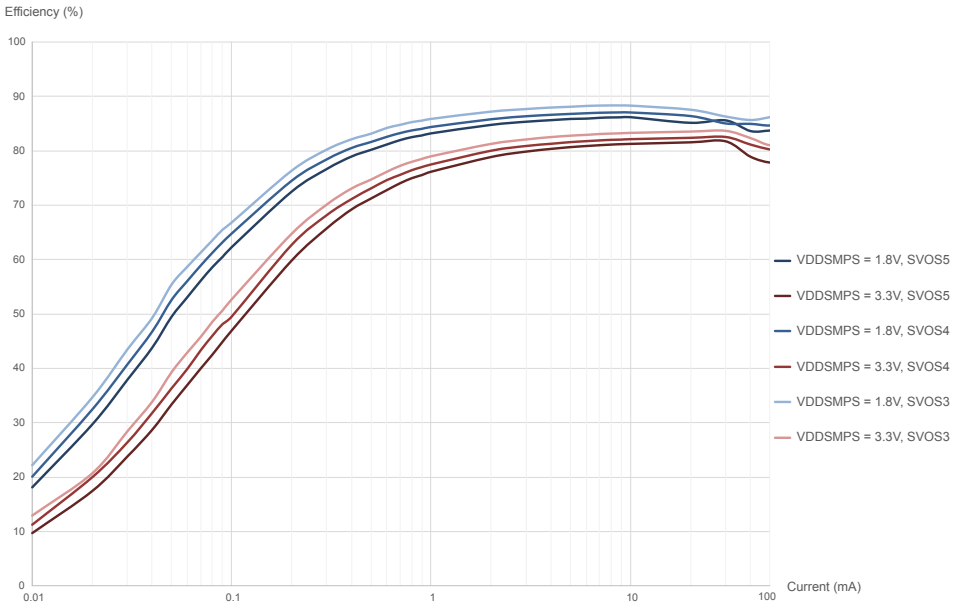


Figure 19. SMPS efficiency vs load current in Stop and DStop2 modes (SVOS3 LP mode, SVOS4, SVOS5), T_J = 30 °C

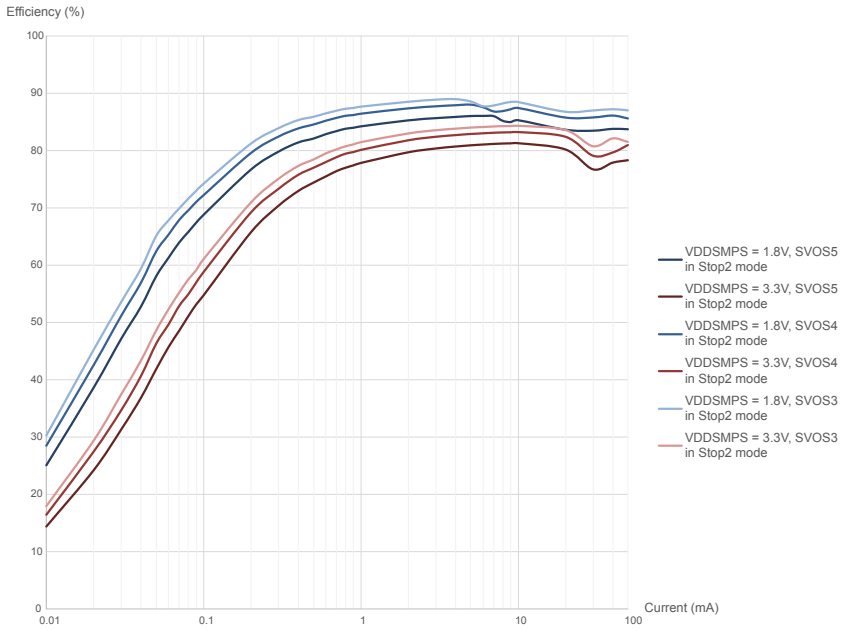
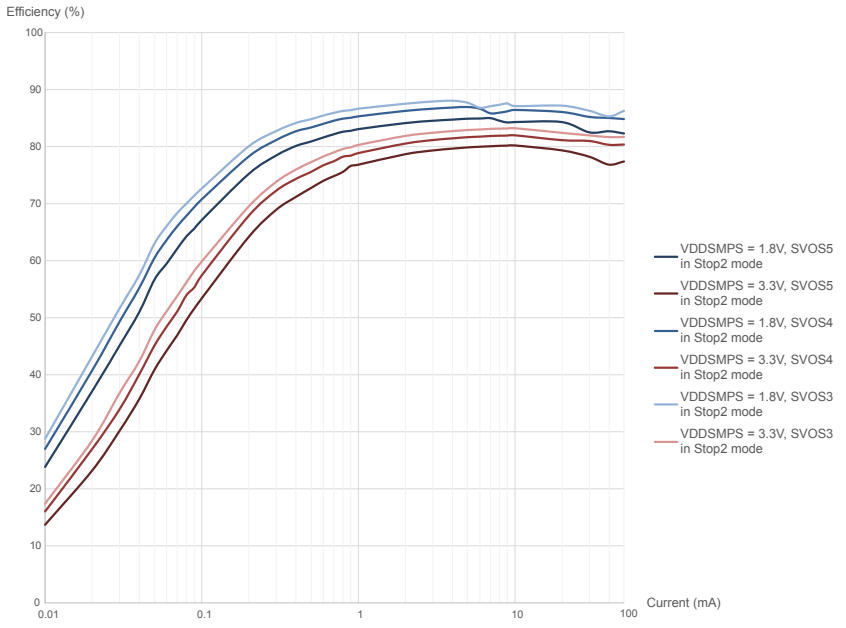


Figure 20. SMPS efficiency vs load current in Stop and DStop2 modes (SVOS3 LP mode, SVOS4, SVOS5), T_J = 130 °C



6.3.4 Operating conditions at power-up / power-down

Subject to general operating conditions for T_A.

Operating conditions at power-up / power-down (regulator ON)

Table 26. Operating conditions at power-up / power-down (regulator ON)

Symbol	Parameter	Min	Max	Unit
t_{VDD}	V_{DD} rise time rate	0	∞	µs/V
	V_{DD} fall time rate	10	∞	
t_{VDDA}	V_{DDA} rise time rate	0	∞	
	V_{DDA} fall time rate	10	∞	
t_{VDDUSB}	V_{DDUSB} rise time rate	0	∞	
	V_{DDUSB} fall time rate	10	∞	
VDDMMC	V_{DDMMC} rise time rate	0	∞	
	V_{DDMMC} fall time rate	10	∞	

6.3.5 Embedded reset and power control block characteristics

The parameters given in Table 27. Reset and power control block characteristics are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in Table 20. General operating conditions.

Table 27. Reset and power control block characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$t_{RSTTEMPO}$(1)	Reset temporization after POR released	-	-	377	550	µs
$V_{POR/PDR}$	Power-on/power-down reset threshold	Rising edge(1)	1.62	1.67	1.71	V
		Falling edge	1.58	1.62	1.68	
V_{BOR1}	Brown-out reset threshold 1	Rising edge	2.04	2.10	2.15	
		Falling edge	1.95	2.00	2.06	
V_{BOR2}	Brown-out reset threshold 2	Rising edge	2.34	2.41	2.47	
		Falling edge	2.25	2.31	2.37	
V_{BOR3}	Brown-out reset threshold 3	Rising edge	2.63	2.70	2.78	
		Falling edge	2.54	2.61	2.68	
V_{PVD0}	Programmable Voltage Detector threshold 0	Rising edge	1.90	1.96	2.01	
		Falling edge	1.81	1.86	1.91	
V_{PVD1}	Programmable Voltage Detector threshold 1	Rising edge	2.05	2.10	2.16	
		Falling edge	1.96	2.01	2.06	
V_{PVD2}	Programmable Voltage Detector threshold 2	Rising edge	2.19	2.26	2.32	
		Falling edge	2.10	2.15	2.21	
V_{PVD3}	Programmable Voltage Detector threshold 3	Rising edge	2.35	2.41	2.47	
		Falling edge	2.25	2.31	2.37	
V_{PVD4}	Programmable Voltage Detector threshold 4	Rising edge	2.49	2.56	2.62	
		Falling edge	2.39	2.45	2.51	
V_{PVD5}	Programmable Voltage Detector threshold 5	Rising edge	2.64	2.71	2.78	
		Falling edge	2.55	2.61	2.68	
V_{PVD6}	Programmable Voltage Detector threshold 6	Rising edge	2.78	2.86	2.94	
		Falling edge in Run mode	2.69	2.76	2.83	
$V_{POR/PDR}$	Hysteresis for power-on/power-down reset	Hysteresis in Run mode	-	43	-	mV

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{hyst_BOR_PVD}$	Hysteresis voltage of BOR	Hysteresis in Run mode	-	100	-	mV
$I_{DD_BOR_PVD}$(1)	BOR and PVD consumption from V_{DD}	-	-	-	0.630	µA
$I_{DD_POR_PDR}$	POR and PDR consumption from V_{DD}	-	0.8	-	1.2	
V_{AVM_0}	Analog voltage detector for VDDA threshold 0	Rising edge	1.66	1.71	1.76	V
		Falling edge	1.56	1.61	1.66	
V_{AVM_1}	Analog voltage detector for VDDA threshold 1	Rising edge	2.06	2.12	2.19	
		Falling edge	1.96	2.02	2.08	
V_{AVM_2}	Analog voltage detector for VDDA threshold 2	Rising edge	2.42	2.50	2.58	
		Falling edge	2.35	2.42	2.49	
V_{AVM_3}	Analog voltage detector for VDDA threshold 3	Rising edge	2.74	2.83	2.91	
		Falling edge	2.64	2.72	2.80	
V_{hyst_VDDA}	Hysteresis of VDDA voltage detector	-	-	100	-	mV
I_{DD_PVM}	PVM consumption from VDD(1)	-	-	-	0.25	µA
I_{DD_VDDA}	Voltage detector consumption on VDDA(1)	Resistor bridge	-	-	2.5	µA

1. *Guaranteed by design.*

6.3.6 Embedded reference voltage

The parameters given in Table 28 are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in Table 20. General operating conditions.

Table 28. Embedded reference voltage

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{REFINT}(1)	Internal reference voltages	−40 °C < T_J < 130 °C	1.180	1.216	1.255	V
$t_{S_vrefint}$(2)(3)	ADC sampling time when reading the internal reference voltage	-	4.3	-	-	µs
t_{S_vbat}(3)	V_{BAT} sampling time when reading the internal V_{BAT} reference voltage	-	9	-	-	
$t_{start_vrefint}$(3)	Start time of reference voltage buffer when ADC is enable	-	-	-	4.4	µs
Irefbuf(3)	Reference Buffer consumption for ADC	V_{DD} = 3.3 V	9	13.5	23	µA
ΔVREFINT(3)	Internal reference voltage spread over the temperature range	−40 °C < T_J < 130 °C	-	5	15	mV
T_{coeff}	Average temperature coefficient	Average temperature coefficient	-	20	70	ppm/°C
$V_{DDcoeff}$	Average Voltage coefficient	3.0 V < V_{DD} < 3.6 V	-	10	1370	ppm/V
V_{REFINT_DIV1}	1/4 reference voltage	-	-	25	-	% V_{REFINT}
V_{REFINT_DIV2}	1/2 reference voltage	-	-	50	-	
V_{REFINT_DIV3}	3/4 reference voltage	-	-	75	-	

1. *Guaranteed by design and tested in production at 3.3 V*
2. *The shortest sampling time for the application can be determined by multiple iterations.*
3. *Guaranteed by design.*

Table 29. Internal reference voltage calibration values

Symbol	Parameter	Memory address
V_{REFIN_CAL}	Raw data acquired at temperature of 30 °C, V_{DDA} = 3.3 V	08FFF810 - 08FFF812

Table 30. USB regulator characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{DD50USB}$	Supply voltage	-	4	5	5,5	V
$I_{DD50USB}$	Current consumption	-	-	13.5	-	µA
$V_{REGOUTV33V}$	Regulated output voltage	-	3	-	3.6	V
I_{OUT}	Output current load sinked by USB block	-	-	-	20	mA
T_{WKUP}	Wakeup time	-	-	120	170	µs

6.3.7 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, ambient temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code.

The current consumption is measured as described in Figure 13. Current consumption measurement scheme.

All the run-mode current consumption measurements given in this section are performed with a CoreMark code.

Typical and maximum current consumption

The MCU is placed under the following conditions:

- All I/O pins are in analog input mode.
- All peripherals are disabled except when explicitly mentioned.
- The flash memory access time is adjusted with the minimum wait states number, depending on the f_{ACLK} frequency (refer to the table “Number of wait states according to CPU clock ($f_{rcc_cpu_ck}$) frequency and V_{CORE} range” available in the reference manual).
- When the peripherals are enabled, the AHB clock frequency is the CPU frequency divided by 2 and the APB clock frequency is AHB clock frequency divided by 2.

The parameters given in the below tables are derived from tests performed under the supply voltage conditions summarized in Table 20. General operating conditions and unless otherwise specified at ambient temperature.

The maximum current consumptions provided in the following tables are given for LDO regulator ON. To obtain the maximum SMPS current consumption, the efficiency curves can be used with the maximum LDO current consumption as entry value (refer to Section 6.3.3 SMPS step-down converter).

Table 31. Inrush current and inrush electric charge characteristics for LDO and SMPS

1. The typical values are given for V_{DDLDO} = V_{DDSMPS} = 3.3 V and for typical decoupling capacitor values of C_{EXT} and C_{OUT}.
2. The product consumption on V_{DDCORE} is not taken into account in the inrush current and inrush electric charge.

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
I_{RUSH}	Inrush current on voltage regulator power-on (POR or wakeup from Standby)	on V_{DDLDO}(1)		-	55	96 (2)	mA
		on V_{DDSMPS}(3)	SMPS supplies the V_{DDCORE}	-	25	92(4)	
	Inrush current on voltage regulator power-on (POR)	on V_{DDSMPS}(3)	SMPS supplies internal LDO V_{OUT} = 1.8 V(5)	-	45	135(4)	
			SMPS supplies internal LDO V_{OUT} = 2.5 V(5)	-		100(4)	
			SMPS supplies external circuit V_{OUT} = 1.8 V(5)	-	25	70(4)	
			SMPS supplies external circuit V_{OUT} = 2.5 V(5)	-		50(4)	
	Inrush current on voltage regulator power-on (wakeup from Standby)	on V_{DDSMPS}(3)	SMPS supplies internal LDO V_{OUT} = 1.8 V	-	70	200(4)	
			SMPS supplies internal LDO V_{OUT} = 2.5 V	-	95	210(4)	
Q_{RUSH}	Inrush current on voltage regulator power-on (POR or wakeup from Standby)	on V_{DDLDO}(1)		-	4.4	5.3(2)	µC
		on V_{DDSMPS}(3)	SMPS supplies the V_{DDCORE}	-	2.9	7(2)	
	Inrush current on voltage regulator power-on (POR)	on V_{DDSMPS}(3)	SMPS supplies internal LDO V_{OUT} = 1.8 V(5)	-	4.0	7.5(2)	
			SMPS supplies internal LDO V_{OUT} = 2.5 V(5)	-		5.7(2)	
			SMPS supplies external circuit VOUT = 1.8 V(5)	-	2.9	5.2(2)	
			SMPS supplies external circuit V_{OUT} = 2.5 V(5)	-		4(2)	
	Inrush current on voltage regulator power-on (wakeup from Standby)	on V_{DDSMPS}(3)	SMPS supplies internal LDO V_{OUT} = 1.8 V	-	8.0	15(2)	
			SMPS supplies internal LDO V_{OUT} = 2.5 V	-	14.5	20.5(2)	

1. *The inrush current and inrush electric charge on V_{DDLDO} are not present in Bypass mode or when the SMPS supplies V_{DDCORE}.*
2. *The maximum value is given for the maximum decoupling capacitor C_{EXT}.*
3. *The inrush current and inrush electric charge on V_{DDSMPS} is not present if the external component (L or C_{OUT}) is not present, that is if the SMPS is not used.*
4. *The maximum value is given for the maximum decoupling capacitor C_{OUT} and the minimum V_{DDSMPS} voltage.*
5. *The inrush current and inrush electric charge due to the transition from 1.2 V to the final V_{OUT} value (1.8 V or 2.5 V) is not taken into account.*

Table 32. Typical and maximum current consumption in Run mode, code with data processing running from ITCM, regulator ON

Data are in DTCM for best computation performance. In this case, the cache has no influence on consumption.

Symbol	Parameter	Conditions		$f_{rcc_cpu_ck}$ (MHz)	Typ LDO	Typ SMPS	Max(1)(2)				unit
							T_J = 25 °C	T_J = 85 °C	T_J = 105 °C	T_J = 130 °C	
I_{DD}	Supply current in Run mode	All peripherals disabled	VOS0	280	69.5	34.0	77	106	128	173	mA
				225	56.5	27.5	64	92	114	159	
			VOS1	225	52.0	24.0	58	80	98	136	
				200	46.5	21.0	52	75	93	130	
				180	42	19.0	47	70	88	125	
				168	39	18.0	45	67	85	122	
				160	37.5	17.0	43	65	83	120	
			VOS2	160	34.0	14.5	38	56	70	101	
				144	30.5	13.0	35	52	67	97	
				88	19.0	8.5	23	41	55	85	
			VOS3	88	18.0	7.5	21	35	46	71	
				60	12.5	5.5	16	29	41	66	
				25	6.0	3.0	9	23	34	59	
		All peripherals enabled	VOS0	280	133.5	63.5	142	173	196	242	
				225	108.0	51.5	115	146	168	214	
			VOS1	225	99.0	45.0	105	129	147	185	
				160	71.5	32.5	77	100	118	156	
			VOS2	160	65.0	27.5	69	87	102	132	
				88	41.5	17.5	45	63	77	108	
			VOS3	88	38.0	15.0	41	55	66	91	

1. *Guaranteed by characterization results, unless otherwise specified.*
2. *The maximum values are given for LDO regulator ON. Refer to Section 6.3.3 SMPS step-down converter for the SMPS maximum current consumption.*

Table 33. Typical and maximum current consumption in Run mode, code with data processing running from flash memory, cache ON

Symbol	Parameter	Conditions		$f_{rcc_cpu_ck}$ (MHz)	Typ LDO(1)	Typ SMPS(1)	Max(1)(2)				unit
							T_J = 25 °C	T_J= 85 °C	T_J = 105 °C	T_J = 130 °C	
I_{DD}	Supply current in Run mode	All peripherals disabled	VOS0	280	69.0	33.5	77	106	128	173	mA
				225	56.0	27.0	64	92	114	158	
			VOS1	225	51.5	23.5	58	80	98	136	
				200	46.5	21.5	52	75	92	129	
				180	42.0	19.0	47	70	88	125	
				168	39.0	18.0	45	67	85	122	
				160	37.5	17.0	43	65	83	120	
			VOS2	160	34.0	14.5	38	56	70	101	
				144	30.5	13.0	35	53	67	97	
				88	19.0	8.5	23	41	55	85	
			VOS3	88	17.5	7.5	21	35	46	71	
				60	12.5	5.0	16	29	41	66	
				25	6.0	2.5	9	23	34	59	
		All peripherals enabled	VOS0	280	132.5	63.5	142	173	195	241	
				225	107.5	51.0	115	145	168	213	
			VOS1	225	99.0	44.5	105	129	147	185	
				160	71.5	32.5	77	100	118	155	
			VOS2	160	65.0	27.5	69	87	102	132	
				88	41.5	17.5	45	63	77	108	
			VOS3	88	38.0	15.0	41	55	66	91	

1. *Guaranteed by characterization results, unless otherwise specified.*
2. *The maximum values are given for LDO regulator ON. Refer to Section 6.3.3 SMPS step-down converter for the SMPS maximum current consumption.*

Table 34. Typical and maximum current consumption in Run mode, code with data processing running from flash memory, cache OFF

Symbol	Parameter	Conditions		$f_{rcc_cpu_ck}$ (MHz)	Typ LDO(1)	Typ SMPS(1)	Max(1)(2)				Unit
							T_J = 25 °C	T_J = 85 °C	T_J = 105 °C	T_J = 130 °C	
I_{DD}	Supply current in Run mode	All peripherals disabled	VOS0	280	56.0	28.0	63	91	113	157	mA
				225	47.0	23.5	54	82	103	148	
			VOS1	225	43.0	21.0	49	71	89	126	
				160	34.0	16.5	39	62	79	116	
			VOS2	160	29.5	13.5	34.	51	65	96	
				88	18.5	9.0	23	40	54	84	
			VOS3	88	16.5	7.5	19	33	44	69	
		All peripherals enabled	VOS0	280	119.5	58.0	127	157	180	225	
				225	98.5	48.0	105	135	157	203	
			VOS1	225	90.5	42.0	96	120	138	176	

Symbol	Parameter	Conditions		$f_{rcc_cpu_ck}$ (MHz)	Typ LDO(1)	Typ SMPS(1)	Max(1)(2)				Unit
							T_J = 25 °C	T_J = 85 °C	T_J = 105 °C	T_J = 130 °C	
I_{DD}	Supply current in Run mode	All peripherals enabled	VOS1	160	68.0	32.0	73	96	114	152	mA
			VOS2	160	60.5	26.5	64	82	97	127	
				88	41.0	18.0	45	62	77	107	
			VOS3	88	36.5	15.0	39	53	64	89	

1. *Guaranteed by characterization results, unless otherwise specified.*
2. *The maximum values are given for LDO regulator ON. Refer to Section 6.3.3 SMPS step-down converter for the SMPS maximum current consumption.*

Table 35. Typical consumption in Run mode and corresponding performance versus code position

Symbol	Parameter	Conditions		$f_{rcc_cpu_c\ k}$ (MHz)	Coremark	Typ LDO	Typ SMPS	Unit	LDO I_{DD}/ Coremark	SMPS I_{DD}/ Coremark	Unit
		Peripheral	Code								
I_{DD}	Supply current in Run mode	All peripherals disabled, cache ON	ITCM	280	1414	69.5	33.8	mA	49.2	23.9	µA/ Coremark
			FLASH	280	1414	69.0	33.4		48.8	23.6	
			AXI SRAM	280	1414	69.5	33.6		49.2	23.8	
			AHB SRAM	280	1414	70.0	33.7		49.5	23.8	
			SRD SRAM	280	1414	70.0	33.7		49.5	23.8	
		All peripherals disabled cache OFF	ITCM	280	1414	69.5	33.8		49.2	23.9	
			FLASH	280	668	56.0	28.0		83.8	41.9	
			AXI SRAM	280	668	62.5	30.2		93.6	45.2	
			AHB SRAM	280	295	59.5	28.8		201.7	97.6	
			SRD SRAM	280	295	59.0	28.5		200.0	96.6	

Table 36. Typical current consumption in Autonomous mode

Symbol	Parameter	Conditions(1)		f_{rcc_hclk4} (AHB4) (MHz)	Typ	Unit
I_{DD}	Supply current in Autonomous mode	Run, DStop mode	VOS3	64	2.98	mA
		Run, DStop2 mode	VOS3	64	2.64	

1. *System in Run mode, CPU domain is DStop or DStop2 mode with memories of the CPU domain shut-off enable or disable.*

Table 37. Typical current consumption in Sleep mode, regulator ON

Symbol	Parameter	Conditions		$f_{rcc_cpu_ck}$ (MHz)	Typ LDO	Typ SMPS	Max(1)(2)				Unit
							T_J = 25 °C	T_J = 85 °C	T_J = 105 °C	T_J= 130 °C	
$I_{DD(Sleep)}$	Supply current in Sleep mode	All peripherals disabled	VOS0	280	18.1	13.0	23	51	72	115	mA
				225	15.0	10.6	20	47	68	112	
			VOS1	225	13.7	9.3	18	40	57	93	
				160	10.3	6.8	14	36	53	90	
			VOS2	160	9.3	5.8	12	30	44	74	
				88	5.8	3.6	9	26	40	70	
			VOS3	88	5.2	3.0	8	21	32	57	

1. *Guaranteed by characterization results.*
2. *The maximum values are given for LDO regulator ON. Refer to Section 6.3.3 SMPS step-down converterfor the SMPS maximum current consumption.*

Table 38. Typical current consumption in System Stop mode

Symbol	Parameter	Conditions		Typ LDO	Typ SMPS	Max(1)(2)				Unit
						T_J = 25 °C	T_J = 85 °C	T_J = 105 °C	T_J= 130 °C	
$I_{DD(Stop)}$	Stop, DStop	Flash memory in low- power mode, memory shut-off disable	SVOS3 Main(3)	0.540	0.487	2.33	14.36	24.52	46.29	mA
			SVOS3 LP	0.495	0.193	2.27	14.21	24.28	45.94	
			SVOS4	0.370	0.137	1.59	10.58	18.52	35.90	
			SVOS5	0.245	0.090	0.98	7.18	13.10	26.61	
		Flash memory in normal mode, memory shut-off disable	SVOS3 Main(3)	0.560	0.504	2.39	14.62	24.93	47.01	
			SVOS3 LP	0.515	0.209	2.33	14.47	24.69	46.65	
			SVOS4	0.390	0.153	1.65	10.84	18.93	36.62	
			SVOS5	0.245	0.090	1.04	7.43	13.51	27.32	
		Flash memory in low- power mode, memory shut-off enable	SVOS3 Main(3)	0.530	0.481	2.31	14.23	24.27	45.71	
			SVOS3 LP	0.480	0.186	2.25	14.09	24.04	45.36	
			SVOS4	0.360	0.134	1.57	10.49	18.32	35.41	
			SVOS5	0.230	0.085	0.96	6.95	12.59	25.26	
		Flash memory in normal mode, memory shut-off enable	SVOS3 Main(3)	0.550	0.498	2.37	14.50	24.68	46.43	
			SVOS3 LP	0.500	0.204	2.31	14.35	24.45	46.07	
			SVOS4	0.380	0.151	1.63	10.75	18.73	36.13	
			SVOS5	0.230	0.085	1.02	7.21	13.00	25.97	
	Stop, DStop2	Flash memory in low- power mode, memory shut-off disable	SVOS3 Main(3)	0.161	0.343	0.32	1.67	2.86	5.58	
			SVOS3 LP	0.115	0.046	0.28	1.62	2.80	5.50	
			SVOS4	0.095	0.037	0.20	1.23	2.19	4.43	
			SVOS5	0.090	0.032	0.14	0.93	1.75	3.80	

Symbol	Parameter	Conditions		Typ LDO	Typ SMPS	Max(1)(2)				Unit
						T_J = 25 °C	T_J = 85 °C	T_J = 105 °C	T_J= 130 °C	
$I_{DD(Stop)}$	Stop, DStop2	Flash memory in low -power mode, memory shut-off enable	SVOS3 Main(3)	0.146	0.337	0.30	1.55	2.63	5.04	mA
			SVOS3 LP	0.100	0.040	0.26	1.51	2.58	4.96	
			SVOS4	0.085	0.033	0.19	1.15	2.01	3.98	
			SVOS5	0.075	0.028	0.12	0.80	1.46	3.02	

1. *Guaranteed by characterization results.*
2. *The maximum values are given for LDO regulator ON. Refer to Section 6.3.3 SMPS step-down converterfor the SMPS maximum current consumption.*
3. *The SMPS is unnecessarily ON in System Stop mode, leading to an additional consumption. It is recommended to use SVOS3 LP mode to optimize power consumption.*

Table 39. Typical current consumption RAM shutoff in Stop mode

Symbol	Parameter	Conditions	Typ LDO			Unit
			SVOS3 LP	SVOS4	SVOS5	
Δ_{IDD}(Stop)	Stop, Dstop or Dstop2	AXISRAM1 shutoff power consumption (power consumption reduction when AXISRAM1 shutoff is enabled)	3.00	1.80	3.00	µA
		AXISRAM2 shutoff power consumption (power consumption reduction when AXISRAM2 shutoff is enabled)	4.40	2.70	4.40	
		AXISRAM13 shutoff power consumption (power consumption reduction when AXISRAM3 shutoff is enabled)	4.40	2.70	4.40	
		AHBSRAM1 shutoff power consumption (power consumption reduction when AHBSRAM1 shutoff is enabled)	0.90	0.50	0.70	
		AHBSRAM2 shutoff power consumption (power consumption reduction when AHBSRAM2 shutoff is enabled)	0.90	0.50	0.70	
		ITCM and ETM shutoff power consumption (power consumption reduction when ITCM and ETM shutoff is enabled)	1.00	0.60	0.90	
		GFXMMU and JPEG shutoff power consumption (power consumption reduction when GFXMMU and JPEG shutoff is enabled)	0.20	0.10	0.10	
		High-speed interface USB and FDCAN shutoff power consumption (power consumption reduction when High-speed interface USB and FDCAN shutoff is enabled)	0.20	0.10	0.10	
		SRDSRAM shutoff power consumption (power consumption reduction when SRDSRAM shutoff is enabled)	0.30	0.30	0.40	

Table 40. Typical and maximum current consumption in Standby mode

Symbol	Parameter	Conditions		Typ				Max (3.6V)(1)				Unit
		Backup SRAM	RTC & LSE(2)	1.62 V	2.4 V(3)	3 V(3)	3.3 V(3)	T_J = 25 °C	T_J = 85 °C	T_J = 105 °C	T_J = 130 °C	
I_{DD} (Standby)	Supply current in Standby mode, **IWDG OFF**	OFF	OFF	1.97	2.76	3.02	3.30	4.0	11.0	22.0	57.0	µA
		ON	OFF	2.78	3.69	4.02	4.40	5.4	13.0	25.0	64.0	
		OFF	ON	2.46	3.37	3.73	4.07	5.0	12.2	23.3	59.0	
		ON	ON	3.27	4.30	4.73	5.17	6.4	14.2	26.3	66.0	

1. *Guaranteed by characterization results.*
2. *The LSE clock is in low-drive mode.*

3. *These values are given for PDR ON. When the PDR is OFF (internal reset OFF), the typical current consumption is reduced (refer to Section 6.3.5 Embedded reset and power control block characteristics).*

Table 41. Typical and maximum current consumption in V_{BAT} mode

Symbol	Parameter	Conditions		Typ				Max (3.6V)(1)				Unit
		Backup SRAM	RTC & LSE(2)	1.2 V	2 V	3 V	3.3 V	T_J = 25 °C	T_J = 85 °C	T_J = 105 °C	T_J= 130 °C	
I_{DD} (VBAT)	Supply current in V_{BAT} mode	OFF	OFF	0.01	0.02	0.03	0.07	0.2	1.9	4.6	14	µA
		ON	OFF	0.85	0.93	1.05	1.14	1.5	3.6	7.5	20.0	
		OFF	ON	0.50	0.63	0.74	0.84	1.2	3.1	5.9	16	
		ON	ON	1.34	1.54	1.76	1.91	2.5	4.8	8.8	22.0	

1. *Guaranteed by characterization results.*
2. *The LSE clock is in low-drive mode.*

I/O system current consumption

I/O static current consumption

All the I/Os used as inputs with pull-up generate a current consumption when the pin is externally held low. The value of this current consumption can be simply computed by using the pull-up/pull-down resistors values given in Table 63. I/O static characteristics.

For the output pins, any external pull-down or external load must also be considered to estimate the current consumption.

An additional I/O current consumption is due to I/Os configured as inputs if an intermediate voltage level is externally applied. This current consumption is caused by the input Schmitt trigger circuits used to discriminate the input value. Unless this specific configuration is required by the application, this supply current consumption can be avoided by configuring these I/Os in analog mode. This is notably the case of ADC input pins which should be configured as analog inputs.

Caution: Any floating input pin can also settle to an intermediate voltage level or switch inadvertently, as a result of external electromagnetic noise. To avoid a current consumption related to floating pins, they must either be configured in analog mode, or forced internally to a definite digital value. This can be done either by using pull-up/down resistors or by configuring the pins in output mode.

I/O dynamic current consumption

In addition to the internal peripheral current consumption (see Table 42. Peripheral current consumption in Run mode), the I/Os used by an application also contribute to the current consumption. When an I/O pin switches, it uses the current from the MCU supply voltage to supply the I/O pin circuitry and to charge/discharge the capacitive load (internal or external) connected to the pin:

$$I_{SW}=V_{DDx} \times f_{sw} \times C_L$$

where

I_{SW} is the current sunk by a switching I/O to charge/discharge the capacitive load

V_{DDx} is the MCU supply voltage

f_{SW} is the I/O switching frequency

C_L is the total capacitance seen by the I/O pin: $C = C_{INT} + C_{EXT}$

The test pin is configured in push-pull output mode and is toggled by software at a fixed frequency.

On-chip peripheral current consumption

The MCU is placed under the following conditions:

- At startup, all I/O pins are in analog input configuration.
- All peripherals are disabled unless otherwise mentioned.
- The I/O compensation cell is enabled.
- $f_{rcc_cpu_ck}$ is the CPU clock. $f_{PCLK} = f_{rcc_cpu_ck}/4$, and $f_{HCLK} = f_{rcc_cpu_ck}/2$.

The given value is calculated by measuring the difference of current consumption

- with all peripherals clocked off
- with only one peripheral clocked on
- $f_{rcc_cpu_ck}$ = 280 MHz (Scale 0), $f_{rcc_cpu_ck}$ = 225 MHz (Scale 1), $f_{rcc_cpu_ck}$ = 160 MHz (Scale 2), $f_{rcc_cpu_ck}$ = 88 MHz (Scale 3)
- The ambient operating temperature is 25 °C and V_{DD}=3.3 V.

Table 42. Peripheral current consumption in Run mode

Peripheral		I_{DD}(Typ)				Unit
		VOS0	VOS1	VOS2	VOS3	
AHB3	MDMA	7.10	6.40	5.90	5.40	µA/MHz
	DMA2D	3.00	2.80	2.50	2.30	
	JPGDEC	4.70	4.40	4.00	3.60	
	FLITF	20.00	19.00	17.00	15.00	
	FMC registers	1.30	1.30	1.20	1.10	
	FMC kernel	10.00	9.30	8.40	7.70	
	OSPI1 registers	0.50	0.60	0.50	0.50	
	OSPI1 kernel	2.30	2.20	2.00	1.80	
	SDMMC1 registers	8.90	8.30	7.60	6.90	
	SDMMC1 kernel	2.20	2.00	1.80	1.60	
	OSPI2 registers	0.70	0.70	0.70	0.60	
	OSPI2 kernel	2.00	1.80	1.60	1.50	
	IOMNGR	0.30	0.30	0.30	0.30	
	OTFDEC1	1.20	1.20	1.10	1.10	
	OTFDEC2	1.40	1.30	1.20	1.20	
	GFXMMU	2.80	2.70	2.40	2.30	
	AXISRAM2	5.30	5.00	4.60	4.20	
	AXISRAM3	5.40	5.10	4.60	4.30	
	DTCM1	1.10	1.10	1.00	1.00	
	DTCM2	0.70	0.80	0.70	0.70	
	ITCM	1.10	1.10	1.00	1.00	
	AXISRAM1	5.30	5.00	4.60	4.20	
	Bridge	0.10	0.10	0.10	0.10	
AHB1	DMA1	0.90	0.90	0.80	0.70	
	DMA2	0.90	0.80	0.80	0.70	
	CRC	0.60	0.60	0.50	0.50	
	ADC12 registers	5.40	4.90	4.50	4.10	
	ADC12 kernel	1.10	1.00	0.90	0.80	
	USB1OTG registers	24.00	22.00	20.00	18.00	
	USB1OTG kernel	9.50	9.30	9.10	8.80	
	USB1ULPI	0.10	0.10	0.10	0.10	
	Bridge	0.10	0.10	0.10	0.10	
AHB2	CRYPT	1.50	1.40	1.30	1.20	
	HASH	1.80	1.60	1.50	1.30	

Peripheral		I_{DD}(Typ)				Unit
		VOS0	VOS1	VOS2	VOS3	
AHB2	DCMI	5.00	4.60	4.20	3.90	µA/MHz
	HSEM	0.10	0.10	0.10	0.10	
	RNG registers	1.50	1.40	1.20	1.10	
	RNG kernel	10.00	9.70	9.50	9.20	
	SDMMC2 registers	6.80	6.30	5.70	5.20	
	SDMMC2 kernel	2.30	2.10	1.90	1.70	
	BDMA1	1.70	1.60	1.50	1.30	
	AHBSRAM1	0.70	0.70	0.60	0.60	
	AHBSRAM2	0.70	0.60	0.60	0.50	
	Bridge	9.10	8.40	7.70	7.00	
AHB4	GPIOA	2.00	1.80	1.70	1.50	
	GPIOB	1.80	1.70	1.50	1.40	
	GPIOC	2.00	1.80	1.70	1.50	
	GPIOD	2.00	1.80	1.70	1.50	
	GPIOE	1.90	1.80	1.60	1.50	
	GPIOF	1.90	1.80	1.60	1.50	
	GPIOG	2.00	1.80	1.70	1.50	
	GPIOH	1.90	1.80	1.60	1.50	
	GPIOI	1.90	1.80	1.60	1.50	
	GPIOJ	1.90	1.80	1.60	1.50	
	GPIOK	2.00	1.80	1.70	1.50	
	BDMA2	4.20	3.90	3.50	3.20	
	SRDSRAM	0.60	0.50	0.50	0.50	
	BKPRAM	0.80	0.70	0.70	0.60	
	IWDG	0.07	0.07	0.07	0.07	
	Bridge	0.10	0.10	0.10	0.10	
APB3	LTDC	12.00	11.00	9.80	8.90	
	WWDG1	1.10	1.00	0.90	0.90	
	Bridge	0.10	0.10	0.10	0.10	
APB1	TIM2	7.50	6.90	6.30	6.20	
	TIM3	6.30	5.90	5.40	4.90	
	TIM4	5.80	5.40	4.90	4.50	
	TIM5	7.20	6.70	6.10	5.60	
	TIM6	1.60	1.50	1.30	1.20	
	TIM7	1.60	1.40	1.30	1.20	
	TIM12	3.60	3.30	3.00	2.80	
	TIM13	2.80	2.60	2.40	2.10	
	TIM14	2.50	2.30	2.10	1.90	
	LPTIM1 registers	0.80	0.80	0.70	0.60	
	LPTIM1 kernel	2.20	2.00	1.80	1.70	

Peripheral		I_{DD}(Typ)				Unit
		VOS0	VOS1	VOS2	VOS3	
APB1	SPI2 registers	2.20	2.00	1.80	1.70	µA/MHz
	SPI2 kernel	0.90	0.80	0.80	0.70	
	SPI3 registers	2.70	2.40	2.30	2.00	
	SPI3 kernel	0.90	0.80	0.70	0.70	
	SPDIFRX1 registers	0.60	0.50	0.50	0.40	
	SPDIFRX1 kernel	2.90	2.70	2.50	2.20	
	USART2 registers	2.00	1.80	1.70	1.50	
	USART2 kernel	4.60	4.30	3.90	3.60	
	USART3 registers	2.00	1.80	1.70	1.50	
	USART3 kernel	4.50	4.20	3.80	3.40	
	UART4 registers	1.70	1.60	1.50	1.30	
	UART4 kernel	3.70	3.40	3.10	2.80	
	UART5 registers	1.80	1.70	1.50	1.40	
	UART5 kernel	3.80	3.50	3.20	2.90	
	I2C1 registers	0.90	0.80	0.80	0.70	
	I2C1 kernel	2.10	2.00	1.80	1.70	
	I2C2 registers	0.90	0.80	0.70	0.70	
	I2C2 kernel	2.10	1.90	1.80	1.60	
	I2C3 registers	0.90	0.80	0.70	0.70	
	I2C3 kernel	2.20	2.00	1.80	1.70	
	HDMICEC registers	0.50	0.50	0.40	0.40	
	HDMICEC kernel	0.10	0.10	0.10	0.10	
	DAC1	1.40	1.30	1.20	1.10	
	UART7 registers	1.80	1.70	1.50	1.40	
	UART7 kernel	3.80	3.50	3.20	2.90	
	UART8 registers	2.10	2.00	1.80	1.70	
	UART8 kernel	3.80	3.50	3.20	2.90	
	Bridge	0.30	0.30	0.20	0.10	
	CRS	0.50	0.40	0.40	0.40	
	SWP registers	2.30	2.10	2.00	1.80	
	SWP kernel	0.10	0.10	0.10	0.10	
	OPAMP	4.20	3.80	3.50	3.20	
	MDIO	3.10	2.90	2.60	2.40	
	FDCAN registers	17.00	16.00	15.00	14.00	
	FDCAN kernel	5.60	4.80	3.50	1.10	
	Bridge	0.10	0.10	0.10	0.10	
APB2	TIM1	9.80	9.10	8.30	7.60	
	TIM8	9.50	8.80	8.00	7.30	
	USART1 registers	0.10	0.10	0.10	0.10	
	USART1 kernel	0.10	0.10	0.10	0.10	

Peripheral		I_{DD}(Typ)				Unit
		VOS0	VOS1	VOS2	VOS3	
APB2	USART6 registers	3.80	4.00	4.50	6.30	µA/MHz
	USART6 kernel	0.10	0.10	0.10	0.10	
	USART10 registers	4.00	4.10	4.60	6.40	
	USART10 kernel	0.10	0.10	0.10	0.10	
	UART9 registers	3.50	3.60	4.00	5.50	
	UART9 kernel	0.10	0.10	0.10	0.10	
	SPI1 registers	2.10	1.90	1.80	1.60	
	SPI1 kernel	0.90	0.80	0.70	0.70	
	SPI4 registers	2.10	1.90	1.70	1.50	
	SPI4 kernel	0.50	0.50	0.40	0.40	
	TIM15	5.30	4.90	4.40	4.00	
	TIM16	4.20	3.90	3.50	3.20	
	TIM17	4.30	4.00	3.60	3.30	
	SPI5 registers	2.00	1.90	1.70	1.50	
	SPI5 kernel	0.50	0.50	0.40	0.40	
	SAI1 registers	1.80	1.60	1.50	1.30	
	SAI1 kernel	1.40	1.30	1.20	1.00	
	SAI2 registers	2.30	2.10	1.90	1.70	
	SAI2 kernel	1.20	1.10	1.00	0.90	
	DFSDM1 registers	10.00	9.60	8.80	8.00	
	DFSDM1 kernel	0.10	0.10	0.10	0.10	
	Bridge	0.50	0.40	0.40	0.30	
APB4	SYSCFG	0.40	0.30	0.30	0.30	
	LPUART1 registers	1.10	1.00	0.90	0.80	
	LPUART1 kernel	2.30	2.10	1.90	1.70	
	SPI6 registers	1.70	1.50	1.40	1.30	
	SPI6 kernel	0.60	0.50	0.50	0.40	
	I2C4 registers	0.80	0.70	0.60	0.60	
	I2C4 kernel	1.90	1.70	1.60	1.40	
	LPTIM2 registers	0.60	0.60	0.50	0.50	
	LPTIM2 kernel	1.90	1.70	1.60	1.40	
	LPTIM3 registers	0.60	0.50	0.50	0.40	
	LPTIM3 kernel	1.50	1.40	1.30	1.20	
	DAC2	0.80	0.70	0.60	0.50	
	COMP12	0.40	0.30	0.30	0.30	
	VREF	0.30	0.30	0.20	0.20	
	RTCAPB	1.90	1.70	1.60	1.40	
	TMPSENS	2.30	2.10	2.00	1.80	
	DFSDM2 registers	1.70	1.50	1.40	1.30	
	DFSDM2 kernel	0.10	0.10	0.10	0.10	

Peripheral		I_{DD}(Typ)				Unit
		VOS0	VOS1	VOS2	VOS3	
APB4	Bridge	0.10	0.10	0.10	0.10	µA/MHz

Table 43. Peripheral current consumption in Stop, Standby and V_{BAT} mode

Symbol	Parameter	Conditions	Typ	Max (3.6 V)				Unit
			3.3 V	T_J = 25 °C	T_J= 85 °C	T_J = 105 °C	T_J = 130 °C	
I_{DD}	RTC+LSE low drive	-	0.77	1.0	1.2	1.3	2.0	µA
	RTC+LSE medium- low drive	-	0.87	1.1	1.3	1.4	2.1	
	RTC+LSE medium- high drive	-	1.03	1.3	1.5	1.6	2.3	
	RTC+LSE High drive	-	1.38	1.6	1.8	1.9	2.6	
	Backup SRAM	-	1.10	1.4	2.0	3.2	7.0	

6.3.8 Wakeup time from low-power modes

The wakeup times given in Table 44. Low-power mode wakeup timings are measured starting from the wakeup event trigger up to the first instruction executed by the CPU:

- For Stop or Sleep modes: the wakeup event is WFE.
- WKUP (PC1) pin is used to wakeup from Standby, Stop and Sleep modes.

All timings are derived from tests performed under ambient temperature and V_{DD}=3.3 V.

Table 44. Low-power mode wakeup timings

Symbol	Parameter	Conditions	Typ(1)	Max(1)(2)	Unit
$t_{WUSLEEP}$(3)	Wakeup from Sleep	-	5.00	5.00	CPU clock cycles
$t_{WUDSTOP}$(3)	Wakeup from DStop	SVOS3 Main, HSI, flash memory in normal mode	4.2	6	µs
		SVOS3 Main, HSI, flash memory in low-power mode	8.3	11	
		SVOS3 LP, HSI, flash memory in normal mode	5.0	7	
		SVOS3 LP, HSI, flash memory in low-power mode	9.0	12	
		SVOS4, HSI, flash memory in normal mode	15.7	19	
		SVOS4, HSI, flash memory in low-power mode	19.7	25	
		SVOS5, HSI, flash memory in normal mode	35.0	43	
		SVOS5, HSI, flash memory in low-power mode	35.0	43	
		SVOS3 Main, CSI, flash memory in normal mode	42.5	52	
		SVOS3 Main, CSI, flash memory in low power mode	48.0	58	
		SVOS3 LP, CSI, flash memory in normal mode	43.3	53	
		SVOS3 LP, CSI, flash memory in low power mode	48.8	59	
		SVOS4, CSI, flash memory in normal mode	54.0	65	
		SVOS4, CSI, flash memory in low-power mode	59.5	72	
		SVOS5, CSI, flash memory in normal mode	74.8	90	
		SVOS5, CSI, flash memory in low-power mode	74.8	90	
$t_{WUDSTOP2}$(3)	Wakeup from DStop2, clock kept running	SVOS3 LP, HSI, flash memory in low-power mode	9.7	13	
		SVOS4, HSI, flash memory in low-power mode	20.4	26	
		SVOS5, HSI, flash memory in low-power mode	35.7	44	

Symbol	Parameter	Conditions	Typ(1)	Max(1)(2)	Unit
$t_{WUDSTOP2}$(3)	Wakeup from DStop2, clock kept running	SVOS3 LP, CSI, flash memory in low-power mode	51.3	62	µs
		SVOS4, CSI, flash memory in low-power mode	62.0	75	
		SVOS5, CSI, flash memory in low-power mode	77.3	93	
$t_{WUSTDBY}$(3)	Wakeup from Standby mode	-	257	330	

1. *Guaranteed by characterization results.*
2. *Measures done at −40 °C in the worst conditions.*
3. *The wakeup times are measured from the wakeup event to the point in which the application code reads the first instruction.*

6.3.9 External clock source characteristics

High-speed external user clock generated from an external source

In bypass mode, the HSE oscillator is switched off and the input pin is a standard I/O.

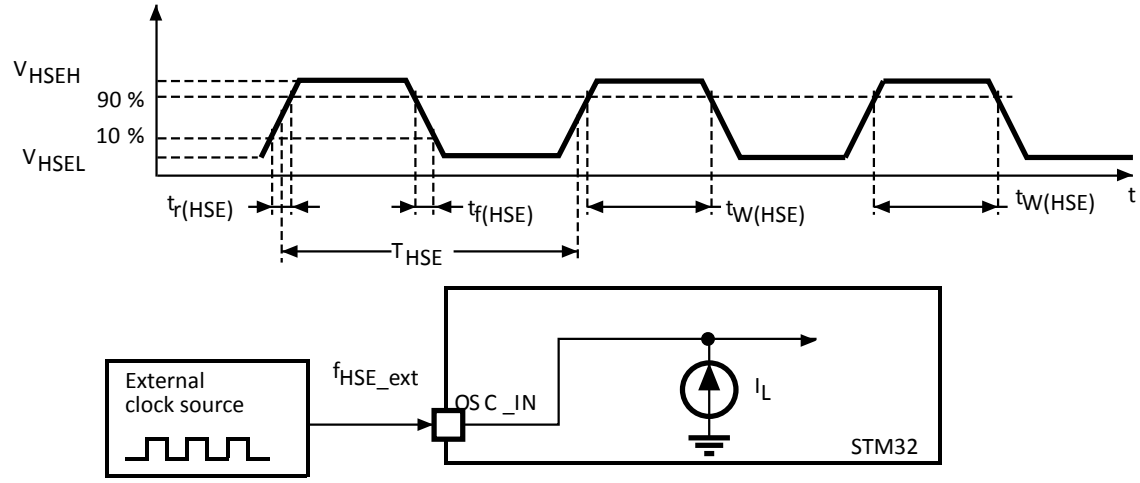
The external clock signal has to respect Table 45. High-speed external user clock characteristics in addition to Table 63. I/O static characteristics. The external clock can be low-swing (analog) or digital. In case of a low-swing analog input clock, the clock squarer must be activated (refer to RM0455).

Table 45. High-speed external user clock characteristics

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)	Unit
f_{HSE_ext}	User external clock source frequency	External digital/analog clock	4	25	50	MHz
V_{HSEH}	Digital OSC_IN input high-level voltage	External digital clock	0.7 V_{DD}	-	V_{DD}	V
V_{HSEL}	Digital OSC_IN input low-level voltage		V_{SS}	-	0.3 V_{DD}	
$t_{W(HSEH)}/t_{W(HSEL)}$(2)	Digital OSC_IN input high or low time	External digital clock	7	-	-	ns
V_{lswHSE} (V_{HSEH}−V_{HSEL})(3)	Analog low-swing OSC_IN peak-to-peak amplitude	External analog low-swing clock	0.2	-	2/3 V_{DD}	V
$DuCy_{HSE}$	Analog low-swing OSC_IN duty cycle		45	50	55	%
$t_{r(HSE)}/t_{f(HSE)}$	Analog low-swing OSC_IN rise and fall times	External analog low-swing clock, 10% to 90%	0.05 / f_{HSE_ext}	-	0.3 / f_{HSE_ext}	ns

1. *Guaranteed by design.*
2. *The rise and fall times for a digital input signal are not specified. However the V_{HSEH} and V_{HSEL} conditions must be fulfilled.*
3. *The DC component of the signal must ensure that the signal peaks are located between V_{DD} and V_{SS}.*

Figure 21. **High-speed external clock source AC timing diagram**



Low-speed external user clock generated from an external source

In bypass mode, the LSE oscillator is switched off and the input pin is a standard I/O. The external clock signal has to respect Table 46. Low-speed external user clock characteristics in addition to Table 63. I/O static characteristics. The external clock can be low-swing (analog) or digital. In case of a low-swing analog input clock, the clock squarer must be activated (refer to RM0455).

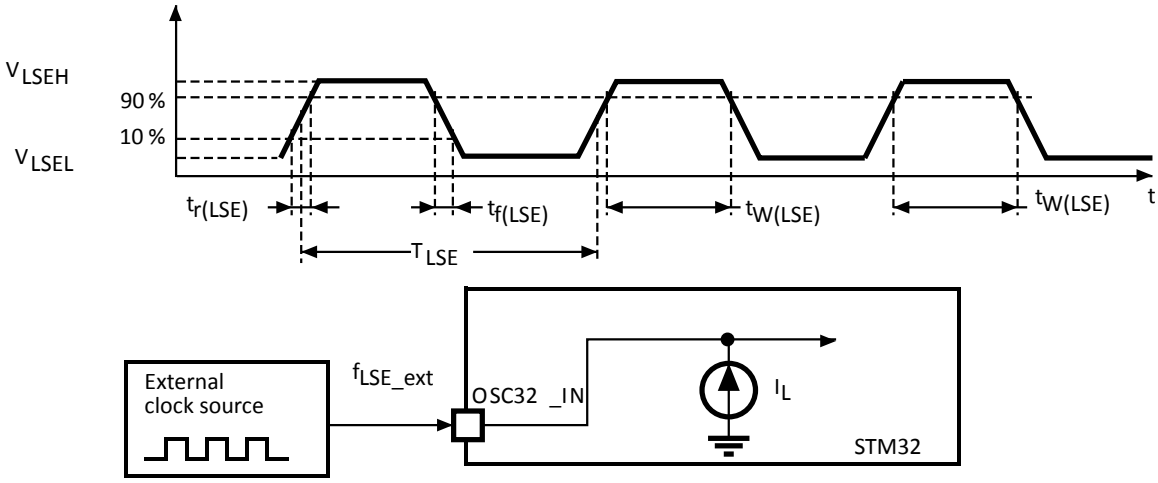
Table 46. **Low-speed external user clock characteristics**

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)	Unit
f_{LSE_ext}	User external clock source frequency	External digital/analog clock	-	32.768	1000	kHz
V_{LSEH}	Digital OSC32_IN input high-level voltage	External digital clock	0.7 V_{DD}	-	V_{DD}	V
V_{LSEL}	OSC32_IN input low-level voltage		V_{SS}	-	0.3 V_{DD}	
$t_{w(LSEH)}$/ $t_{w(LSEL)}$	OSC32_IN high or low time	External digital clock	250	-	-	ns
V_{lsw_H}	Analog low-swing OSC_IN high-level voltage	External analog low-swing clock	0.6	-	1.225	V
V_{lsw_L}	Analog low-swing OSC_IN low-level voltage		0.35	-	0.8	
V_{lswLSE} (V_{LSEH}–V_{LSEL})	Analog low-swing OSC_IN peak-to-peak amplitude		0.2	-	0.875	
$DuCy_{LSE}$	Analog low-swing OSC_IN duty cycle		45	50	55	%
$t_{r(LSE)}$/$t_{f(LSE)}$	Analog low-swing OSC_IN rise and fall times	External analog low-swing clock, 10% to 90%	-	100	200	ns

1. *Guaranteed by design.*

Note: For information on selecting the crystal, refer to the application note AN2867 “Oscillator design guide for ST microcontrollers” available from the ST website www.st.com.

Figure 22. **Low-speed external clock source AC timing diagram**



High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with a 4 to 50 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in Table 47. 4-50 MHz HSE oscillator characteristics. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Table 47. **4-50 MHz HSE oscillator characteristics**

Symbol	Parameter	Operating conditions[1]	Min[2]	Typ[2]	Max[2]	Unit
F	Oscillator frequency	-	4	-	50	MHz
R_F	Feedback resistor	-	-	200	-	kΩ
$I_{DD(HSE)}$	HSE current consumption	During startup[3]	-	-	4	mA
		V_{DD}=3 V, Rm=30 Ω C_L=10 pF at 4 MHz	-	0.35	-	
		V_{DD}=3 V, Rm=30 Ω C_L=10 pF at 8 MHz	-	0.40	-	
		V_{DD}=3 V, Rm=30 Ω C_L=10 pF at 16 MHz	-	0.45	-	
		V_{DD}=3 V, Rm=30 Ω C_L=10 pF at 32 MHz	-	0.65	-	
		V_{DD}=3 V, Rm=30 Ω C_L=10 pF at 48 MHz	-	0.95	-	
$Gm_{critmax}$	Maximum critical crystal gm	Startup	-	-	1.5	mA/V
t_{SU}[4]	Start-up time	V_{DD} is stabilized	-	2	-	ms

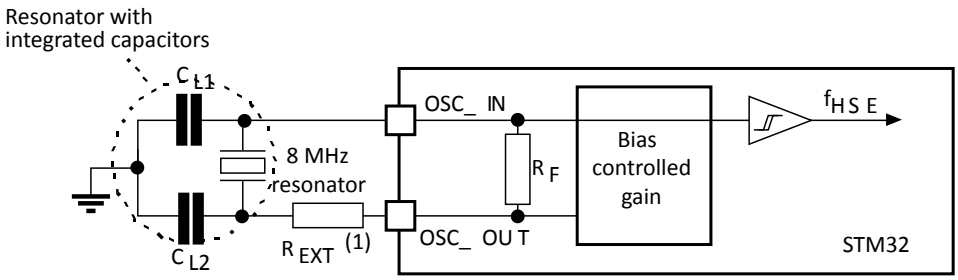
1. *Resonator characteristics given by the crystal/ceramic resonator manufacturer.*
2. *Guaranteed by design.*
3. *This consumption level occurs during the first 2/3 of the $t_{SU(HSE)}$ startup time.*

4. *$t_{SU(HSE)}$ is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer.*

For C_{L1} and C_{L2}, it is recommended to use high-quality external ceramic capacitors in the 5 pF to 25 pF range (typical), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see Figure 23. Typical application with an 8 MHz crystal). C_{L1} and C_{L2} are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of C_{L1} and C_{L2}. The PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing C_{L1} and C_{L2}.

Note: For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

Figure 23. Typical application with an 8 MHz crystal



1. R_{EXT} value depends on the crystal characteristics.

Low-speed external clock generated from a crystal/ceramic resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in Table 48. Low-speed external user clock characteristics. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Table 48. Low-speed external user clock characteristics

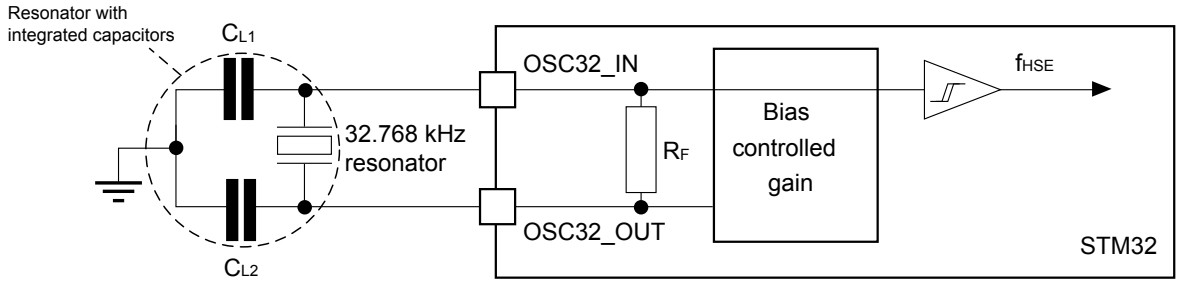
Symbol	Parameter	Operating conditions[1]	Min[2]	Typ[2]	Max[2]	Unit
F	Oscillator frequency	-	-	32.768	-	kHz
I_{DD}	LSE current consumption	LSEDRV[1:0] = 00, Low drive capability	-	290	-	nA
		LSEDRV[1:0] = 01, Medium Low drive capability	-	390	-	
		LSEDRV[1:0] = 10, Medium high drive capability	-	550	-	
		LSEDRV[1:0] = 11, High drive capability	-	900	-	
$Gm_{critmax}$	Maximum critical crystal gm	LSEDRV[1:0] = 00, Low drive capability	-	-	0.5	µA/V
		LSEDRV[1:0] = 01, Medium Low drive capability	-	-	0.75	
		LSEDRV[1:0] = 10,	-	-	1.7	

Symbol	Parameter	Operating conditions[1]	Min[2]	Typ[2]	Max[2]	Unit
$Gm_{critmax}$	Maximum critical crystal gm	Medium high drive capability				µA/V
		LSEDRV[1:0] = 11, High drive capability	-	-	2.7	
t_{SU}[3]	Startup time	VDD is stabilized	-	2	-	s

1. *Refer to the note and caution paragraphs below the table, and to the application note AN2867 “Oscillator design guide for ST microcontrollers.*
2. *Guaranteed by design.*
3. *t_{SU} is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768k Hz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer.*

Note: For information on selecting the crystal, refer to the application note AN2867 “Oscillator design guide for ST microcontrollers” available from the ST website www.st.com.

Figure 24. Typical application with a 32.768 kHz crystal



1. An external resistor is not required between OSC32_IN and OSC32_OUT and it is forbidden to add one.

6.3.10 Internal clock source characteristics

The parameters given in Table 49. HSI48 oscillator characteristics to Table 52. LSI oscillator characteristics are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in Table 20. General operating conditions.

48 MHz high-speed internal RC oscillator (HSI48)

Table 49. HSI48 oscillator characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{HSI48}	HSI48 frequency	V_{DD} = 3.3 V, T_J = 30 °C	47.5[1]	48	48.5[1]	MHz
TRIM[2]	User trimming step	-	-	0.175	0.250	%
USER TRIM COVERAGE[3]	User trimming coverage	± 32 steps	±4,70	±5.6	-	%
DuCy(HSI48)[2]	Duty cycle	-	45	-	55	%
ACCHSI48_REL[3]	Accuracy of the HSI48 oscillator over temperature (reference is 30 °C)	T_J = −40 to 130 °C	−4.5	-	4	%
ΔV_{DD}(HSI48)[2]	HSI48 oscillator frequency drift with V_{DD} (reference is 3.3 V)	V_{DD} = 3 to 3.6 V	-	0.025	0.05	%
		V_{DD} = 1.62 to 3.6 V	-	0.05	0.1	
t_{su}(HSI48)[2]	HSI48 oscillator startup time	-	-	2.1	4.0	µs
I_{DD}(HSI48)[2]	HSI48 oscillator power consumption	-	-	350	400	µA
N_T jitter[2]	Next transition jitter accumulated jitter on 28 cycles	-	-	± 0.15	-	ns

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
P_T jitter(2)	Paired transition jitter Accumulated jitter on 56 cycles(6)	-	-	± 0.25	-	ns

1. *Calibrated during manufacturing tests.*
2. *Guaranteed by design.*
3. *Guaranteed by characterization results.*
4. *Δf_{HSI} = ACCHSI48_REL + ΔV_{DD}*
5. *These values are obtained by using the formula: (Freq(3.6 V) − Freq(3.0 V)) / Freq(3.0 V) or (Freq(3.6 V) − Freq(1.62 V)) / Freq(1.62 V).*
6. *Jitter measurements are performed without clock sources activated in parallel.*

64 MHz high-speed internal RC oscillator (HSI)

Table 50. HSI oscillator characteristics

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)	Unit
f_{HSI}	HSI frequency	V_{DD}=3.3 V, T_J=30 °C	63.7(2)	64	64.3(2)	MHz
TRIM	HSI user trimming step	Trimming is not a multiple of 32(3)	-	0.24	0.32	%
		Trimming is 128, 256 and 384(3)	−5.2	−1.8	-	
		Trimming is 64, 192, 320 and 448(3)	−1.4	−0.8	-	
		Other trimming are a multiple of 32 (not including multiple of 64 and 128)(3)	−0.6	−0.25	-	
DuCy(HSI)	Duty Cycle	-	45	-	55	%
$\Delta_{VDD\ (HSI)}$	HSI oscillator frequency drift over V_{DD} (reference is 3.3 V)	V_{DD}=1.62 to 3.6 V	−0.12	-	0.03	%
$\Delta_{TEMP\ (HSI)}$	HSI oscillator frequency drift over temperature (reference is 64 MHz)	T_J=-20 to 105 °C	−1(4)	-	1(4)	%
		T_J=−40 to $T_J$max °C	−2(4)	-	1(4)	
t_{su}(HSI)	HSI oscillator start-up time	-	-	1.4	2	µs
t_{stab}(HSI)	HSI oscillator stabilization time	at 1 % of target frequency	-	4	8	µs
		at 5 % of target frequency	-	-	4	
I_{DD}(HSI)	HSI oscillator power consumption	-	-	300	400	µA

1. *Guaranteed by design, unless otherwise specified.*
2. *Calibrated during manufacturing tests.*
3. *Trimming value of HSICAL[8:0] (refer to RM0455).*
4. *Guaranteed by characterization results.*

4 MHz low-power internal RC oscillator (CSI)

Table 51. CSI oscillator characteristics

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)	Unit
f_{CSI}	CSI frequency	V_{DD} = 3.3 V, T_J = 30 °C	3.96(2)	4	4.04(2)	MHz
TRIM	CSI user trimming step	Trimming is not a multiple of 16	-	0.40	0.75	-
		Trimming is a multiple of 32	−4,75	−2,75	0.75	-
		Other trimming are a multiple of 16 (not including multiple of 32)	−0,43	0.00	0.75	%
DuCy(CSI)	Duty Cycle	-	45	-	55	%
Δ_{TEMP} (CSI)	CSI oscillator frequency drift over temperature	T_J = 0 to 85 °C	−3.7(3)	-	4,5(3)	%
		T_J = −40 to 130 °C	−11(3)	-	7,5(3)	

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)	Unit
ΔV_{DD}(CSI)	CSI oscillator frequency drift over V_{DD}	V_{DD} = 1.62 to 3.6 V	−0.06	-	0.06	%
t_{su}(CSI)	CSI oscillator startup time	-	-	1	2	µs
t_{stab}(CSI)	CSI oscillator stabilization time (to reach ± 3 % of f_{CSI})	-	-	-	4	cycle
I_{DD}(CSI)	CSI oscillator power consumption	-	-	23	30	µA

1. *Guaranteed by design, unless otherwise specified.*
2. *Calibrated during manufacturing tests.*
3. *Guaranteed by characterization results.*

Low-speed internal (LSI) RC oscillator

Table 52. LSI oscillator characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{LSI}	LSI frequency	V_{DD} = 3.3 V, T_J = 25 °C	31,4(1)	32	32,6(1)	kHz
		T_J = –40 to 110 °C, V_{DD} = 1.62 to 3.6 V	29,76(2)	-	33,6(2)	
		T_J = –40 to 130 °C, V_{DD} = 1.62 to 3.6 V	29,4(2)	-	33,6(2)	
t_{su}(LSI)(3)	LSI oscillator startup time	-	-	80	130	µs
t_{stab}(LSI)(3)	LSI oscillator stabilization time (5% of final value)	-	-	120	170	
I_{DD}(LSI)(3)	LSI oscillator power consumption	-	-	130	280	nA

1. *Calibrated during manufacturing tests.*
2. *Guaranteed by characterization results.*
3. *Guaranteed by design.*

6.3.11 PLL characteristics

The parameters given in Table 53. PLL characteristics (wide VCO frequency range) are derived from tests performed under temperature and V_{DD} supply voltage conditions summarized in Table 20. General operating conditions.

Table 53. PLL characteristics (wide VCO frequency range)

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)	Unit
f_{PLL_IN}	PLL input clock	-	2	-	16	MHz
	PLL input clock duty cycle	-	10	-	90	%
$f_{PLL_P_OUT}$	PLL multiplier output clock P, Q, R	VOS0	1	-	280(2)	MHz
		VOS1	1	-	225(2)	
		VOS2	1	-	160(2)	
		VOS3	1	-	88(2)	
f_{VCO_OUT}	PLL VCO output	-	128	-	560(3)	
t_{LOCK}	PLL lock time	Normal mode	-	45	100(3)	µs
		Sigma-delta mode (f_{PLL_IN} ≥ 8 MHz)	-	60	120(3)	
Jitter	Cycle-to-cycle jitter	f_{VCO_OUT} = 128 MHz	-	60	-	±ps
		f_{VCO_OUT} = 200 MHz	-	50	-	
		f_{VCO_OUT} = 400 MHz	-	20	-	

Symbol	Parameter	Conditions		Min(1)	Typ(1)	Max(1)	Unit
Jitter	Cycle-to-cycle jitter	f_{VCO_OUT} = 560 MHz		-	15	-	±ps
	Long term jitter	Normal mode (f_{PLL_IN} = 2 MHz), f_{VCO_OUT} = 560 MHz		-	±0.2	-	%
		Normal mode (f_{PLL_IN} = 16 MHz), f_{VCO_OUT} = 560 MHz		-	±0.8	-	
		Sigma-delta mode (f_{PLL_IN} = 2 MHz), f_{VCO_OUT} = 560 MHz		-	±0.2	-	
		Sigma-delta mode (f_{PLL_IN} = 16 MHz), f_{VCO_OUT} = 560 MHz		-	±0.8	-	
$I_{DD(PLL)}$	PLL power consumption	f_{VCO_OUT} = 560 MHz	V_{DD}	-	330	420	µA
			V_{CORE}	-	630	-	
		f_{VCO_OUT} = 128 MHz	V_{DD}	-	155	230	
			V_{CORE}	-	170	-	

1. *Guaranteed by design, unless otherwise specified.*
2. *This value must be limited to the maximum frequency due to the product limitation.*
3. *Guaranteed by characterization results.*

Table 54. PLL characteristics (medium VCO frequency range)

Symbol	Parameter	Conditions		Min(1)	Typ(1)	Max(1)	Unit
f_{PLL_IN}	PLL input clock	-		1	-	2	MHz
	PLL input clock duty cycle	-		10	-	90	%
f_{PLL_OUT}	PLL multiplier output clock P, Q, R	VOS0		1.17	-	210	MHz
		VOS1		1.17	-	210	
		VOS2		1.17	-	160(2)	
		VOS3		1.17	-	88(2)	
f_{VCO_OUT}	PLL VCO output	-		150	-	420	
t_{LOCK}	PLL lock time	Normal mode		-	45	80(3)	µs
		Sigma-delta mode		forbidden			
Jitter	Cycle-to-cycle jitter	f_{VCO_OUT} = 150 MHz	-	-	60	-	±ps
		f_{VCO_OUT} = 200 MHz	-	-	40	-	
		f_{VCO_OUT} = 400 MHz	-	-	18	-	
		f_{VCO_OUT} = 420 MHz	-	-	15	-	
	Period jitter	f_{VCO_OUT} = 150 MHz	f_{PLL_OUT} = 50 MHz	-	75	-	±-ps
		f_{VCO_OUT} = 400 MHz		-	25	-	
	Long term jitter	Normal mode, f_{VCO_OUT} = 400 MHz		-	±0.2	-	%
$I_{DD(PLL)}$	PLL power consumption on V_{DD}	f_{VCO_OUT} = 420 MHz	V_{DD}	-	275	360	µA
			V_{CORE}	-	450	-	
		f_{VCO_OUT} = 150 MHz	V_{DD}	-	160	240	
			V_{CORE}	-	165	-	

1. *Guaranteed by design, unless otherwise specified.*
2. *This value must be limited to the maximum frequency due to the product limitation.*
3. *Guaranteed by characterization results.*

6.3.12 Memory characteristics

Flash memory

The characteristics are given at T_J = –40 to 130 °C unless otherwise specified.
The devices are shipped to customers with the flash memory erased.

Table 55. Flash memory characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
I_{DD}	Supply current	Word program	-	2.5	4	mA
		Sector erase	-	1.8	3	
		Mass erase	-	2.0	3	

Table 56. Flash memory programming

Symbol	Parameter	Conditions	Min	Typ	Max(1)	Unit
t_{prog}	Word program time	128 bits (user area)	-	-	20	µs
		16 bits (OTP area)	-	-	20	
$t_{ERASE8KB}$	Sector erase time (8 Kbytes)	-	-	-	2.2	ms
t_{ME}	Bank mass erase time		-	-	10	
V_{prog}	Programming voltage		1.62	-	3.6	V

1. *Guaranteed by characterization results.*

Table 57. Flash memory endurance and data retention

Symbol	Parameter	Conditions	Value	Unit
			Min(1)	
N_{END}	Endurance	T_J = –40 to +130 °C	10	kcycles
t_{RET}	Data retention	1 kcycle at T_A = 85 °C	30	Years
	-	10 kcycles at T_A = 55 °C	20	

1. *Guaranteed by characterization results.*

6.3.13 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports). the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- **Electrostatic discharge (ESD)** (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- **FTB**: A burst of fast transient voltage (positive and negative) is applied to V_{DD} and V_{SS} through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in Table 58. EMS characteristics. They are based on the EMS levels and classes defined in application note AN1709.

Table 58. EMS characteristics

Symbol	Parameter	Conditions	Level/ Class
V_{FESD}	Voltage limits to be applied on any I/O pin to induce a functional disturbance	V_{DD} = 3.3 V, T_A = +25 °C, LQFP144, $f_{rcc_cpu_ck}$ = 216 MHz, conforms to IEC 61000-4-2	3B
V_{FTB}	Fast transient voltage burst limits to be applied through 100 pF on V_{DD} and V_{SS} pins to induce a functional disturbance		5A

As a consequence, it is recommended to add a serial resistor (1 kΩ) located as close as possible to the MCU to the pins exposed to noise (connected to tracks longer than 50 mm on PCB).

Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical Data corruption (control registers...)

Prequalification trials

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the Oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

Electromagnetic Interference (EMI)

The electromagnetic field emitted by the device are monitored while a simple application, executing EEMBC code, is running. This emission test is compliant with SAE IEC61967-2 standard which specifies the test board and the pin loading.

Table 59. EMI characteristics for f_{HSE} = 8 MHz and f_{HCLK} = 64 MHz

Symbol	Parameter	Conditions	Monitored frequency band	Value	Unit
S_{EMI}	Peak(1)	V_{DD} = 3.6 V, T_A = 25 °C, LQFP64 package, compliant with IEC 61967-2	0.1 to 30 MHz	7	dBµV
			30 to 130 MHz	–1	
			130 MHz to 1 GHz	8	
			1 GHz to 2 GHz	7	
	Level(2)		0.1 GHz to 2 GHz	2.5	-

1. *Refer to the EMI radiated test chapter of application note AN1709 "EMC design guide for STM8, STM32 and Legacy MCUs" available from the ST website www.st.com..*
2. *Refer to the EMI level classification chapter of application note AN1709 "EMC design guide for STM8, STM32 and Legacy MCUs" available from the ST website www.st.com.*

6.3.14 Absolute maximum ratings (electrical sensitivity)

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse) are applied to the pins of each sample according to each pin combination. This test conforms to the ANSI/ESDA/JEDEC JS-001 and ANSI/ESDA/JEDEC JS-002 standards.

Table 60. ESD absolute maximum ratings

Symbol	Ratings	Conditions	Packages	Class	Maximum value	Unit
$V_{ESD(HBM)}$	Electrostatic discharge voltage (human body model)	T_A = +25 °C conforming to ANSI/ESDA/JEDEC JS-001	Packages with SMPS	1C	1000(2)	V
			Packages without SMPS	2	2000	
$V_{ESD(CDM)}$	Electrostatic discharge voltage (charge device model)	T_A = +25 °C conforming to ANSI/ESDA/JEDEC JS-002	All LQFP packages	C1	250	
			All BGA packages	C2a	500	

1. *Guaranteed by characterization results.*
2. *The electrostatic discharge is 2000 V for all pins, except V_{FBSMPS}, for which the test fails at 2000 V and passes at 1600 V.*

Static latchup

Two complementary static tests are required on six parts to assess the latchup performance:

- A supply overvoltage is applied to each power supply pin
- A current injection is applied to each input, output and configurable I/O pin

These tests are compliant with JESD78 IC latchup standard.

Table 61. Electrical sensitivities

Symbol	Parameter	Conditions	Class
LU	Static latchup class	T_J= +130 °C, conforming to JESD78	II level A

6.3.15 I/O current injection characteristics

As a general rule, a current injection to the I/O pins, due to external voltage below V_{SS} or above V_{DD} (for standard, 3.3 V-capable I/O pins) should be avoided during the normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when an abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during the device characterization.

Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error above a certain limit (higher than 5 LSB TUE), out of conventional limits of induced leakage current on adjacent pins (out of –5 µA/+0 µA range), or other functional failure (for example reset, oscillator frequency deviation).

The following tables are the compilation of the SIC1/SIC2 and functional ESD results.

Negative induced A negative induced leakage current is caused by negative injection and positive induced leakage current by positive injection.

Table 62. I/O current injection susceptibility

Symbol	Description	Functional susceptibility		Unit
		Negative injection	Positive injection	
I_{INJ}	PF2, PI12	0	NA	mA
	PG1, PE9, PB0, PA7, PC4, PC5, PE7, PE8, PA4, PA5, PA6, PF2, PI12, PC2_C, PC3_C, PA0_C, PA1_C, BOOT0	0	0	
	All other I/Os	5	NA	

6.3.16 I/O port characteristics

General input/output characteristics

Unless otherwise specified, the parameters given in Table 63. I/O static characteristics are derived from tests performed under the conditions summarized in Table 20. General operating conditions. All I/Os are CMOS and TTL compliant (except for BOOT0).

Note: For information on GPIO configuration, refer to the application note AN4899 "STM32 GPIO configuration for hardware settings and low-power consumption" available from the ST website www.st.com.

Table 63. I/O static characteristics

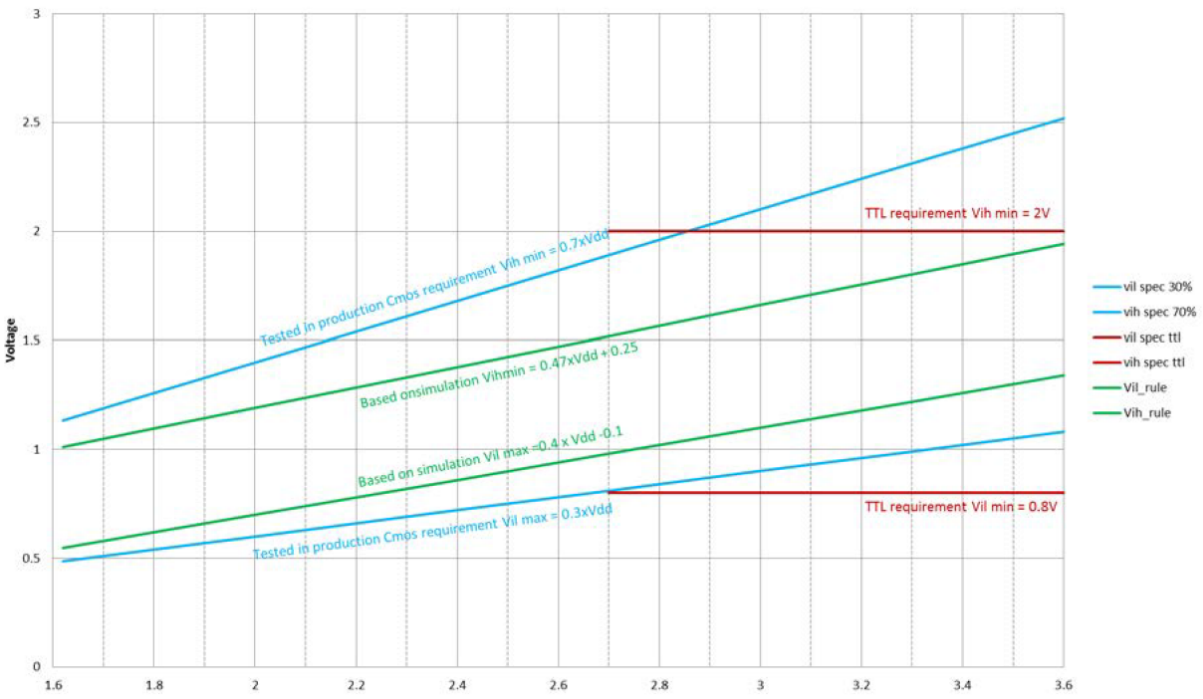
Symbol	Parameter	Condition	Min	Typ	Max	Unit
V_{IL}	I/O input low-level voltage except BOOT0	1.62 V < V_{DDIOx} < 3.6 V	-	-	$0.3V_{DD}$ (1)	V
	I/O input low-level voltage except BOOT0		-	-	$0.4V_{DD}-0.1$(2)	
	BOOT0 I/O input low level voltage		-	-	$0.19V_{DD}+0.1$(2)	
V_{IH}	I/O input high level voltage except BOOT0	1.62 V < V_{DDIOx} < 3.6 V	$0.7V_{DD}$ (1)	-	-	V
	I/O input high level voltage except BOOT0		$0.47V_{DD}+0.25$(2)	-	-	
	BOOT0 I/O input high level voltage		$0.17V_{DD}+0.6$(2)	-	-	
V_{HYS}(2)	TT_xx, FT_xxx and NRST I/O input hysteresis	1.62 V < V_{DDIOx} < 3.6 V	-	250	-	mV
	BOOT0 I/O input hysteresis		-	200	-	
I_{lkg}	FT_xx input leakage current(2)	0 < V_{IN} ≤ Max(V_{DDxxx})(5)	-	-	±250	nA
		Max(V_{DDxxx}) < V_{IN} ≤ 5.5 V (3)(4)(5)	-	-	1500	
	FT_u I/O	0 < V_{IN} ≤ Max(V_{DDxxx})(5)	-	-	±350	

Symbol	Parameter	Condition	Min	Typ	Max	Unit
I_{lkg}	FT_u I/O	$Max(V_{DDxxx}) < V_{IN} \le 5.5$ V (3)(4)(8)(5)	-	-	5000(6)	nA
	TT_xx input leakage current	$0 < V_{IN} \le Max(V_{DDxxx})$(5)	-	-	±250	
	VPP (BOOT0 alternate function)	$0 < V_{IN} \le V_{DDIOx}$	-	-	15	uA
		$V_{DDIOx} < V_{IN} \le 9$ V	-	-	35	
RPU	Weak pull-up equivalent resistor(7)	$V_{IN} = V_{SS}$	30	40	50	kΩ
RPD	Weak pull-down equivalent resistor(7)	$V_{IN} = V_{DD}$(5)	30	40	50	
CIO	I/O pin capacitance	-	-	5	-	pF

1. *Compliant with CMOS requirements.*
2. *Guaranteed by design.*
3. *All FT_xx IO except FT_lu and FT_u.*
4. *V_{IN} must be less than $Max(V_{DDxxx})$ + 3.6 V.*
5. *$Max(V_{DDxxx})$ is the maximum value of all the I/O supplies.*
6. *To sustain a voltage higher than $MIN(V_{DD}, V_{DDA}, V_{DD33USB})$ +0.3 V, the internal pull-up and pull-down resistors must be disabled.*
7. *The pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This PMOS/NMOS contribution to the series resistance is minimal (~10%).*

All I/Os are CMOS and TTL compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters. The coverage of these requirements for FT I/Os is shown in Figure 25. V_{IL}/V_{IH} for all I/Os except BOOT0.

Figure 25. V_{IL}/V_{IH} for all I/Os except BOOT0



Output driving current

The GPIOs (general purpose input/outputs) can sink or source up to ±8 mA, and sink or source up to ±20 mA (with a relaxed V_{OL}/V_{OH}).

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in Section 6.2 Absolute maximum ratings. In particular:

- The sum of the currents sourced by all the I/Os on V_{DD}, plus the maximum Run consumption of the MCU sourced on V_{DD}, cannot exceed the absolute maximum rating ΣI_{VDD} (see Table 18. Current characteristics).
- The sum of the currents sunk by all the I/Os on V_{SS} plus the maximum Run consumption of the MCU sunk on V_{SS} cannot exceed the absolute maximum rating ΣI_{VSS} (see Table 18. Current characteristics).

Output voltage levels

Unless otherwise specified, the parameters given in Table 64. Output voltage characteristics for all I/Os except PC13, PC14, PC15 and PI8 and Table 65. Output voltage characteristics for PC13, PC14, PC15 and PI8 are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in Table 20. General operating conditions. All I/Os are CMOS and TTL compliant.

Table 64. Output voltage characteristics for all I/Os except PC13, PC14, PC15 and PI8

The IIO current sourced or sunk by the device must always respect the absolute maximum rating specified in Table 17. Voltage characteristics, and the sum of the currents sourced or sunk by all the I/Os (I/O ports and control pins) must always respect the absolute maximum ratings ΣIIO.

Symbol	Parameter	Conditions(1)	Min	Max	Unit
V_{OL}	Output low level voltage	CMOS port(2) I_{IO}=8 mA 2.7 V≤ V_{DD} ≤3.6 V	-	0.4	V
V_{OH}	Output high level voltage	CMOS port(2) I_{IO}= −8 mA 2.7 V≤ V_{DD} ≤3.6 V	V_{DD}−0.4	-	
V_{OL}(3)	Output low level voltage	TTL port(2) I_{IO}=8 mA 2.7 V≤ V_{DD} ≤3.6 V	-	0.4	
V_{OH}(3)	Output high level voltage	TTL port(2) I_{IO}=-8 mA 2.7 V≤ V_{DD} ≤3.6 V	2.4	-	
V_{OL}(3)	Output low level voltage	I_{IO}=20 mA 2.7 V≤ V_{DD} ≤3.6 V	-	1.3	
V_{OH}(3)	Output high level voltage	I_{IO}=−20 mA 2.7 V≤ V_{DD} ≤3.6 V	V_{DD}−1.3	-	
V_{OL}(3)	Output low level voltage	I_{IO}= 4 mA 1.62 V≤ V_{DD} ≤3.6 V	-	0.4	
V_{OH} (3)	Output high level voltage	I_{IO}= −4 mA 1.62 V≤V_{DD}<3.6 V	V_{DD}−0.4	-	
V_{OLFM+}(3)	Output low level voltage for an FTf I/O pin in FM+ mode	I_{IO}= 20 mA 2.3 V≤ V_{DD}≤3.6 V	-	0.4	
		I_{IO}= 10 mA 1.62 V≤ V_{DD} ≤3.6 V	-	0.4	

1. *The I_{IO} current sourced or sunk by the device must always respect the absolute maximum rating specified in Table 17. Voltage characteristics, and the sum of the currents sourced or sunk by all the I/Os (I/O ports and control pins) must always respect the absolute maximum ratings ΣI_{IO}.*
2. *TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.*
3. *Guaranteed by design.*

Table 65. Output voltage characteristics for PC13, PC14, PC15 and PI8

Symbol	Parameter	Conditions(1)	Min	Max	Unit
V_{OL}	Output low level voltage	CMOS port(2) I_{IO}=3 mA, 2.7 V≤ VDD ≤ 3.6 V	-	0.4	V
V_{OH}	Output high level voltage	CMOS port(2) I_{IO}= −3 mA, 2.7 V≤ V_{DD} ≤ 3.6 V	V_{DD}−0.4	-	

Symbol	Parameter	Conditions[1]	Min	Max	Unit
V_{OL}[3]	Output low level voltage	TTL port[2] I_{IO} = 3 mA, 2.7 V≤ V_{DD} ≤ 3.6 V	-	0.4	V
V_{OH}[3]	Output high level voltage	TTL port[2] I_{IO}=−3 mA, 2.7 V ≤ V_{DD} ≤ 3.6 V	2.4	-	
V_{OL}[3]	Output low level voltage	I_{IO} = 1.5 mA, 1.62 V ≤ V_{DD} ≤ 3.6 V	-	0.4	
V_{OH} [3]	Output high level voltage	I_{IO} = −1.5 mA, 1.62 V ≤ V_{DD} < 3.6 V	VDD−0.4	-	

1. *The I_{IO} current sourced or sunk by the device must always respect the absolute maximum rating specified in Table 17. Voltage characteristics, and the sum of the currents sourced or sunk by all the I/Os (I/O ports and control pins) must always respect the absolute maximum ratings ΣI_{IO}.*
2. *TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.*
3. *Guaranteed by design.*

Output buffer timing characteristics (HSLV option disabled)

The HSLV bit of SYSCFG_CCCSR register can be used to optimize the I/O speed when the product voltage is below 2.7 V.

Table 66. Output timing characteristics (HSLV OFF)

The frequency of the GPIOs that can be supplied in VBAT mode (PC13, PC14, PC15 and PI8) is limited to 2 MHz.

Speed	Symbol	Parameter	conditions	Min[1]	Max[1]	Unit
00	F_{max}[2]	Maximum frequency	C=50 pF, 2.7 V≤ V_{DD}≤3.6 V	-	12	MHz
			C=50 pF, 1.62 V≤V_{DD}≤2.7 V	-	3	
			C=30 pF, 2.7 V≤V_{DD}≤3.6 V	-	12	
			C=30 pF, 1.62 V≤V_{DD}≤2.7 V	-	3	
			C=10 pF, 2.7 V≤V_{DD}≤3.6 V	-	16	
			C=10 pF, 1.62 V≤V_{DD}≤2.7 V	-	4	
	t_r/t_f [3]	Output high to low level fall time and output low to high level rise time	C=50 pF, 2.7 V≤ V_{DD}≤3.6 V	-	16.6	ns
			C=50 pF, 1.62 V≤V_{DD}≤2.7 V	-	33.3	
			C=30 pF, 2.7 V≤V_{DD}≤3.6 V	-	13.3	
			C=30 pF, 1.62 V≤V_{DD}≤2.7 V	-	25	
			C=10 pF, 2.7 V≤V_{DD}≤3.6 V	-	10	
			C=10 pF, 1.62 V≤V_{DD}≤2.7 V	-	20	
01	F_{max}[2]	Maximum frequency	C=50 pF, 2.7 V≤ V_{DD}≤3.6 V	-	60	MHz
			C=50 pF, 1.62 V≤V_{DD}≤2.7 V	-	15	
			C=30 pF, 2.7 V≤V_{DD}≤3.6 V	-	80	
			C=30 pF, 1.62 V≤V_{DD}≤2.7 V	-	15	
			C=10 pF, 2.7 V≤V_{DD}≤3.6 V	-	110	
			C=10 pF, 1.62 V≤V_{DD}≤2.7 V	-	20	
	t_r/t_f [3]	Output high to low level fall time and output low to high level rise time	C=50 pF, 2.7 V≤ V_{DD}≤3.6 V	-	5.2	ns
			C=50 pF, 1.62 V≤V_{DD}≤2.7 V	-	10	
			C=30 pF, 2.7 V≤V_{DD}≤3.6 V	-	4.2	
			C=30 pF, 1.62 V≤V_{DD}≤2.7 V	-	7.5	
			C=10 pF, 2.7 V≤V_{DD}≤3.6 V	-	2.8	
			C=10 pF, 1.62 V≤V_{DD}≤2.7 V	-	5.2	

Speed	Symbol	Parameter	conditions	Min(1)	Max(1)	Unit
10	F_{max}(2)	Maximum frequency	C=50 pF, 2.7 V≤V_{DD}≤3.6 V(4)	-	85	MHz
			C=50 pF, 1.62 V≤V_{DD}≤2.7 V(4)	-	35	
			C=30 pF, 2.7 V≤V_{DD}≤3.6 V(4)	-	110	
			C=30 pF, 1.62 V≤V_{DD}≤2.7 V(4)	-	40	
			C=10 pF, 2.7 V≤V_{DD}≤3.6 V(4)	-	166	
			C=10 pF, 1.62 V≤V_{DD}≤2.7 V(4)	-	100	
	t_r/t_f (3)	Output high to low level fall time and output low to high level rise time	C=50 pF, 2.7 V≤V_{DD}≤3.6 V(4)	-	3.8	ns
			C=50 pF, 1.62 V≤V_{DD}≤2.7 V(4)	-	6.9	
			C=30 pF, 2.7 V≤V_{DD}≤3.6 V(4)	-	2.8	
			C=30 pF, 1.62 V≤V_{DD}≤2.7 V(4)	-	5.2	
			C=10 pF, 2.7 V≤V_{DD}≤3.6 V(4)	-	1.8	
			C=10 pF, 1.62 V≤V_{DD}≤2.7 V(4)	-	3.3	
11	F_{max}(2)	Maximum frequency	C=50 pF, 2.7 V≤V_{DD}≤3.6 V(4)	-	100	MHz
			C=50 pF, 1.62 V≤V_{DD}≤2.7 V(4)	-	50	
			C=30 pF, 2.7 V≤V_{DD}≤3.6 V(4)	-	133	
			C=30 pF, 1.62 V≤V_{DD}≤2.7 V(4)	-	66	
			C=10 pF, 2.7 V≤V_{DD}≤3.6 V(4)	-	220	
			C=10 pF, 1.62 V≤V_{DD}≤2.7 V(4)	-	85	
	t_r/t_f (3)	Output high to low level fall time and output low to high level rise time	C=50 pF, 2.7 V≤V_{DD}≤3.6 V(4)	-	3.3	ns
			C=50 pF, 1.62 V≤V_{DD}≤2.7 V(4)	-	6.6	
			C=30 pF, 2.7 V≤V_{DD}≤3.6 V(4)	-	2.4	
			C=30 pF, 1.62 V≤V_{DD}≤2.7 V(4)	-	4.5	
			C=10 pF, 2.7 V≤V_{DD}≤3.6 V(4)	-	1.5	
			C=10 pF, 1.62 V≤V_{DD}≤2.7 V(4)	-	2.7	

1. *Guaranteed by design.*
2. *The maximum frequency is defined with the following conditions: (t_r+t_f) ≤ 2/3 T, skew ≤ 1/20 T, 45%<Duty cycle<55%*
3. *The fall and rise times are defined between 90% and 10% and between 10% and 90% of the output waveform, respectively.*
4. *Compensation system enabled.*

Output buffer timing characteristics (HSLV option enabled)

Table 67. Output timing characteristics (HSLV ON)

Speed	Symbol	Parameter	conditions	Min(1)	Max(1)	Unit
00	F_{max}(2)	Maximum frequency	C=50 pF, 1.62 V≤VDD≤2.7 V	-	10	MHz
			C=30 pF, 1.62 V≤VDD≤2.7 V	-	10	
			C=10 pF, 1.62 V≤VDD≤2.7 V	-	10	
	t_r/t_f(3)	Output high to low level fall time and output low to high level rise time	C=50 pF, 1.62 V≤VDD≤2.7 V	-	11	ns
			C=30 pF, 1.62 V≤VDD≤2.7 V	-	9	
			C=10 pF, 1.62 V≤VDD≤2.7 V	-	6.6	

Speed	Symbol	Parameter	conditions	Min(1)	Max(1)	Unit
01	F_{max}(2)	Maximum frequency	C=50 pF, 1.62 V≤VDD≤2.7 V	-	50	MHz
			C=30 pF, 1.62 V≤VDD≤2.7 V	-	58	
			C=10 pF, 1.62 V≤VDD≤2.7 V	-	66	
	t_r/t_f(3)	Output high to low level fall time and output low to high level rise time	C=50 pF, 1.62 V≤VDD≤2.7 V	-	6.6	ns
			C=30 pF, 1.62 V≤VDD≤2.7 V	-	4.8	
			C=10 pF, 1.62 V≤VDD≤2.7 V	-	3	
10	F_{max}(2)	Maximum frequency	C=50 pF, 1.62 V≤VDD≤2.7 V(4)	-	55	MHz
			C=30 pF, 1.62 V≤VDD≤2.7 V(4)	-	80	
			C=10 pF, 1.62 V≤VDD≤2.7 V(4)	-	133	
	t_r/t_f(3)	Output high to low level fall time and output low to high level rise time	C=50 pF, 1.62 V≤VDD≤2.7 V(4)	-	5.8	ns
			C=30 pF, 1.62 V≤VDD≤2.7 V(4)	-	4	
			C=10 pF, 1.62 V≤VDD≤2.7 V(4)	-	2.4	
11	F_{max}(2)	Maximum frequency	C=50 pF, 1.62 V≤VDD≤2.7 V(4)	-	60	MHz
			C=30 pF, 1.62 V≤VDD≤2.7 V(4)	-	90	
			C=10 pF, 1.62 V≤VDD≤2.7 V(4)	-	175	
	t_r/t_f(3)	Output high to low level fall time and output low to high level rise time	C=50 pF, 1.62 V≤VDD≤2.7 V(4)	-	5.3	ns
			C=30 pF, 1.62 V≤VDD≤2.7 V(4)	-	3.6	
			C=10 pF, 1.62 V≤VDD≤2.7 V(4)	-	1.9	

1. *Guaranteed by design.*
2. *The maximum frequency is defined with the following conditions: (t_r+t_f) ≤ 2/3 T, skew ≤ 1/20 T, 45%<Duty cycle<55%*
3. *The fall and rise times are defined between 90% and 10% and between 10% and 90% of the output waveform, respectively.*
4. *Compensation system enabled.*

Analog switch between ports Pxy_C and Pxy

PA0_C, PA1_C, PC2_C and PC3_C can be connected internally to PA0, PA1, PC2 and PC3, respectively (refer to SYSCFG_PMCR register in RM0468 reference manual). The switch is controlled by VDDSWITCH voltage level. It is defined through BOOSTVDDSEL bit of SYSCFG_PMCR. If the switch is closed the switch characteristics are given in the table below.

Table 68. Pxy_C and Pxy analog switch characteristics

Parameter	Conditions		Min	Typ	Max	Unit
Switch impedance	Switch control boosted		-	-	315	Ω
	Switch control not boosted	$V_{DDSWITCH}$ > 2.7 V	-	-	315	
		$V_{DDSWITCH}$ > 2.4 V	-	-	335	
		$V_{DDSWITCH}$ > 2.0 V	-	-	390	
		$V_{DDSWITCH}$ > 1.8 V	-	-	445	
		$V_{DDSWITCH}$ > 1.62 V	-	-	550	

6.3.17 NRST pin characteristics

The NRST pin input driver uses CMOS technology. It is connected to a permanent pull-up resistor, R_{PU} (see Table 63. I/O static characteristics).

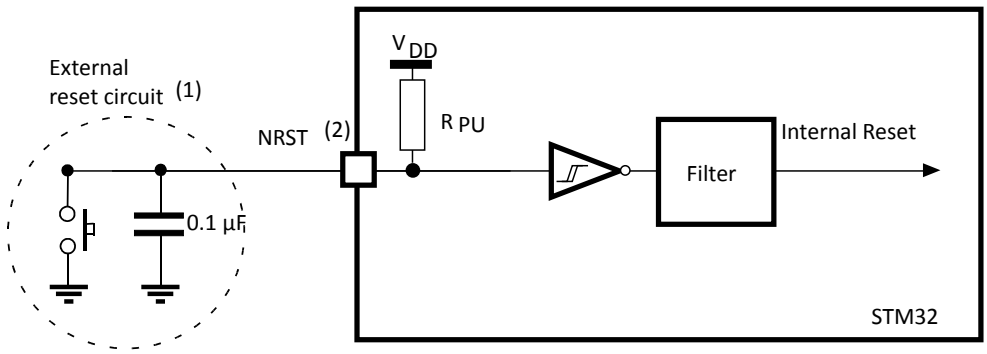
Unless otherwise specified, the parameters given in Table 69. NRST pin characteristics are derived from tests performed under the ambient temperature and V_{DD} supply voltage conditions summarized in Table 20. General operating conditions.

Table 69. NRST pin characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
R_{PU}(1)	Weak pull-up equivalent resistor(2)	$V_{IN} = V_{SS}$	30	40	50	kΩ
$V_{F(NRST)}$(1)	NRST Input filtered pulse	1.71 V < V_{DD} < 3.6 V	-	-	50	ns
$V_{NF(NRST)}$(1)	NRST Input not filtered pulse	1.71 V < V_{DD} < 3.6 V	350	-	-	ns
		1.62 V < V_{DD} < 3.6 V	1000	-	-	ns

1. *Guaranteed by design.*
2. *The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance must be minimum (~10%).*

Figure 26. Recommended NRST pin protection



1. The reset network protects the device against parasitic resets.
2. The user must ensure that the level on the NRST pin can go below the $V_{IL(NRST)}$ max level specified in Table 63. I/O static characteristics. Otherwise the reset is not taken into account by the device.

6.3.18 FMC characteristics

Unless otherwise specified, the parameters given in the below tables for the FMC interface are derived from tests performed under the ambient temperature, f_{HCLK} frequency and V_{DD} supply voltage conditions summarized in Table 20. General operating conditions, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 11
- Measurement points are done at CMOS levels: $0.5V_{DD}$
- IO Compensation cell activated.
- HSLV activated when $V_{DD} \leq 2.7$ V
- VOS level set to VOS0.

Note: At VOS1, the performance in some FMC modes can be degraded by up to 5 % compared to VOS0. This is indicated by a footnote when applicable.

Refer to Section 6.3.16 I/O port characteristics for more details on the input/output alternate function characteristics.

Asynchronous waveforms and timings

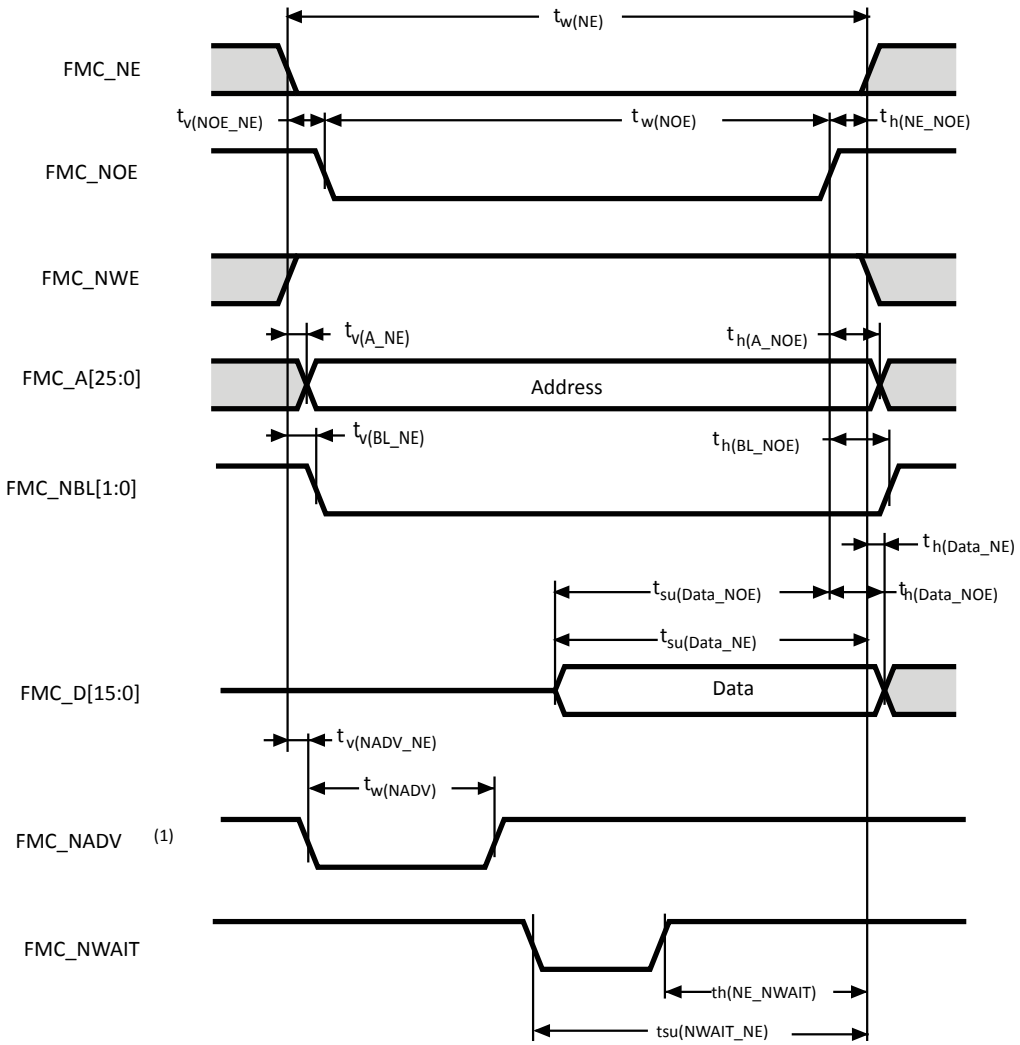
Figure 27 through Figure 29 represent asynchronous waveforms and Table 70 through Table 77 provide the corresponding timings. The results shown in these tables are obtained with the following FMC configuration:

- AddressSetupTime = 0x1
- AddressHoldTime = 0x1
- DataSetupTime = 0x1 (except for asynchronous NWAIT mode , DataSetupTime = 0x5)
- BusTurnAroundDuration = 0x0

- Capacitive load C_L = 30 pF

In all timing tables, $T_{fmc_ker_ck}$ is the kernel clock period.

Figure 27. Asynchronous non-multiplexed SRAM/PSRAM/NOR read waveforms



1. Mode 2/B, C and D only. In Mode 1, FMC_NADV is not used.

Table 70. Asynchronous non-multiplexed SRAM/PSRAM/NOR read timings

Symbol	Parameter	Min(1)	Max(1)	Unit
$t_{w(NE)}$	FMC_NE low time	$3T_{fmc_ker_ck} - 1$	$3T_{fmc_ker_ck} + 1$	ns
$t_{v(NOE_NE)}$	FMC_NEx low to FMC_NOE low	0	0.5	
$t_{w(NOE)}$	FMC_NOE low time	$2T_{fmc_ker_ck} - 1$	$2T_{fmc_ker_ck} + 1$	
$t_{h(NE_NOE)}$	FMC_NOE high to FMC_NE high hold time	0	-	
$t_{v(A_NE)}$	FMC_NEx low to FMC_A valid	-	0.5	
$t_{h(A_NOE)}$	Address hold time after FMC_NOE high	0	-	
$t_{su(Data_NE)}$	Data to FMC_NEx high setup time	13	-	

Symbol	Parameter	Min(1)	Max(1)	Unit
$t_{su(Data_NOE)}$	Data to FMC_NOEx high setup time	11	-	ns
$t_{h(Data_NOE)}$	Data hold time after FMC_NOE high	0	-	
$t_{h(Data_NE)}$	Data hold time after FMC_NEx high	0	-	
$t_{v(NADV_NE)}$	FMC_NEx low to FMC_NADV low	-	0	
$t_{w(NADV)}$	FMC_NADV low time	-	$T_{fmc_ker_ck}$ + 1	

1. *Guaranteed by characterization results.*

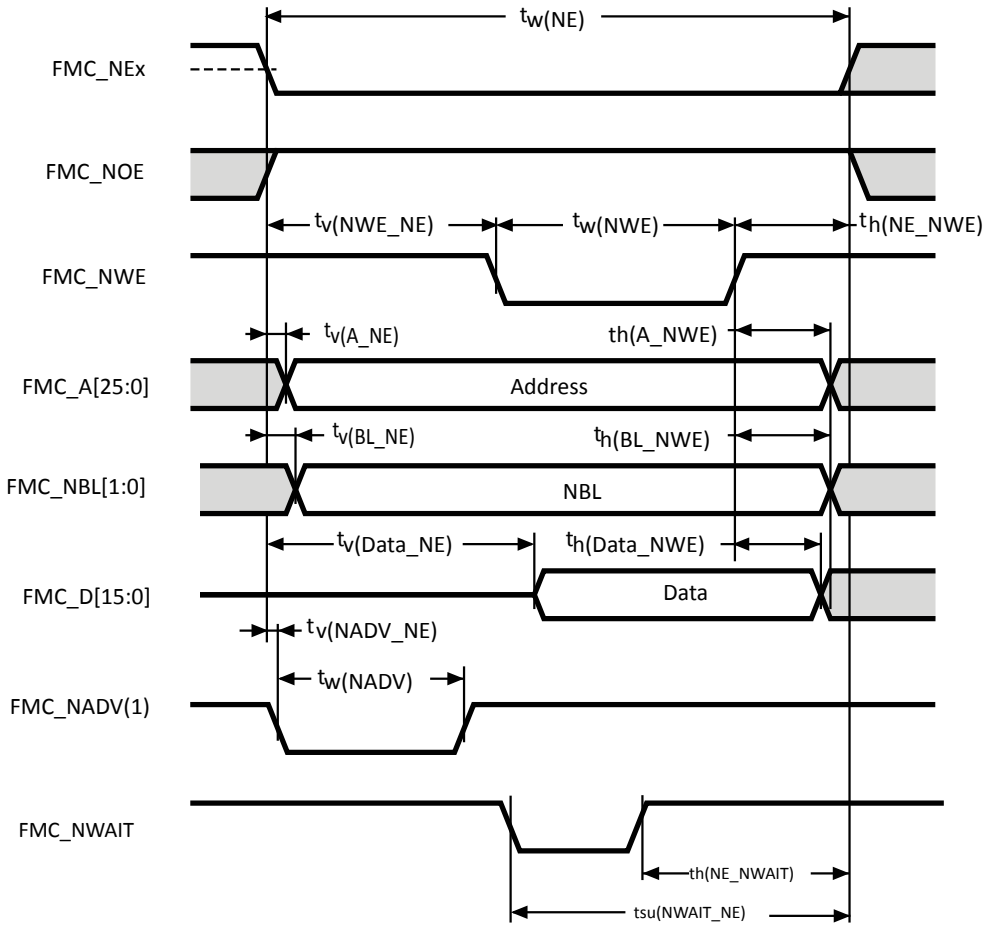
Table 71. Asynchronous non-multiplexed SRAM/PSRAM/NOR read-NWAIT timings

NWAIT pulse width is equal to 1 AHB cycle.

Symbol	Parameter	Min(1)	Max(1)	Unit
$t_{w(NE)}$	FMC_NE low time	$7T_{fmc_ker_ck}$ +1	$7T_{fmc_ker_ck}$ +1	ns
$t_{w(NOE)}$	FMC_NWE low time	$5T_{fmc_ker_ck}$ −1	$5T_{fmc_ker_ck}$ +1	
$t_{w(NWAIT)}$	FMC_NWAIT low time	$T_{fmc_ker_ck}$ −1 −0.5	-	
$t_{su(NWAIT_NE)}$	FMC_NWAIT valid before FMC_NEx high	$4T_{fmc_ker_ck}$ +9	-	
$t_{h(NE_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$3T_{fmc_ker_ck}$+12	-	

1. *Guaranteed by characterization results.*

Figure 28. Asynchronous non-multiplexed SRAM/PSRAM/NOR write waveforms



1. Mode 2/B, C and D only. In Mode 1, FMC_NADV is not used.

Table 72. Asynchronous non-multiplexed SRAM/PSRAM/NOR write timings

Symbol	Parameter	Min(1)	Max(1)	Unit
$t_{w(NE)}$	FMC_NE low time	$3T_{fmc_ker_ck} - 1$	$3T_{fmc_ker_ck} + 1$	ns
$t_{v(NWE_NE)}$	FMC_NEx low to FMC_NWE low	$T_{fmc_ker_ck} - 1$	$T_{fmc_ker_ck}$	
$t_{w(NWE)}$	FMC_NWE low time	$T_{fmc_ker_ck} - 0.5$	$T_{fmc_ker_ck} + 0.5$	
$t_{h(NE_NWE)}$	FMC_NWE high to FMC_NE high hold time	$T_{fmc_ker_ck}$	-	
$t_{v(A_NE)}$	FMC_NEx low to FMC_A valid	-	2	
$t_{h(A_NWE)}$	Address hold time after FMC_NWE high	$T_{fmc_ker_ck} + 0.5$	-	
$t_{v(BL_NE)}$	FMC_NEx low to FMC_BL valid	-	0.5	
$t_{h(BL_NWE)}$	FMC_BL hold time after FMC_NWE high	$T_{fmc_ker_ck} - 0.5$	-	
$t_{v(Data_NE)}$	Data to FMC_NEx low to Data valid	-	$T_{fmc_ker_ck} + 3$	
$t_{h(Data_NWE)}$	Data hold time after FMC_NWE high	$T_{fmc_ker_ck+1}$	-	
$t_{v(NADV_NE)}$	FMC_NEx low to FMC_NADV low	-	0	
$t_{w(NADV)}$	FMC_NADV low time	-	$T_{fmc_ker_ck} + 1$	

1. *Guaranteed by characterization results.*

Table 73. Asynchronous non-multiplexed SRAM/PSRAM/NOR write-NWAIT timings

NWAIT pulse width is equal to 1 AHB cycle.

Symbol	Parameter	Min(1)	Max(1)	Unit
$t_{w(NE)}$	FMC_NE low time	$8T_{fmc_ker_ck} - 1$	$8T_{fmc_ker_ck} + 1$	ns
$t_{w(NWE)}$	FMC_NWE low time	$6T_{fmc_ker_ck} - 1$	$6T_{fmc_ker_ck} + 1$	
$t_{su(NWAIT_NE)}$	FMC_NWAIT valid before FMC_NEx high	$5T_{fmc_ker_ck} + 13$	-	
$t_{h(NE_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$4T_{fmc_ker_ck} + 12$	-	

1. *Guaranteed by characterization results.*

Figure 29. Asynchronous multiplexed PSRAM/NOR read waveforms

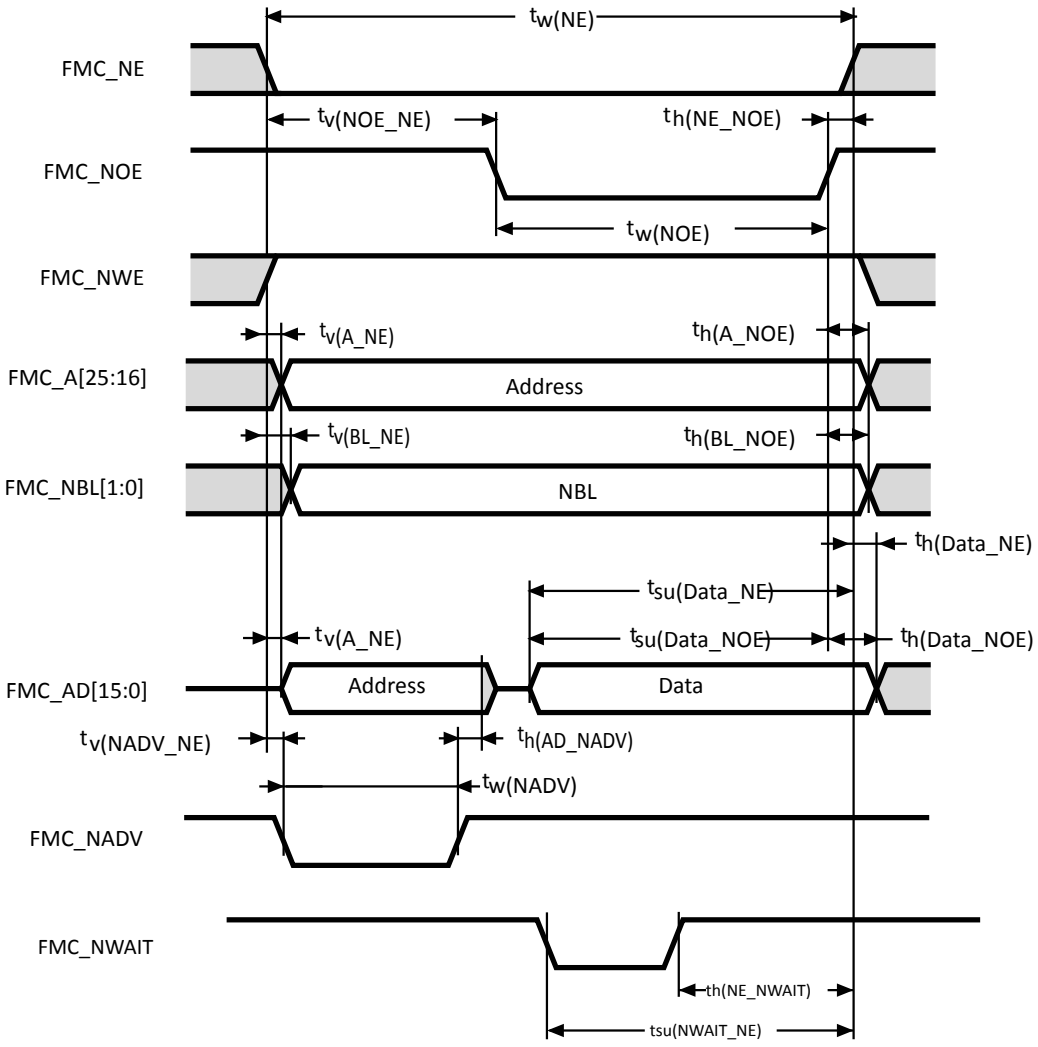


Table 74. Asynchronous multiplexed PSRAM/NOR read timings

Symbol	Parameter	Min(1)	Max(1)	Unit
$t_{w(NE)}$	FMC_NE low time	$4T_{fmc_ker_ck} - 1$	$4T_{fmc_ker_ck} + 1$	ns

Symbol	Parameter	Min(1)	Max(1)	Unit
$t_{v(NOE_NE)}$	FMC_NEx low to FMC_NOE low	$2T_{fmc_ker_ck}$	$2T_{fmc_ker_ck + 0.5}$	ns
$t_{w(NOE)}$	FMC_NOE low time	$T_{fmc_ker_ck} - 1$	$T_{fmc_ker_ck} + 1$	
$t_{h(NE_NOE)}$	FMC_NOE high to FMC_NE high hold time	0	-	
$t_{v(A_NE)}$	FMC_NEx low to FMC_A valid	-	0.5	
$t_{v(NADV_NE)}$	FMC_NEx low to FMC_NADV low	0	0.5	
$t_{w(NADV)}$	FMC_NADV low time	$T_{fmc_ker_ck} - 0.5$	$T_{fmc_ker_ck}+1$	
$t_{h(AD_NADV)}$	FMC_AD(address) valid hold time after FMC_NADV high	$T_{fmc_ker_ck} + 0.5$	-	
$t_{h(A_NOE)}$	Address hold time after FMC_NOE high	$T_{fmc_ker_ck} - 0.5$	-	
$t_{su(Data_NE)}$	Data to FMC_NEx high setup time	13	-	
$t_{su(Data_NOE)}$	Data to FMC_NOE high setup time	11	-	
$t_{h(Data_NE)}$	Data hold time after FMC_NEx high	0	-	
$t_{h(Data_NOE)}$	Data hold time after FMC_NOE high	0	-	

1. *Guaranteed by characterization results.*

Table 75. Asynchronous multiplexed PSRAM/NOR read - NWAIT timings

Symbol	Parameter	Min(1)	Max(1)	Unit
$t_{w(NE)}$	FMC_NE low time	$8T_{fmc_ker_ck} - 1$	$8T_{fmc_ker_ck} + 1$	ns
$t_{w(NOE)}$	FMC_NWE low time	$5T_{fmc_ker_ck} - 1$	$5T_{fmc_ker_ck} + 1$	
$t_{su(NWAIT_NE)}$	FMC_NWAIT valid before FMC_NEx high	$4T_{fmc_ker_ck} + 9$	-	
$t_{h(NE_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$4T_{fmc_ker_ck} + 12$	-	

1. *Guaranteed by characterization results.*

Table 76. Asynchronous multiplexed PSRAM/NOR write timings

Symbol	Parameter	Min(1)	Max(1)	Unit
$t_{w(NE)}$	FMC_NE low time	$4T_{fmc_ker_ck} - 1$	$4T_{fmc_ker_ck}$	ns
$t_{v(NWE_NE)}$	FMC_NEx low to FMC_NWE low	Tfmc_ker_c − 1	$T_{fmc_ker_ck} + 0.5$	
$t_{w(NWE)}$	FMC_NWE low time	$2T_{fmc_ker_ck}- 0.5$	$2T_{fmc_ker_ck} + 0.5$	
$t_{h(NE_NWE)}$	FMC_NWE high to FMC_NE high hold time	$T_{fmc_ker_ck} - 0.5$	-	
$t_{v(A_NE)}$	FMC_NEx low to FMC_A valid	-	0	
$t_{v(NADV_NE)}$	FMC_NEx low to FMC_NADV low	0	0.5	
$t_{w(NADV)}$	FMC_NADV low time	$T_{fmc_ker_ck}$	$T_{fmc_ker_ck}+ 1$	
$t_{h(AD_NADV)}$	FMC_AD(address) valid hold time after FMC_NADV high	$T_{fmc_ker_ck} + 0.5$	-	
$t_{h(A_NWE)}$	Address hold time after FMC_NWE high	$T_{fmc_ker_ck} + 0.5$	-	
$t_{h(BL_NWE)}$	FMC_BL hold time after FMC_NWE high	$T_{fmc_ker_ck} - 0.5$	-	
$t_{v(BL_NE)}$	FMC_NEx low to FMC_BL valid	-	0.5	
$t_{v(Data_NADV)}$	FMC_NADV high to Data valid	-	$T_{fmc_ker_ck} + 2$	
$t_{h(Data_NWE)}$	Data hold time after FMC_NWE high	$T_{fmc_ker_ck}+0.5$	-	

1. *Guaranteed by characterization results.*

Table 77. Asynchronous multiplexed PSRAM/NOR write - NWAIT timings

Symbol	Parameter	Min(1)	Max(1)	Unit
$t_{w(NE)}$	FMC_NE low time	$9T_{fmc_ker_ck} - 1$	$9T_{fmc_ker_ck}$	ns
$t_{w(NWE)}$	FMC_NWE low time	$7T_{fmc_ker_ck} - 0.5$	$7T_{fmc_ker_ck} + 0.5$	
$t_{su(NWAIT_NE)}$	FMC_NWAIT valid before FMC_NEx high	$5T_{fmc_ker_ck} + 9$	-	
$t_{h(NE_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$4T_{fmc_ker_ck} + 12$	-	

1. *Guaranteed by characterization results.*

Synchronous waveforms and timings

Figure 30 through Figure 33 represent synchronous waveforms and Table 78 through Table 81 provide the corresponding timings. The results shown in these tables are obtained with the following FMC configuration:

- BurstAccessMode = FMC_BurstAccessMode_Enable
- MemoryType = FMC_MemoryType_CRAM
- WriteBurst = FMC_WriteBurst_Enable
- CLKDivision = 1
- DataLatency = 1 for NOR flash; DataLatency = 0 for PSRAM

In all timing tables, $T_{fmc_ker_ck}$ is the kernel clock period, with the following FMC_CLK maximum values:

- For 2.7 V < V_{DD} < 3.6 V, FMC_CLK = 125 MHz at 20 pF
- For 1.8 V < V_{DD} < 1.9 V, FMC_CLK = 100 MHz at 20 pF
- For 1.62 V < V_{DD}<1.8 V, FMC_CLK = 100 MHz at 15 pF

Figure 30. Synchronous multiplexed NOR/PSRAM read timings

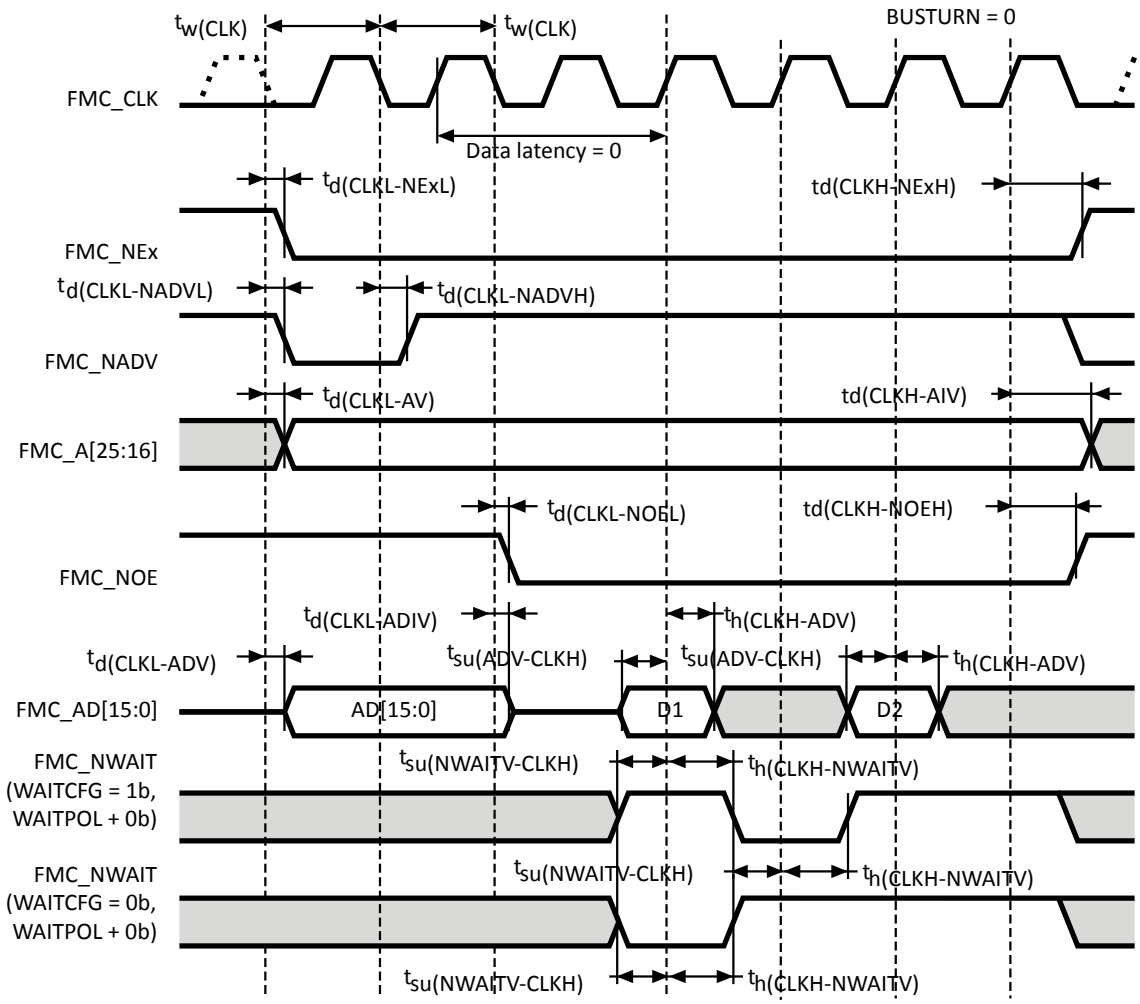


Table 78. Synchronous multiplexed NOR/PSRAM read timings

Symbol	Parameter	Min(1)	Max(1)(2)	Unit
$t_{w(CLK)}$	FMC_CLK period	$2T_{fmc_ker_ck}$ –0.5	-	ns
$t_{d(CLKL-NExL)}$	FMC_CLK low to FMC_NEx low (x=0..2)	-	2	
$t_{d(CLKH_NExH)}$	FMC_CLK high to FMC_NEx high (x= 0…2)	$T_{fmc_ker_ck}$+1.5	-	
$t_{d(CLKL-NADVL)}$	FMC_CLK low to FMC_NADV low	-	1	
$t_{d(CLKL-NADVH)}$	FMC_CLK low to FMC_NADV high	0	-	
$t_{d(CLKL-AV)}$	FMC_CLK low to FMC_Ax valid (x=16…25)	-	2.0	
$t_{d(CLKH-AIV)}$	FMC_CLK high to FMC_Ax invalid (x=16…25)	$T_{fmc_ker_ck}$+1.5	-	
$t_{d(CLKL-NOEL)}$	FMC_CLK low to FMC_NOE low	-	1.5	
$t_{d(CLKH-NOEH)}$	FMC_CLK high to FMC_NOE high	$T_{fmc_ker_ck}$ +1.5	-	
$t_{d(CLKL-ADV)}$	FMC_CLK low to FMC_AD[15:0] valid	-	3	
$t_{d(CLKL-ADIV)}$	FMC_CLK low to FMC_AD[15:0] invalid	0	-	
$t_{su(ADV-CLKH)}$	FMC_A/D[15:0] valid data before FMC_CLK high	3	-	

Symbol	Parameter	Min(1)	Max(1)(2)	Unit
$t_{h(CLKH-ADV)}$	FMC_A/D[15:0] valid data after FMC_CLK high	0.5	-	ns
$t_{su(NWAIT-CLKH)}$	FMC_NWAIT valid before FMC_CLK high	3	-	
$t_{h(CLKH-NWAIT)}$	FMC_NWAIT valid after FMC_CLK high	1	-	

1. *Guaranteed by characterization results.*
2. *At VOS1, these values are degraded by up to 5 %.*

Figure 31. Synchronous multiplexed PSRAM write timings

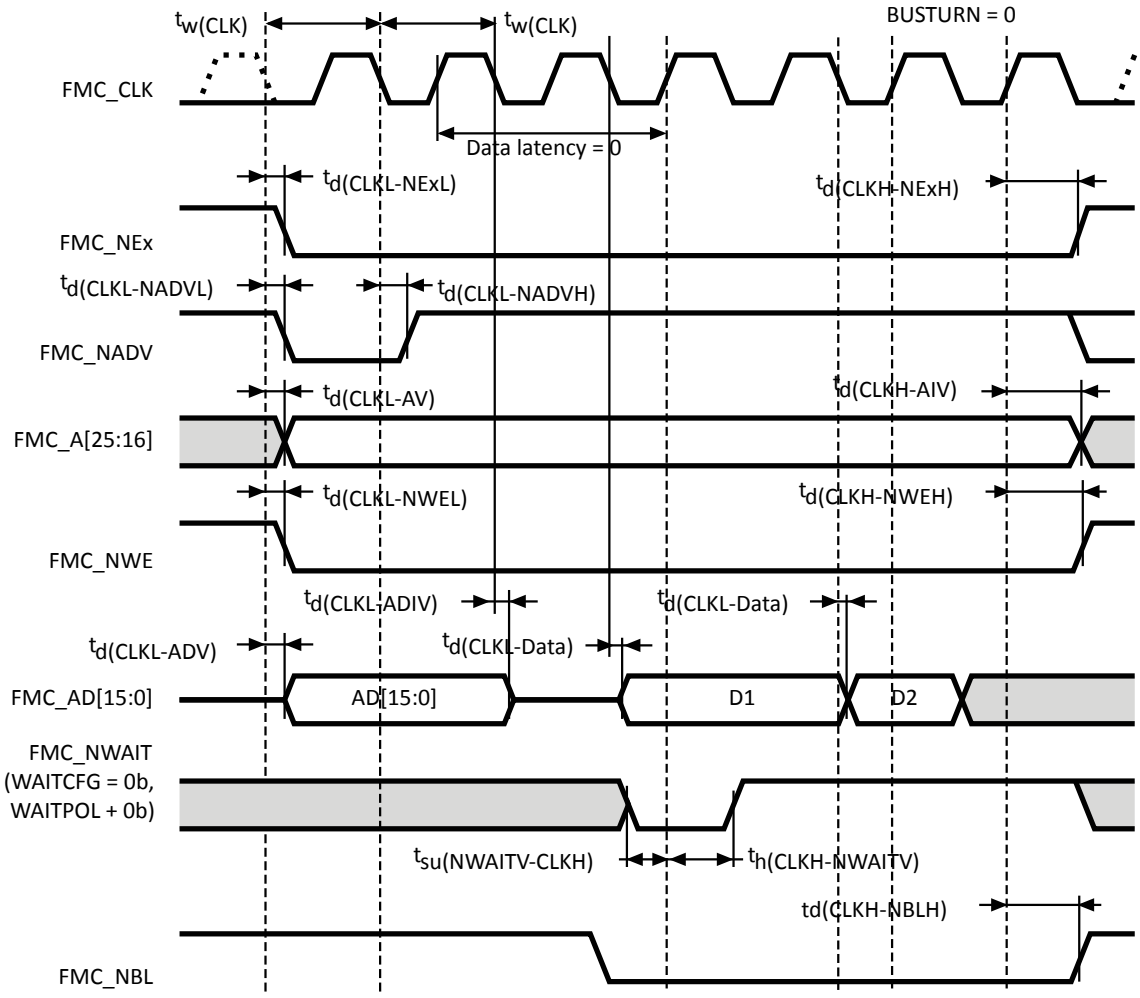


Table 79. Synchronous multiplexed PSRAM write timings

Symbol	Parameter	Min(1)	Max(1)(2)	Unit
$t_{w(CLK)}$	FMC_CLK period	$2T_{fmc_ker_ck}-1$	-	ns
$t_{d(CLKL-NExL)}$	FMC_CLK low to FMC_NEx low (x =0..2)	-	2	
$t_{d(CLKH-NExH)}$	FMC_CLK high to FMC_NEx high (x = 0…2)	$T_{fmc_ker_ck}+1.5$	-	
$t_{d(CLKL-NADVL)}$	FMC_CLK low to FMC_NADV low	-	1.5	

Symbol	Parameter	Min(1)	Max(1)(2)	Unit
$t_{d(CLKL-NADVH)}$	FMC_CLK low to FMC_NADV high	0	-	ns
$t_{d(CLKL-AV)}$	FMC_CLK low to FMC_Ax valid (x =16…25)	-	2	
$t_{d(CLKH-AIV)}$	FMC_CLK high to FMC_Ax invalid (x =16…25)	$T_{fmc_ker_ck}$ +1.5	-	
$t_{d(CLKL-NWEL)}$	FMC_CLK low to FMC_NWE low	-	1.5	
$t_{(CLKH-NWEH)}$	FMC_CLK high to FMC_NWE high	$T_{fmc_ker_ck}$ +1	-	
$t_{d(CLKL-ADV)}$	FMC_CLK low to to FMC_AD[15:0] valid	-	2.5	
$t_{d(CLKL-ADIV)}$	FMC_CLK low to FMC_AD[15:0] invalid	0	-	
$t_{d(CLKL-DATA)}$	FMC_A/D[15:0] valid data after FMC_CLK low	-	3	
$t_{d(CLKL-NBLL)}$	FMC_CLK low to FMC_NBL low	-	2	
$t_{d(CLKH-NBLH)}$	FMC_CLK high to FMC_NBL high	$T_{fmc_ker_ck}$ +0.5	-	
$t_{su(NWAIT-CLKH)}$	FMC_NWAIT valid before FMC_CLK high	2	-	
$t_{h(CLKH-NWAIT)}$	FMC_NWAIT valid after FMC_CLK high	2	-	

1. *Guaranteed by characterization results.*
2. *At VOS1, these values are degraded by up to 5 %.*

Figure 32. Synchronous non-multiplexed NOR/PSRAM read timings

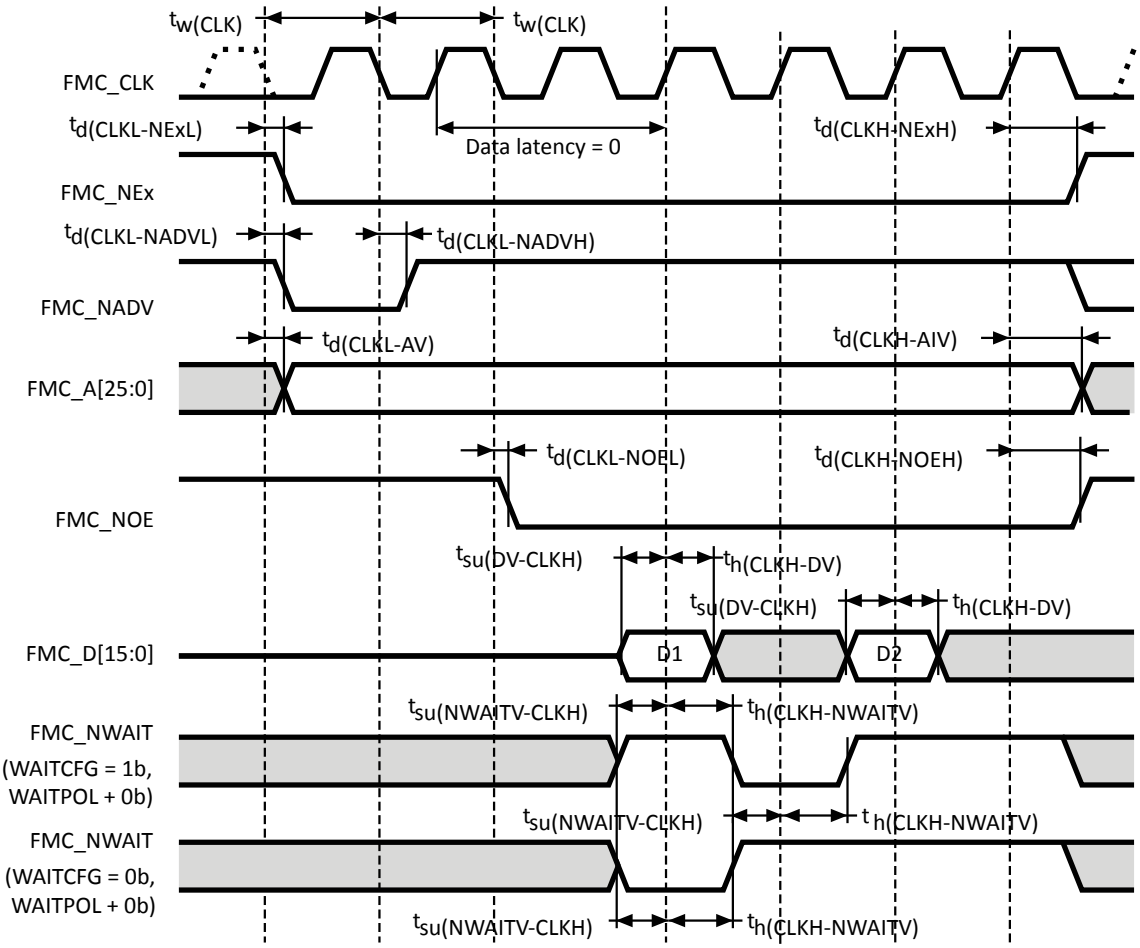


Table 80. Synchronous non-multiplexed NOR/PSRAM read timings

Symbol	Parameter	Min(1)	Max(1)(2)	Unit
$t_{w(CLK)}$	FMC_CLK period	$2T_{fmc_ker_ck}$ –0.5	-	ns
$t_{(CLKL-NExL)}$	FMC_CLK low to FMC_NEx low (x=0..2)	-	2	
$t_{d(CLKH-NExH)}$	FMC_CLK high to FMC_NEx high (x= 0…2)	$T_{fmc_ker_ck}$+1.5	-	
$t_{d(CLKL-NADVL)}$	FMC_CLK low to FMC_NADV low	-	1.5	
$t_{d(CLKL-NADVH)}$	FMC_CLK low to FMC_NADV high	0	-	
$t_{d(CLKL-AV)}$	FMC_CLK low to FMC_Ax valid (x=16…25)	-	2	
$t_{d(CLKH-AIV)}$	FMC_CLK high to FMC_Ax invalid (x=16…25)	$T_{fmc_ker_ck}$+1.5	-	
$t_{d(CLKL-NOEL)}$	FMC_CLK low to FMC_NOE low	-	1.5	
$t_{d(CLKH-NOEH)}$	FMC_CLK high to FMC_NOE high	$T_{fmc_ker_ck}$+1	-	
$t_{su(DV-CLKH)}$	FMC_D[15:0] valid data before FMC_CLK high	3	-	
$t_{h(CLKH-DV)}$	FMC_D[15:0] valid data after FMC_CLK high	0.5	-	
$t_{(NWAIT-CLKH)}$	FMC_NWAIT valid before FMC_CLK high	3	-	

Symbol	Parameter	Min(1)	Max(1)(2)	Unit
$t_{h(CLKH-NWAIT)}$	FMC_NWAIT valid after FMC_CLK high	1	-	ns

1. *Guaranteed by characterization results.*
2. *At VOS1, these values are degraded by up to 5 %.*

Figure 33. Synchronous non-multiplexed PSRAM write timings

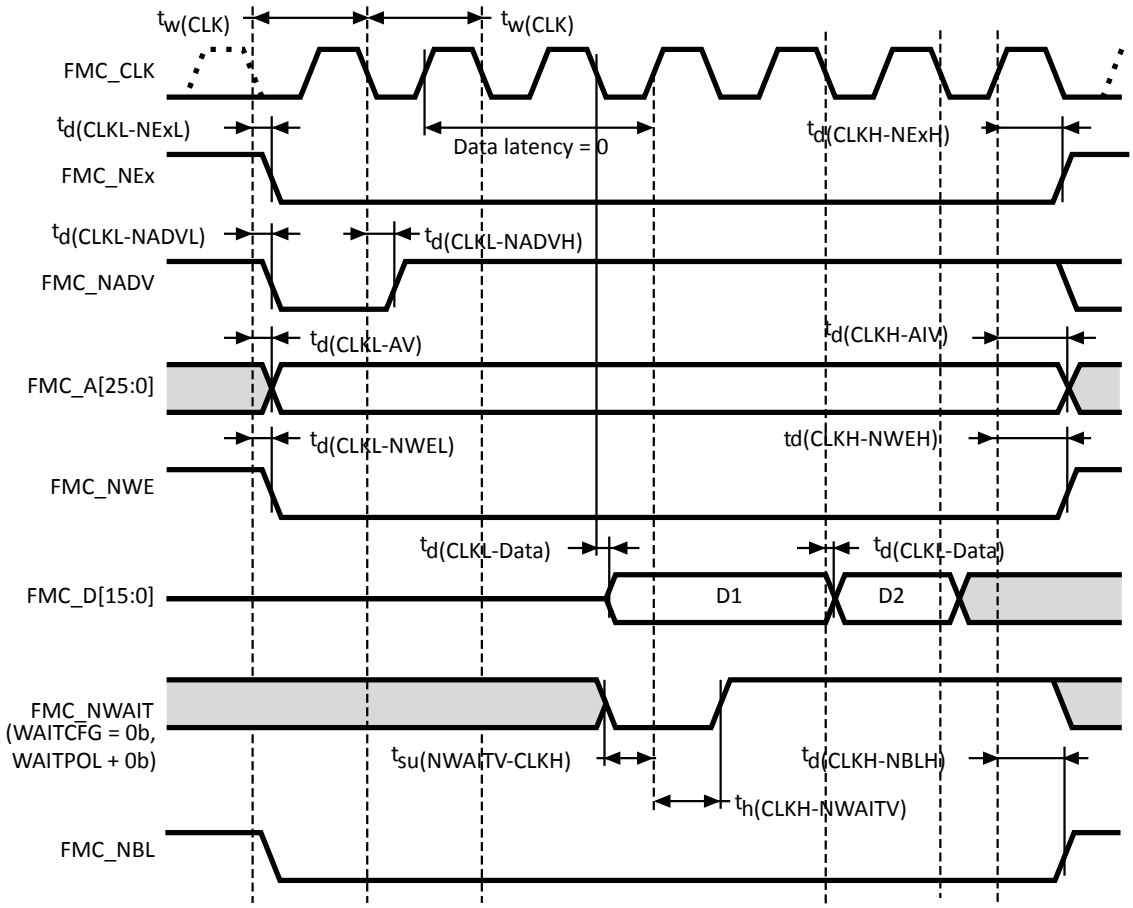


Table 81. Synchronous non-multiplexed PSRAM write timings

Symbol	Parameter	Min(1)	Max(1)(2)	Unit
$t_{(CLK)}$	FMC_CLK period	$2T_{fmc_ker_ck} - 0.5$	-	ns
$t_{d(CLKL-NExL)}$	FMC_CLK low to FMC_NEx low (x=0..2)	-	2	
$t_{(CLKH-NExH)}$	FMC_CLK high to FMC_NEx high (x= 0…2)	$T_{fmc_ker_ck} + 1.5$	-	
$t_{d(CLKL-NADVL)}$	FMC_CLK low to FMC_NADV low	-	1.5	
$t_{d(CLKL-NADVH)}$	FMC_CLK low to FMC_NADV high	0	-	
$t_{d(CLKL-AV)}$	FMC_CLK low to FMC_Ax valid (x=16…25)	-	2	
$t_{d(CLKH-AIV)}$	FMC_CLK high to FMC_Ax invalid (x=16…25)	$T_{fmc_ker_ck} + 1.5$	-	
$t_{d(CLKL-NWEL)}$	FMC_CLK low to FMC_NWE low	-	1.5	
$t_{d(CLKH-NWEH)}$	FMC_CLK high to FMC_NWE high	$T_{fmc_ker_ck} + 1$	-	

Symbol	Parameter	Min(1)	Max(1)(2)	Unit
$t_{d(CLKL-Data)}$	FMC_D[15:0] valid data after FMC_CLK low	-	3	ns
$t_{d(CLKL-NBLL)}$	FMC_CLK low to FMC_NBL low	-	2	
$t_{d(CLKH-NBLH)}$	FMC_CLK high to FMC_NBL high	$T_{fmc_ker_ck}$ + 0.5	-	
$t_{su(NWAIT-CLKH)}$	FMC_NWAIT valid before FMC_CLK high	2	-	
$t_{h(CLKH-NWAIT)}$	FMC_NWAIT valid after FMC_CLK high	2	-	

1. *Guaranteed by characterization results.*
2. *At VOS1, these values are degraded by up to 5 %.*

NAND controller waveforms and timings

Figure 34 through Figure 37 represent synchronous waveforms, and Table 82 and Table 83 provide the corresponding timings. The results shown in this table are obtained with the following FMC configuration:

- COM.FMC_SetupTime = 0x01
- COM.FMC_WaitSetupTime = 0x03
- COM.FMC_HoldSetupTime = 0x02
- COM.FMC_HiZSetupTime = 0x01
- ATT.FMC_SetupTime = 0x01
- ATT.FMC_WaitSetupTime = 0x03
- ATT.FMC_HoldSetupTime = 0x02
- ATT.FMC_HiZSetupTime = 0x01
- Bank = FMC_Bank_NAND
- MemoryDataWidth = FMC_MemoryDataWidth_16b
- ECC = FMC_ECC_Enable
- ECCPageSize = FMC_ECCPageSize_512Bytes
- TCLRSetupTime = 0
- TARSetupTime = 0
- Capacitive load C_L = 30 pF

In all timing tables, $T_{fmc_ker_ck}$ is the kernel clock period.

Figure 34. NAND controller waveforms for read access

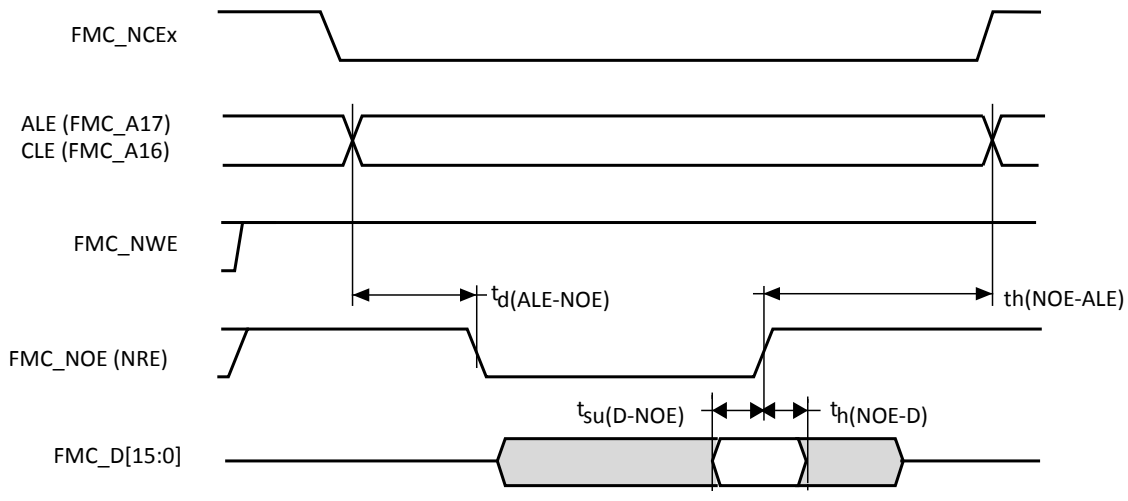


Figure 35. **NAND controller waveforms for write access**

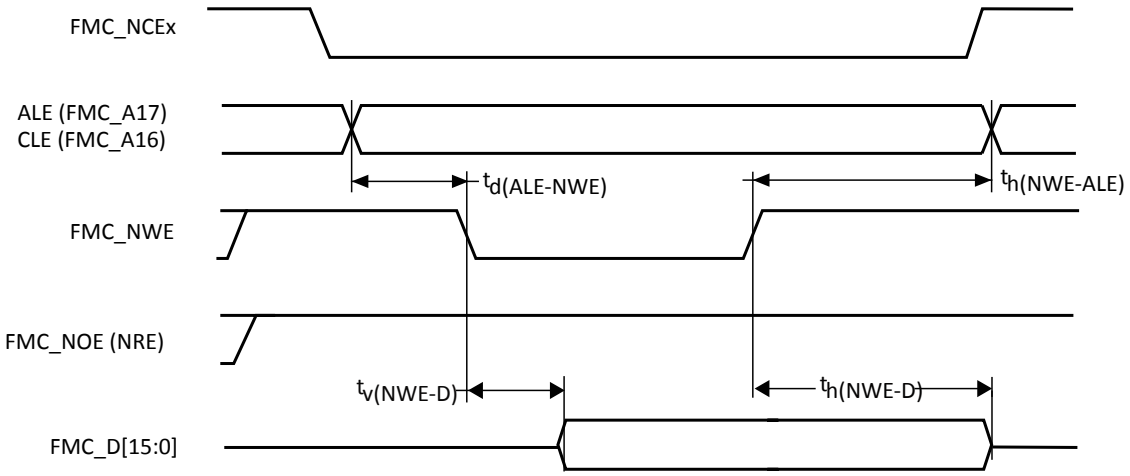


Figure 36. **NAND controller waveforms for common memory read access**

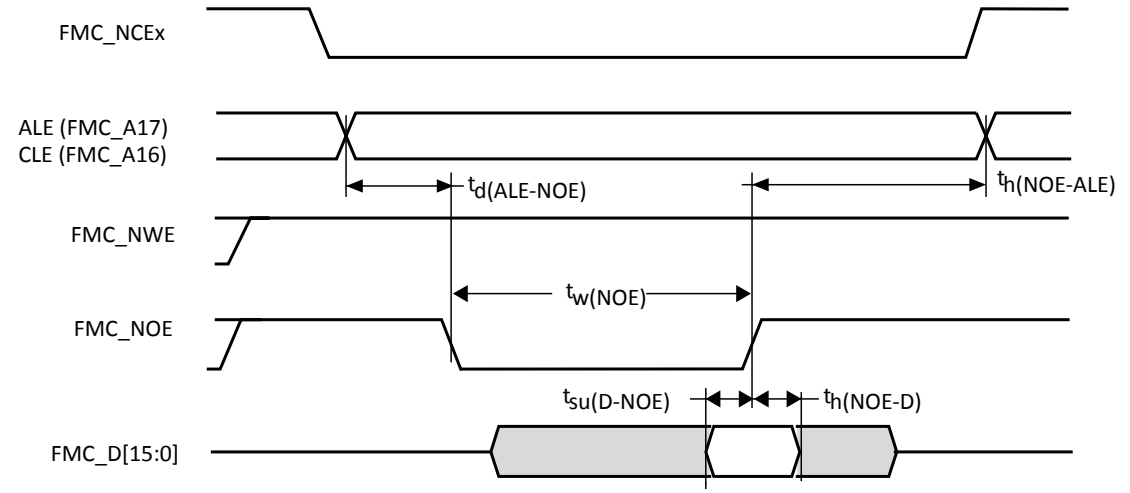


Figure 37. **NAND controller waveforms for common memory write access**

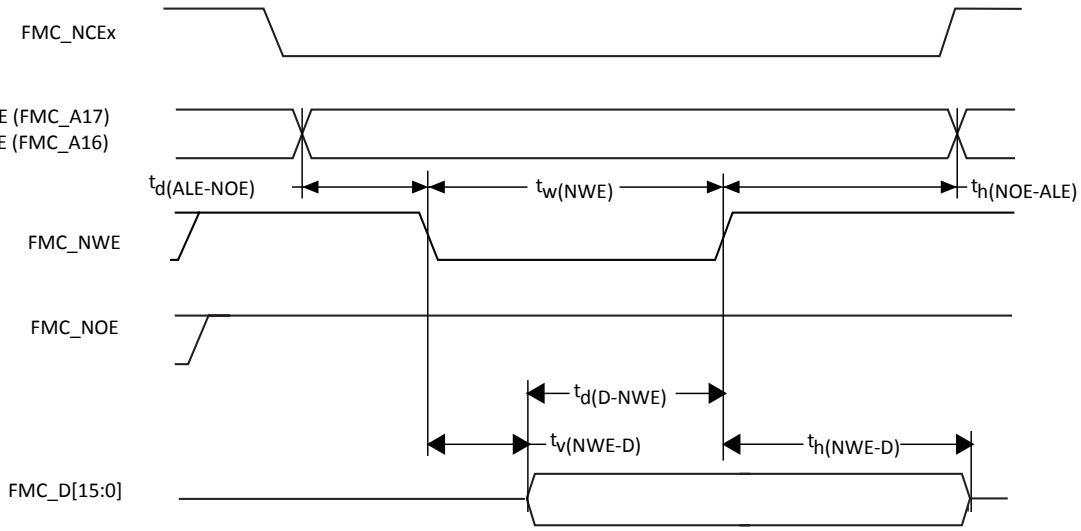


Table 82. **Switching characteristics for NAND flash memory read cycles**

Symbol	Parameter	Min(1)	Max(1)	Unit
$t_{w(N0E)}$	FMC_NOE low width	$4T_{fmc_ker_ck} - 0.5$	$4T_{fmc_ker_ck}+0.5$	ns
$t_{su(D-NOE)}$	FMC_D[15-0] valid data before FMC_NOE high	8	-	
$t_{h(NOE-D)}$	FMC_D[15-0] valid data after FMC_NOE high	0	-	
$t_{d(ALE-NOE)}$	FMC_ALE valid before FMC_NOE low	-	$3T_{fmc_ker_ck}$ +0.5	
$t_{h(NOE-ALE)}$	FMC_NWE high to FMC_ALE invalid	$4T_{fmc_ker_ck}$ –1	-	

1. *Guaranteed by characterization results.*

Table 83. **Switching characteristics for NAND flash write cycles**

Symbol	Parameter	Min(1)	Max(1)	Unit
$t_{w(NWE)}$	FMC_NWE low width	$4T_{fmc_ker_ck} - 0.5$	$4T_{fmc_ker_ck}$ +0.5	ns
$t_{v(NWE-D)}$	FMC_NWE low to FMC_D[15-0] valid	0	-	
$t_{h(NWE-D)}$	FMC_NWE high to FMC_D[15-0] invalid	$2T_{fmc_ker_ck}$ + 1.5	-	
$t_{d(D-NWE)}$	FMC_D[15-0] valid before FMC_NWE high	$5T_{fmc_ker_ck} - 2$	-	
$t_{d(ALE-NWE)}$	FMC_ALE valid before FMC_NWE low	-	$3T_{fmc_ker_ck}$ +0.5	
$t_{h(NWE-ALE)}$	FMC_NWE high to FMC_ALE invalid	$2T_{fmc_ker_ck}$ + 0.5	-	

1. *Guaranteed by characterization results.*

SDRAM waveforms and timings

In all timing tables, $T_{fmc_ker_ck}$ is the kernel clock period, with the following FMC_SDCLK maximum values:

- For 2.7 V < V_{DD} <3.6 V: FMC_CLK =110 MHz at 20 pF
- For 1.8 V < V_{DD} <1.9 V: FMC_CLK =100 MHz at 20 pF
- For 1.62 V <$_{DD}$ <1.8 V, FMC_CLK =100 MHz at 15 pF

Figure 38. **SDRAM read access waveforms (CL = 1)**

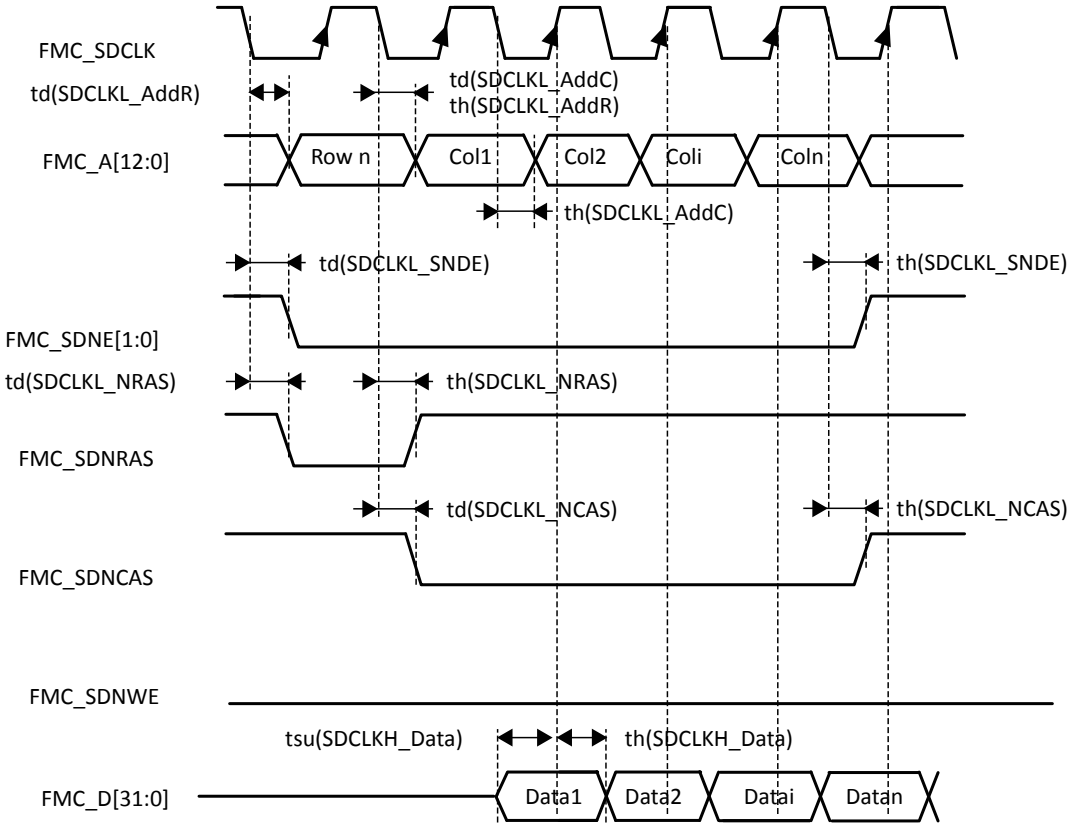


Table 84. **SDRAM read timings**

Symbol	Parameter	Min(1)	Max(1)(2)	Unit
$t_{w(SDCLK)}$	FMC_SDCLK period	$2T_{fmc_ker_ck} - 1$	$2T_{fmc_ker_ck} + 0.5$	ns
$t_{su(SDCLKH_Data)}$	Data input setup time	3	-	
$t_{h(SDCLKH_Data)}$	Data input hold time	0	-	
$t_{d(SDCLKL_Add)}$	Address valid time	-	1.5	
$t_{d(SDCLKL-SDNE)}$	Chip select valid time	-	2(3)	
$t_{h(SDCLKL_SDNE)}$	Chip select hold time	0.5	-	
$t_{d(SDCLKL_SDNRAS)}$	SDNRAS valid time	-	2	
$t_{h(SDCLKL_SDNRAS)}$	SDNRAS hold time	0.5	-	
$t_{d(SDCLKL_SDNCAS)}$	SDNCAS valid time	-	0.5	
$t_{h(SDCLKL_SDNCAS)}$	SDNCAS hold time	0	-	

1. *Guaranteed by characterization results.*
2. *At VOS1, these values are degraded by up to 5 %.*
3. *Using PC2_C I/O adds 4.5 ns to this timing.*

Table 85. LPSDR SDRAM read timings

Symbol	Parameter	Min(1)	Max(1)(2)	Unit
$t_{W(SDCLK)}$	FMC_SDCLK period	$2T_{fmc_ker_ck} - 1$	$2T_{fmc_ker_ck}+0.5$	ns
$t_{su(SDCLKH_Data)}$	Data input setup time	3	-	
$t_{h(SDCLKH_Data)}$	Data input hold time	0.5	-	
$t_{d(SDCLKL_Add)}$	Address valid time	-	3.5	
$t_{d(SDCLKL_SDNE)}$	Chip select valid time	-	2.5(3)	
$t_{h(SDCLKL_SDNE)}$	Chip select hold time	0	-	
$t_{d(SDCLKL_SDNRAS}$	SDNRAS valid time	-	1	
$t_{h(SDCLKL_SDNRAS)}$	SDNRAS hold time	0	-	
$t_{d(SDCLKL_SDNCAS)}$	SDNCAS valid time	-	1.5	
$t_{h(SDCLKL_SDNCAS)}$	SDNCAS hold time	0	-	

1. *Guaranteed by characterization results.*
2. *At VOS1, these values are degraded by up to 5 %.*
3. *Using PC2_C I/O adds 16.5 ns to this timing.*

Figure 39. SDRAM write access waveforms

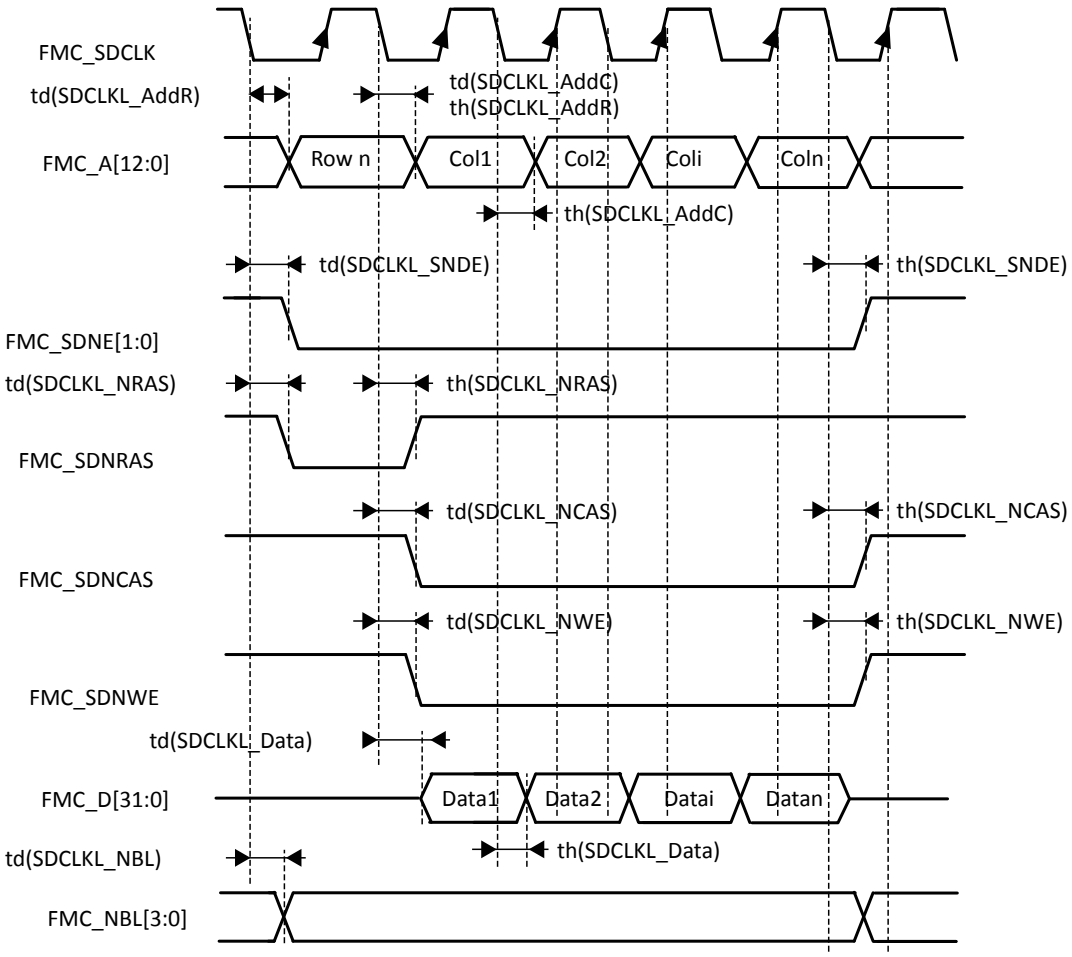


Table 86. SDRAM write timings

Symbol	Parameter	Min(1)	Max(1)(2)	Unit
$t_{w(SDCLK)}$	FMC_SDCLK period	$2T_{fmc_ker_ck} - 1$	$2T_{fmc_ker_ck}+0.5$	ns
$t_{d(SDCLKL_Data)}$	Data output valid time	-	2.5	
$t_{h(SDCLKL_Data)}$	Data output hold time	0	-	
$t_{d(SDCLKL_Add)}$	Address valid time	-	2	
$t_{d(SDCLKL_SDNWE)}$	SDNWE valid time	-	2.5	
$t_{h(SDCLKL_SDNWE)}$	SDNWE hold time	0.5	-	
$t_{d(SDCLKL_SDNE)}$	Chip select valid time	-	2(3)	
$t_{h(SDCLKL_SDNE)}$	Chip select hold time	0.5	-	
$t_{d(SDCLKL_SDNRAS)}$	SDNRAS valid time	-	1.5	
$t_{h(SDCLKL_SDNRAS)}$	SDNRAS hold time	0.5	-	
$t_{d(SDCLKL_SDNCAS)}$	SDNCAS valid time	-	1.5	
$t_{d(SDCLKL_SDNCAS)}$	SDNCAS hold time	0.5	-	

1. *Guaranteed by characterization results.*
2. *At VOS1, these values are degraded by up to 5 %.*
3. *Using PC2_C I/O adds 4.5 ns to this timing.*

Table 87. LPSDR SDRAM write timings

Symbol	Parameter	Min(1)	Max(1)(2)	Unit
$t_{w(SDCLK)}$	FMC_SDCLK period	$2T_{fmc_ker_ck} - 1$	$2T_{fmc_ker_ck}+0.5$	ns
$t_{d(SDCLKL_Data)}$	Data output valid time	-	2.5	
$t_{h(SDCLKL_Data)}$	Data output hold time	0	-	
$t_{d(SDCLKL_Add)}$	Address valid time	-	2.5	
$t_{d(SDCLKL-SDNWE)}$	SDNWE valid time	-	3	
$t_{h(SDCLKL-SDNWE)}$	SDNWE hold time	0	-	
$t_{d(SDCLKL-SDNE)}$	Chip select valid time	-	3(3)	
$t_{h(SDCLKL-SDNE)}$	Chip select hold time	0	-	
$t_{d(SDCLKL-SDNRAS)}$	SDNRAS valid time	-	2	
$t_{h(SDCLKL-SDNRAS)}$	SDNRAS hold time	0	-	
$t_{d(SDCLKL-SDNCAS)}$	SDNCAS valid time	-	2	
$t_{d(SDCLKL-SDNCAS)}$	SDNCAS hold time	0	-	

1. *Guaranteed by characterization results.*
2. *At VOS1, these values are degraded by up to 5 %.*
3. *Using PC2_C I/O adds 16.5 ns to this timing.*

6.3.19 Octo-SPI interface characteristics

Unless otherwise specified, the parameters given in Table 88. OCTOSPI characteristics in SDR mode and Table 89. OCTOSPI characteristics in DTR mode (with DQS)/Octal and Hyperbus for the OCTOSPI interface are derived from tests performed under the ambient temperature, f_{HCLK} frequency and V_{DD} supply voltage summarized in Table 20. General operating conditions, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 11
- Measurement points are done at CMOS levels: $0.5V_{DD}$
- I/O compensation cell activated.
- HSLV activated when $V_{DD} \leq 2.7$ V
- VOS level set to VOS0

Note: At VOS1, the performance can be degraded by up to 5 % compared to VOS0. This is indicated by a footnote when applicable.

Refer to Section 6.3.16 I/O port characteristics for more details on the input/output alternate function characteristics.

Table 88. OCTOSPI characteristics in SDR mode

Delay block bypassed.

Symbol	Parameter	Conditions	Min(1)(2)	Typ(1)(2)	Max(1)(2)(3)	Unit
$F_{(CLK)}$	OCTOSPI clock frequency	1.62 V < V_{DD} < 3.6 V, VOS0, C_{LOAD} = 15 pF	-	-	90	MHz
		1.62 V < V_{DD} < 3.6 V, VOS0, C_{LOAD} = 20 pF	-	-	80	
		2.7 V < V_{DD} < 3.6 V, VOS0, C_{LOAD} = 20 pF	-	-	140	
$t_{w(CLKH)}$	OCTOSPI clock high and low time	PRESCALER[7:0] = n = 0,1,3,5	$t_{(CLK)}/2$	-	$t_{(CLK)}/2+1$	ns
$t_{w(CLKL)}$			$t_{(CLK)}/2-1$	-	$t_{(CLK)}/2$	
$t_{w(CLKH)}$	OCTOSPI clock high and low time	PRESCALER[7:0] = n = 2,4,6,8	$(n/2)*t_{(CLK)}/(n+1)$	-	$(n/2)*t_{(CLK)}/(n+1)+1$	
$t_{w(CLKL)}$			$(n/2+1)*t_{(CLK)}/(n+1)-1$	-	$(n/2+1)*t_{(CLK)}/(n+1)$	
$t_{s(IN)}$(4)	Data input setup time	2.7 V < V_{DD} < 3.6 V	2	-	-	
		1.62 V < V_{DD} < 3.6 V	2	-	-	
$t_{h(IN)}$(4)	Data input hold time	2.7 V < V_{DD} < 3.6 V	4.5	-	-	
		1.62 V < V_{DD} < 3.6 V	5	-	-	
$t_{v(OUT)}$	Data output valid time	-	-	1	1.5(4)	
$t_{h(OUT)}$	Data output hold time	-	0	-	-	

1. *All values apply to Octal and Quad-SPI mode.*
2. *Guaranteed by characterization results.*
3. *At VOS1, these values are degraded by up to 5 %.*
4. *Using PC2, PC3, PI11, PF0 or PF1 I/O in the data bus adds 3.5 ns to this timing value.*

Figure 40. **OctoSPI timing diagram - SDR mode**

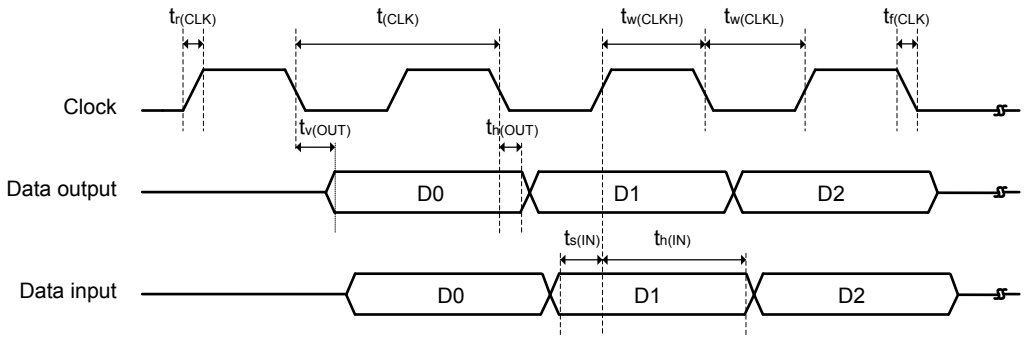


Table 89. **OCTOSPI characteristics in DTR mode (with DQS)/Octal and Hyperbus**

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)(2)	Unit
$F_{(CLK)}$(3)(4)	OCTOSPI clock frequency	1.71 V < V_{DD} < 3.6 V, VOS0, C_{LOAD} = 15 pF	-	-	120(5)	MHz
		2.7 V < V_{DD} < 3.6 V, VOS0, C_{LOAD} = 20 pF	-	-	100	
		1.62 V < V_{DD} < 2.5 V, VOS0, C_{LOAD} = 20 pF	-	-	100/45(6)	
$t_{w(CLKH)}$	OCTOSPI clock high and low time	PRESCALER[7:0] = n = 0,1,3,5	$t_{(CLK)}/2$	-	$t_{(CLK)}/2+1$	ns
$t_{w(CLKL)}$			$t_{(CLK)}/2-1$	-	$t_{(CLK)}/2$	
$t_{w(CLKH)}$	OCTOSPI clock high and low time	PRESCALER[7:0] = n = 2,4,6,8	$(n/2)*t_{(CLK)}/(n+1)$	-	$(n/2)*t_{(CLK)}/(n+1)+1$	
$t_{w(CLKL)}$			$(n/2+1)*t_{(CLK)}/(n+1)-1$	-	$(n/2+1)*t_{(CLK)}/(n+1)$	
$t_{v(CLK)}$	Clock valid time	-	-	-	$t_{(CLK)}+1$	
$t_{h(CLK)}$	Clock hold time	-	$t_{(CLK)}/2-0.5$ $t_{(CLK)}/2$	-	-	
$t_{w(CS)}$	Chip select high time	-	3 x $t_{(CLK)}$	-	-	
$t_{v(DQ)}$	Data input vallid time	-	0	-	-	
$t_{v(DS)}$	Data strobe input valid time	-	0	-	-	
$t_{h(DS)}$	Data strobe input hold time	-	0	-	-	
$t_{v(RWDS)}$	Data strobe output valid time	-	-	-	3 x $t_{(CLK)}$	
$t_{sr(DQ)}$ $t_{sf(DQ)}$(7)	Data input setup time	-	−1	-	-	
$t_{hr(DQ)}$	Data input hold time	Rising edge	3	-	-	
$t_{hf(DQ)}$(7)		Falling edge	3.5	-	-	
$t_{vr(OUT)}$ $t_{vf(OUT)}$	Data output valid time	DHQC = 0	-	5.5	7(8)	
		DHQC = 1, PRESCALER[7:0]=1,2...	-	$t_{(CLK)}/4+1$	$t_{(CLK)}/4+2$(8)	
$t_{hr(OUT)}$ $t_{hf(OUT)}$	Data output hold time	DHQC = 0	4.5	-	-	

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)(2)	Unit
$t_{hr(OUT)}$ $t_{hf(OUT)}$	Data output hold time	DHQC = 1, PRESCALER[7:0]=1,2...	$t_{(CLK)}/4$	-	-	ns

1. *Guaranteed by characterization results, unless otherwise specified.*
2. *At VOS1, these values are degraded by up to 5 %.*
3. *The maximum frequency values are given for a maximum RWDS to DQ skew ≤ ±1.0 ns.*
4. *DHQC must be set to reach the mentioned frequency.*
5. *Guaranteed by design.*
6. *Using PC2, PC3, PI11, PF0 or PF1 I/Os limits the maximum clock frequency.*
7. *Delay block bypassed.*
8. *Using PC2, PC3, PI11, PF0 or PF1 I/O in the data bus adds 3.5 ns to this timing value.*

Figure 41. OctoSPI timing diagram - DTR mode

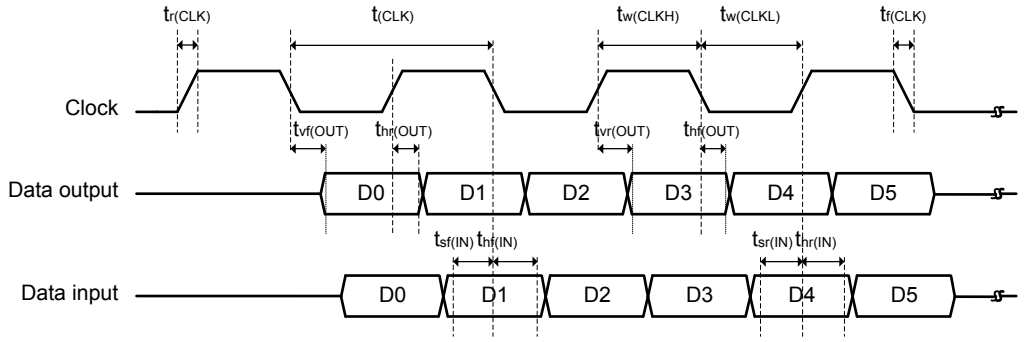


Figure 42. OctoSPI Hyperbus clock

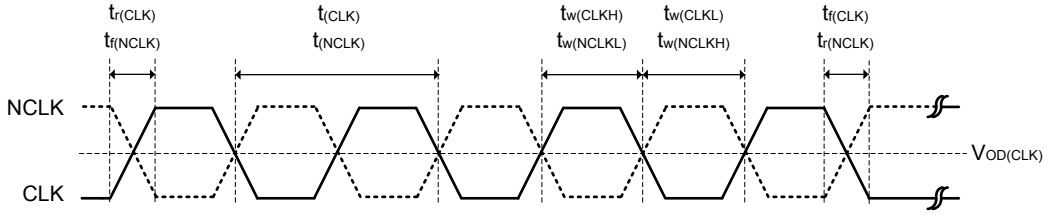


Figure 43. OctoSPI Hyperbus read

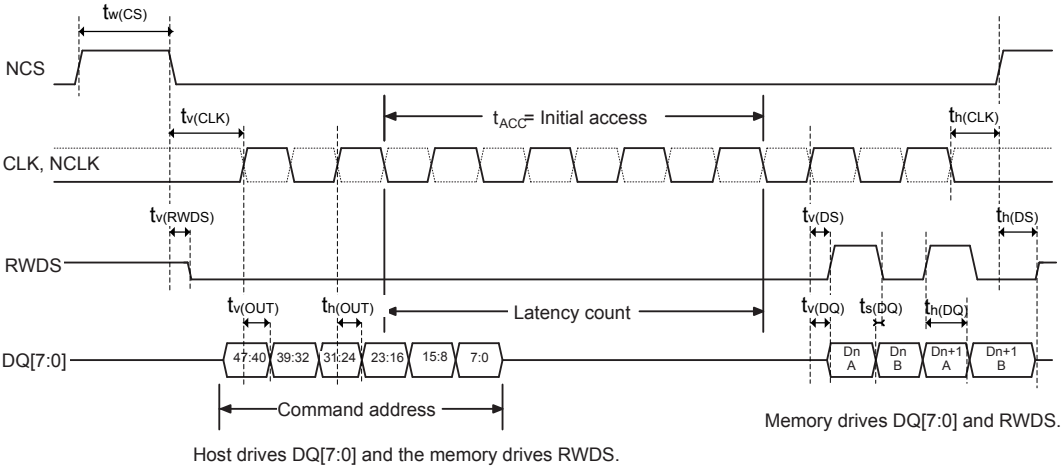
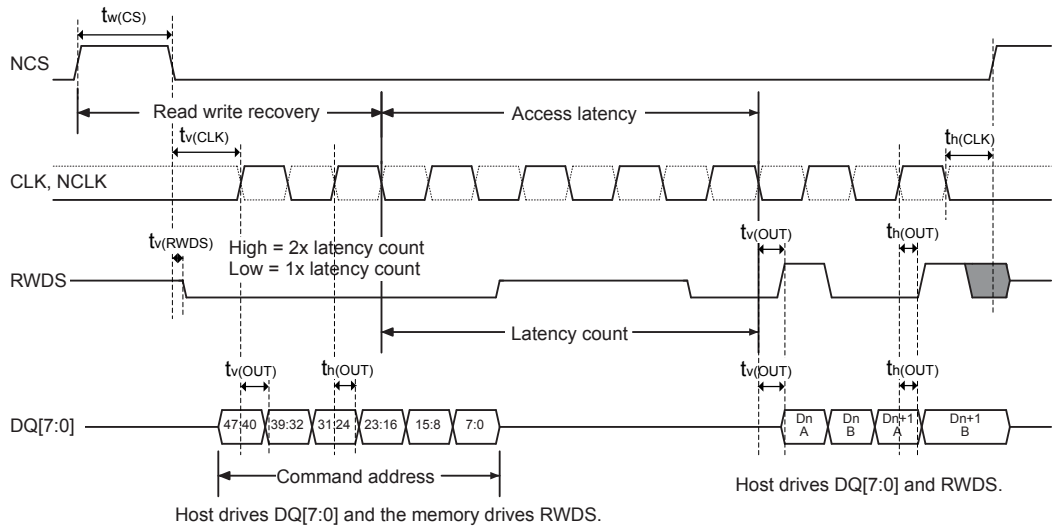


Figure 44. **OctoSPI Hyperbus write**



6.3.20 Delay block (DLYB) characteristics

Unless otherwise specified, the parameters given in Table 90. Delay Block characteristics for Delay Block are derived from tests performed under the ambient temperature, $f_{rcc_cpu_ck}$ frequency and VDD supply voltage summarized in Table 20. General operating conditions, with the following configuration:

Table 90. **Delay Block characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
t_{init}	Initial delay	-	1400	1700	2700	ps
t_{Δ}	Unit Delay	-	40	47	59	

6.3.21 16-bit ADC characteristics

Unless otherwise specified, the parameters given in Table 91. ADC characteristics are derived from tests performed under the ambient temperature, f_{PCLK2} frequency and V_{DDA} supply voltage conditions summarized in Table 20. General operating conditions.

Table 91. **ADC characteristics**

Symbol	Parameter	Conditions		Min(1)	Typ(1)	Max(1)	Unit
V_{DDA}	Analog power supply for ADC ON	-		1.62	-	3.6	V
V_{REF+}(2)	Positive reference voltage	$V_{DDA} \geq 2$ V		1.62	-	V_{DDA}	V
		$V_{DDA} < 2$ V		V_{DDA}			
V_{REF-}(2)	Negative reference voltage	-		V_{SSA}			
f_{ADC}	ADC clock frequency	1,62 V ≤V_{DDA} ≤ 3.6 V	BOOST = 11	0.12	-	50	MHz
			BOOST = 10	0.12	-	25	
			BOOST = 01	0.12	-	12.5	
			BOOST = 00	-	-	6.25	

Symbol	Parameter	Conditions				Min(1)	Typ(1)	Max(1)	Unit
f_S(3)	Sampling rate for Direct channels	Resolution = 16 bits, V_{DDA} > 2.5 V	T_J = 90 °C	f_{ADC} = 36 MHz	SMP = 1.5	-	-	3.60	MSPS
		Resolution = 16 bits		f_{ADC} = 37 MHz	SMP = 2.5	-	-	3.35	
		Resolution = 14 bits	T_J = 125 °C	f_{ADC} = 50 MHz	SMP = 2.5	-	-	5.00	
		Resolution = 12 bits		f_{ADC}=50 MHz	SMP = 2.5	-	-	5.50	
		Resolution = 10 bits		f_{ADC}=50 MHz	SMP=1.5	-	-	7.10	
		Resolution = 8 bits		f_{ADC}=50 MHz	SMP=1.5	-	-	8.30	
	Sampling rate for Fast channels	Resolution = 16 bits, V_{DDA}>2.5V	T_J = 90 °C	f_{ADC}=32 MHz	SMP=2.5	-	-	2.90	
		Resolution = 16 bits		f_{ADC}=31 MHz	SMP=2.5	-	-	2.80	
		Resolution = 14 bits	TJ = 125°C	f_{ADC}=33 MHz	SMP=2.5	-	-	3.30	
		Resolution = 12 bits		f_{ADC}=39 MHz	SMP=2.5	-	-	4.30	
		Resolution = 10 bits		f_{ADC}=48 MHz	SMP=2.5	-	-	6.00	
		Resolution = 8 bits		f_{ADC}=50 MHz	SMP=2.5	-	-	7.10	
	Sampling rate for Slow channels, BOOST = 00, f_{ADC} = 10 MHz	Resolution = 16 bits	T_J = 90 °C	f_{ADC}=10 MHz	SMP=1.5	-	-	1.00	
		Resolution = 14 bits	T_J = 125 °C			-	-		
		Resolution = 12 bits				-	-		
		Resolution = 10 bits				-	-		
		Resolution = 8 bits				-	-		
t_{TRIG}	External trigger period	Resolution = 16 bits				-	-	10	$1/f_{ADC}$
V_{AIN}(4)	Conversion voltage range	-				0	-	V_{REF+}	V
V_{CMIV}	Common mode input voltage	-				$V_{REF}/2$ − 10%	$V_{REF}/2$	$V_{REF}/2$ + 10%	V
R_{AIN}(5)	External input impedance	Resolution = 16 bits, T_J = 125 °C				-	-	170	Ω
		Resolution = 14 bits, T_J= 125 °C				-	-	435	
		Resolution = 12 bits, TJ_J = 125 °C				-	-	1150	
		Resolution = 10 bits, T_J = 125 °C				-	-	5650	
		Resolution = 8 bits, T_J = 125 °C				-	-	26500	
C_{ADC}	Internal sample and hold capacitor	-				-	4	-	pF
$t_{ADCVREG_STUP}$	ADC LDO startup time	-				-	5	10	µs

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)	Unit
t_{STAB}	ADC power-up time	LDO already started	1	-	-	conversion cycle
t_{CAL}	Offset and linearity calibration time	-	165010			$1/f_{ADC}$
t_{OFF_CAL}	Offset calibration time	-	1280			
t_{LATR}	Trigger conversion latency for regular and injected channels without aborting the conversion	CKMODE = 00	1.5	2	2.5	
		CKMODE = 01	-	-	2.5	
		CKMODE = 10	-	-	2.5	
		CKMODE = 11	-	-	2.25	
$t_{LATRINJ}$	Trigger conversion latency for regular and injected channels when a regular conversion is aborted	CKMODE = 00	2.5	3	3.5	
		CKMODE = 01	-	-	3.5	
		CKMODE = 10	-	-	3.5	
		CKMODE = 11	-	-	3.25	
t_S	Sampling time	-	1.5	-	810.5	
t_{CONV}	Total conversion time (including sampling time)	N-bits resolution	tS + 0.5 + N/2	-	-	
$I_{DDA_D(ADC)}$	ADC consumption on V_{DDA}, BOOST=11, Differential mode	Resolution = 16 bits, f_{ADC}=25 MHz	-	1440	-	µA
		Resolution = 14 bits, f_{ADC}=30 MHz	-	1350	-	
		Resolution = 12 bits, f_{ADC}=40 MHz	-	990	-	
	ADC consumption on V_{DDA}, BOOST=10, Differential mode f_{ADC}=25 MHz	Resolution = 16 bits	-	1080	-	
		Resolution = 14 bits	-	810	-	
		Resolution = 12 bits	-	585	-	
	ADC consumption on V_{DDA}, BOOST=01, Differential mode f_{ADC}=12.5 MHz	Resolution = 16 bits	-	630	-	
		Resolution = 14 bits	-	432	-	
		Resolution = 12 bits	-	315	-	
	ADC consumption on V_{DDA}, BOOST=00, Differential mode f_{ADC}=6.25 MHz	Resolution = 16 bits	-	360	-	
		Resolution = 14 bits	-	270	-	
		Resolution = 12 bits	-	225	-	
$I_{DDA_SE(ADC)}$	ADC consumption on V_{DDA}, BOOST=11, Single-ended mode	Resolution = 16 bits, f_{ADC}=25 MHz	-	720	-	µA
		Resolution = 14 bits, f_{ADC}=30 MHz	-	675	-	
		Resolution = 12 bits, f_{ADC}=40 MHz	-	495	-	
	ADC consumption on V_{DDA}, BOOST=10, Single-ended mode f_{ADC}=25 MHz	Resolution = 16 bits	-	540	-	
		Resolution = 14 bits	-	405	-	
		Resolution = 12 bits	-	292.5	-	
	ADC consumption on V_{DDA}, BOOST=01, Single-ended mode f_{ADC}=12.5 MHz	Resolution = 16 bits	-	315	-	
		Resolution = 14 bits	-	216	-	
		Resolution = 12 bits	-	157.5	-	
	ADC consumption on V_{DDA}, BOOST=00, Single-ended mode f_{ADC}=6.25 MHz	Resolution = 16 bits	-	180	-	
		Resolution = 14 bits	-	135	-	
		Resolution = 12 bits	-	112.5	-	
$I_{DD(ADC)}$	ADC consumption on V_{DD}	f_{ADC}=50 MHz	-	400	-	

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)	Unit
$I_{DD(ADC)}$	ADC consumption on V_{DD}	f_{ADC}=25 MHz	-	220	-	µA
		f_{ADC}=12.5 MHz	-	180	-	
		f_{ADC}=6.25 MHz	-	120	-	
		f_{ADC}=3.125 MHz	-	80	-	

1. *Guaranteed by design.*
2. *Depending on the package, V_{REF+} can be internally connected to V_{DDA} and V_{REF-} to V_{SSA}.*
3. *These values are valid UFBGA176+25 and one ADC. The values for other packages and multiple ADCs might be different*
4. *The voltage booster on ADC switches must be used for V_{DDA} < 2.4 V (embedded I/O switches).*
5. *The tolerance is 10 LSBs for 16-bit resolution, 4 LSBs for 14-bit resolution, and 2 LSBs for 12-bit, 10-bit and 8-bit resolutions.*

Table 92. Minimum sampling time vs R_{AIN}

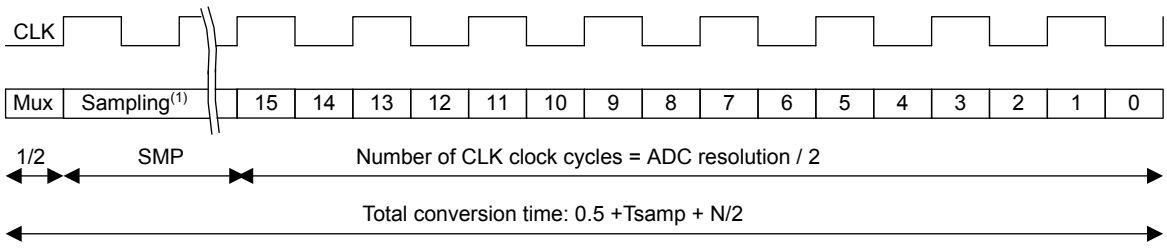
Data valid up to 130 °C, with a 47 pF PCB capacitor and V_{DDA}=1.6 V.

Resolution	RAIN (Ω)	Minimum sampling time (s)		
		Direct channels(1)(2)	Fast channels(1)(3)	Slow channels(1)(4)
16 bits	47	7.37E-08	1.14E-07	1.72E-07
14 bits	47	6.29E-08	9.74E-08	1.55E-07
	68	6.84E-08	1.02E-07	1.58E-07
	100	7.80E-08	1.12E-07	1.62E-07
	150	9.86E-08	1.32E-07	1.80E-07
	220	1.32E-07	1.61E-07	2.01E-07
12 bits	47	5.32E-08	8.00E-08	1.29E-07
	68	5.74E-08	8.50E-08	1.32E-07
	100	6.58E-08	9.31E-08	1.40E-07
	150	8.37E-08	1.10E-07	1.51E-07
	220	1.11E-07	1.34E-07	1.73E-07
	330	1.56E-07	1.78E-07	2.14E-07
	470	2.16E-07	2.39E-07	2.68E-07
	680	3.01E-07	3.29E-07	3.54E-07
10 bits	47	4.34E-08	6.51E-08	1.08E-07
	68	4.68E-08	6.89E-08	1.11E-07
	100	5.35E-08	7.55E-08	1.16E-07
	150	6.68E-08	8.77E-08	1.26E-07
	220	8.80E-08	1.08E-07	1.40E-07
	330	1.24E-07	1.43E-07	1.71E-07
	470	1.69E-07	1.89E-07	2.13E-07
	680	2.38E-07	2.60E-07	2.80E-07
	1000	3.45E-07	3.66E-07	3.84E-07
	1500	5.15E-07	5.35E-07	5.48E-07
	2200	7.42E-07	7.75E-07	7.78E-07
	3300	1.10E-06	1.14E-06	1.14E-06
8 bits	47	3.32E-08	5.10E-08	8.61E-08
	68	3.59E-08	5.35E-08	8.83E-08

Resolution	RAIN (Ω)	Minimum sampling time (s)		
		Direct channels(1)(2)	Fast channels(1)(3)	Slow channels(1)(4)
8 bits	100	4.10E-08	5.83E-08	9.22E-08
	150	5.06E-08	6.76E-08	9.95E-08
	220	6.61E-08	8.22E-08	1.11E-07
	330	9.17E-08	1.08E-07	1.32E-07
	470	1.24E-07	1.40E-07	1.63E-07
	680	1.74E-07	1.91E-07	2.12E-07
	1000	2.53E-07	2.70E-07	2.85E-07
	1500	3.73E-07	3.93E-07	4.05E-07
	2200	5.39E-07	5.67E-07	5.75E-07
	3300	8.02E-07	8.36E-07	8.38E-07
	4700	1.13E-06	1.18E-06	1.18E-06
	6800	1.62E-06	1.69E-06	1.68E-06
	10000	2.36E-06	2.47E-06	2.45E-06
	15000	3.50E-06	3.69E-06	3.65E-06

1. *Guaranteed by design.*
2. *Direct channels are connected to analog I/Os (PA0_C, PA1_C, PC2_C and PC3_C) to optimize ADC performance.*
3. *Fast channels correspond for ADCx_INPx to PA6, PB1, PC4, PF11, PF13 and for ADCx_INNx to PA7, PB0, PC5, PF12, PF14*
4. *Slow channels correspond to all ADC inputs except for the Direct and Fast channels.*

Figure 45. ADC conversion timing diagram



1. The sampling time defines the minimum sampling clock cycles (SMP) to be programmed in the ADC (refer to the product reference manual for details).

Table 93. ADC accuracy

Data guaranteed by characterization for BGA packages. The values for LQFP packages might differ. ADC DC accuracy values are measured after internal calibration.

Symbol	Parameter	Conditions(1)		Min	Typ	Max	Unit
ET	Total undadjusted error	Direct channel	Single ended	-	+10/–20	-	LSB
			Differential	-	±15	-	
		Fast channel	Single ended	-	+10/–20	-	
			Differential	-	±15	-	
		Slow channel	Single ended	-	±10	-	
			Differential	-	±10	-	

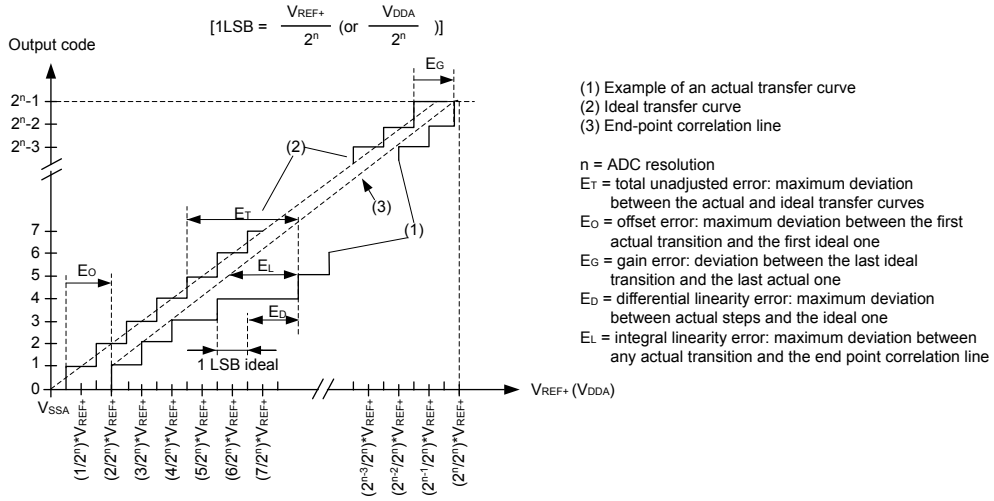
Symbol	Parameter	Conditions[1]		Min	Typ	Max	Unit
EO	Offset error	-		-	±10	-	LSB
EG	Gain error	-		-	±15	-	
ED	Differential linearity error	Single ended		-	+3/–1	-	
		Differential		-	+4.5/–1	-	
EL	Integral linearity error	Direct channel	Single ended	-	±11	-	
			Differential	-	±7	-	
		Fast channel	Single ended	-	±13	-	
			Differential	-	±7	-	
		Slow channel	Single ended	-	±10	-	
			Differential	-	±6	-	
ENOB	Effective number of bits	Single ended		-	12.2	-	Bits
		Differential		-	13.2	-	
SINAD	Signal-to-noise and distortion ratio	Single ended		-	75.2	-	dB
		Differential		-	81.2	-	
SNR	Signal-to-noise ratio	Single ended		-	77.0	-	
		Differential		-	81.0	-	
THD	Total harmonic distortion	Single ended		-	87	-	
		Differential		-	90	-	

1. *ADC clock frequency = 25 MHz, ADC resolution = 16 bits, V_{DDA}=V_{REF+}=3.3 V and BOOST=11.*

Note: ADC accuracy vs. negative injection current: injecting a negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative currents.

Any positive injection current within the limits specified for $I_{INJ(PIN)}$ and $\Sigma I_{INJ(PIN)}$ in Section 6.3.15 I/O current injection characteristics does not affect the ADC accuracy.

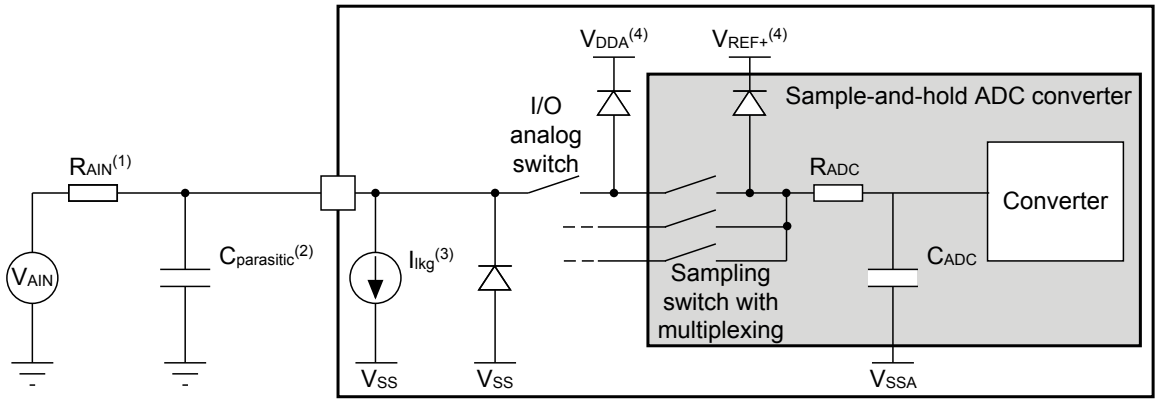
Figure 46. ADC accuracy characteristics



1. Example of an actual transfer curve.

2. Ideal transfer curve.
3. End point correlation line.
4. E_T = Total Unadjusted Error: maximum deviation between the actual and the ideal transfer curves.
5. EO = Offset Error: deviation between the first actual transition and the first ideal one.
6. EG = Gain Error: deviation between the last ideal transition and the last actual one.
7. ED = Differential Linearity Error: maximum deviation between actual steps and the ideal one.
8. EL = Integral Linearity Error: maximum deviation between any actual transition and the end point correlation line.

Figure 47. Typical connection diagram using the ADC with FT/TT pins featuring analog switch function



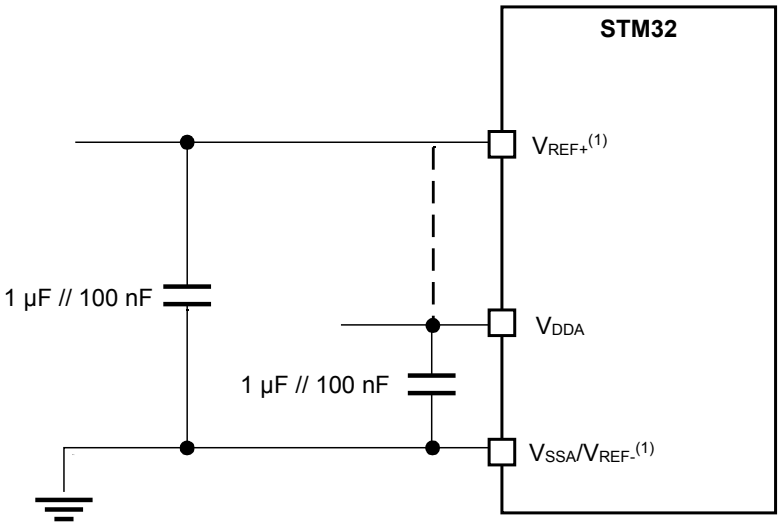
MSv67871V3

1. Refer to Table 91. ADC characteristics for the values of R_{AIN} and C_{ADC}.
2. $C_{parasitic}$ represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (refer to Table 63. I/O static characteristics). A high $C_{parasitic}$ value downgrades conversion accuracy. To remedy this, f_{ADC} should be reduced.
3. Refer to Table 63. I/O static characteristics for the value of I_{lkg}.
4. Refer to .Figure 12. Power supply scheme.

General PCB design guidelines

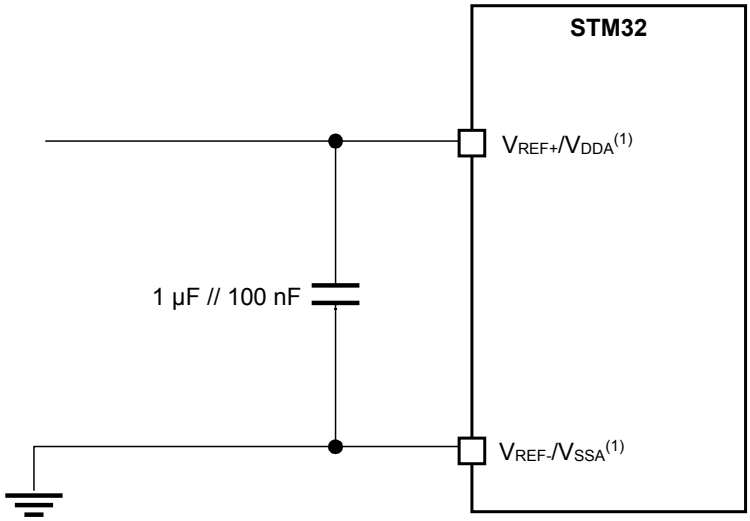
Power supply decoupling should be performed as shown in Figure 48. Power supply and reference decoupling (V_{REF+} not connected to V_{DDA}) or Figure 49. Power supply and reference decoupling (V_{REF+} connected to V_{DDA}), depending on whether V_{REF+} is connected to V_{DDA} or not. The 100 nF capacitors should be ceramic (good quality). They should be placed them as close as possible to the chip.

Figure 48. Power supply and reference decoupling (V_{REF+} not connected to V_{DDA})



1. V_{REF+} input is not available on all package (refer to Table 1. STM32H7B0xB features and peripheral counts) whereas V_{REF-} is available only on UFBGA176+25, TFBGA225 with SMPS and TFBGA216. When V_{REF-} is not available, it is internally connected to V_{SSA}.

Figure 49. Power supply and reference decoupling (V_{REF+} connected to V_{DDA})



1. V_{REF+} input is not available on all package (refer to Table 1. STM32H7B0xB features and peripheral counts) whereas V_{REF-} is available only on UFBGA176+25, TFBGA225 with SMPS and TFBGA216. When V_{REF-} is not available, it is internally connected to V_{SSA}.

6.3.22 DAC characteristics

Table 94. DAC characteristics

Symbol	Parameter	Conditions		Min(1)	Typ(1)	Max(1)	Unit
V_{DDA}	Analog supply voltage	-		1.8	3.3	3.6	V
V_{REF+}	Positive reference voltage	-		1.80	-	V_{DDA}	
V_{REF-}	Negative reference voltage	-		-	V_{SSA}	-	
R_L	Resistive Load	DAC output buffer ON	connected to V_{SSA}	5	-	-	kΩ
			connected to V_{DDA}	25	-	-	
R_O	Output Impedance	DAC output buffer OFF		10.3	13	16	
R_{BON}	Output impedance sample and hold mode, output buffer ON	DAC output buffer ON	V_{DD} = 2.7 V	-	-	1.6	kΩ
			V_{DD} = 2.0 V	-	-	2.6	
R_{BOFF}	Output impedance sample and hold mode, output buffer OFF	DAC output buffer OFF	V_{DD} = 2.7 V	-	-	17.8	kΩ
			V_{DD} = 2.0 V	-	-	18.7	
C_L	Capacitive Load	DAC output buffer OFF		-	-	50	pF
C_{SH}		Sample and Hold mode		-	0.1	1	µF
V_{DAC_OUT}	Voltage on DAC_OUT output	DAC output buffer ON		0.2	-	V_{DDA} −0.2	V
		DAC output buffer OFF		0	-	V_{REF+}	
$t_{SETTLING}$	Settling time (full scale: for a 12-bit code transition between the lowest and the highest input codes when DAC_OUT reaches the final value of ±0.5LSB, ±1LSB, ±2LSB, ±4LSB, ±8LSB)	Normal mode, DAC output buffer ON, $C_L \le 50$ pF, $R_L \ge 5$ kΩ	±0.5 LSB	-	2.05	-	µs
			±1 LSB	-	1.97	-	
			±2 LSB	-	1.67	-	
			±4 LSB	-	1.66	-	
			±8 LSB	-	1.65	-	
		Normal mode, DAC output buffer OFF, ±1LSB C_L=10 pF		-	1.7	2	
t_{WAKEUP}(2)	Wakeup time from off state (setting the ENx bit in the DAC Control register) until the final value of ±1LSB is reached	Normal mode, DAC output buffer ON, $C_L \le 50$ pF, R_L = 5 kΩ		-	5	7.5	µs
		Normal mode, DAC output buffer OFF, $C_L \le 10$ pF		-	2	5	
PSRR	DC V_{DDA} supply rejection ratio	Normal mode, DAC output buffer ON, $C_L \le 50$ pF, R_L = 5 kΩ		-	−80	−28	dB
t_{SAMP}	Sampling time in Sample and Hold mode C_L=100 nF (code transition between the lowest input code and the highest input code when DAC_OUT reaches the ±1LSB final value)	MODE<2:0>_V12=100/101 (BUFFER ON)		-	0.7	2.6	ms
		MODE<2:0>_V12=110 (BUFFER OFF)		-	11.5	18.7	
		MODE<2:0>_V12=111(3) (INTERNAL BUFFER OFF)		-	0.3	0.6	µs
I_{leak}	Output leakage current	-		-	-	(4)	nA
C_{lint}	Internal sample and hold capacitor	-		1.8	2.2	2.6	pF
t_{TRIM}	Middle code offset trim time	Minimum time to verify the each code		50	-	-	µs
V_{offset}	Middle code offset for 1 trim code step	V_{REF+} = 3.6 V		-	850	-	µV
		V_{REF+} = 1.8 V		-	425	-	

Symbol	Parameter	Conditions		Min(1)	Typ(1)	Max(1)	Unit
$I_{DDA(DAC)}$	DAC quiescent consumption from V_{DDA}	DAC output buffer ON	No load, middle code (0x800)	-	360	-	µA
			No load, worst code (0xF1C)	-	490	-	
		DAC output buffer OFF	No load, middle/worst code (0x800)	-	20	-	
		Sample and Hold mode, C_{SH}=100 nF		-	360*T_{ON}/(T_{ON}+T_{OFF})(5)	-	
$I_{DDV(DAC)}$	DAC consumption from V_{REF+}	DAC output buffer ON	No load, middle code (0x800)	-	170	-	
			No load, worst code (0xF1C)	-	170	-	
		DAC output buffer OFF	No load, middle/worst code (0x800)	-	160	-	
		Sample and Hold mode, Buffer ON, C_{SH}=100 nF (worst code)		-	170*T_{ON}/(T_{ON}+T_{OFF})(5)	-	
		Sample and Hold mode, Buffer OFF, C_{SH}=100 nF (worst code)		-	160*T_{ON}/(T_{ON}+T_{OFF})(5)	-	

1. *Guaranteed by design, unless otherwise specified.*
2. *In buffered mode, the output can overshoot above the final value for low input code (starting from the minimum value).*
3. *DACx_OUT pin is not connected externally (internal connection only).*
4. *Refer to Table 63. I/O static characteristics.*
5. *T_{ON} is the refresh phase duration, while T_{OFF} is the hold phase duration. Refer to the product reference manual for more details.*

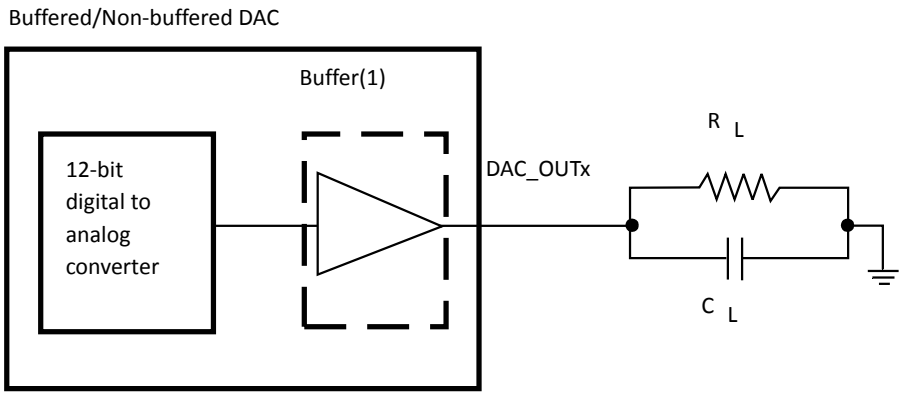
Table 95. DAC accuracy

Symbol	Parameter	Conditions		Min(1)	Typ(1)	Max(1)	Unit
DNL	Differential non linearity(2)	DAC output buffer ON		−2	-	2	LSB
		DAC output buffer OFF		−2	-	2	
INL	Integral non linearity(3)	DAC output buffer ON, C_L ≤ 50 pF, R_L ≥ 5 kΩ		−4	-	4	LSB
		DAC output buffer OFF, C_L ≤ 50 pF, no R_L		−4	-	4	
Offset	Offset error at code 0x800(3)	DAC output buffer ON, C_L ≤ 50 pF, R_L ≥ 5 kΩ	V_{REF+} = 3.6 V	-	-	±12	LSB
			V_{REF+} = 1.8 V	-	-	±25	
		DAC output buffer OFF, C_L ≤ 50 pF, no R_L	-	-	-	±8	
Offset1	Offset error at code 0x001(4)	DAC output buffer OFF, C_L ≤ 50 pF, no R_L		-	-	±5	LSB
OffsetCal	Offset error at code 0x800 after factory calibration	DAC output buffer ON, C_L ≤ 50 pF, R_L ≥ 5 kΩ	V_{REF+} = 3.6 V	-	-	±5	LSB
			VR_{EF+} = 1.8 V	-	-	±7	
Gain	Gain error(5)	DAC output buffer ON,C_L ≤ 50 pF, R_L ≥ 5 kΩ		-	-	±1	%
		DAC output buffer OFF, C_L ≤ 50 pF, no R_L		-	-	±1	
TUE	Total undajusted error	DAC output buffer ON,C_L ≤ 50 pF, R_L ≥ 5 kΩ		-	-	±30	LSB
		DAC output buffer OFF, C_L ≤ 50 pF, no R_L		-	-	±12	
TUECal	Total undajusted error after calibration	DAC output buffer ON C_L ≤ 50pF, R_L ≥ 5kΩ		-	-	±23	LSB
SNR	Signal-to-noise ratio(6)	DAC output buffer ON C_L ≤ 50pF, R_L ≥ 5kΩ 1 kHz, BW 500KHz		-	67.8	-	dB
		DAC output buffer OFF C_L ≤ 50pF, no R_L 1kHz, BW 500KHz		-	67.8	-	

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)	Unit
THD	Total harmonic distorsion(6)	DAC output buffer ON $C_L \le 50pF$, $R_L \ge 5k\Omega$, 1 kHz	-	−78,6	-	dB
		DAC output buffer OFF $C_L \le 50pF$, no R_L, 1 kHz	-	−78,6	-	
SINAD	Signal-to-noise and distorsion ratio(6)	DAC output buffer ON $C_L \le 50pF$, $R_L \ge 5k\Omega$, 1 kHz	-	67.5	-	dB
		DAC output buffer OFF $C_L \le 50pF$, no R_L, 1 kHz	-	67.5	-	
ENOB	Effective number of bits	DAC output buffer ON $C_L \le 50pF$, $R_L \ge 5k\Omega$, 1 kHz	-	10.9	-	dB
		DAC output buffer OFF $C_L \le 50pF$, no R_L, 1 kHz	-	10.9	-	

1. *Guaranteed by design, unless otherwise specified.*
2. *Difference between two consecutive codes minus 1 LSB.*
3. *Difference between the value measured at Code i and the value measured at Code i on a line drawn between Code 0 and last Code 4095.*
4. *Difference between the value measured at Code (0x001) and the ideal value.*
5. *Difference between the ideal slope of the transfer function and the measured slope computed from code 0x000 and 0xFFF when the buffer is OFF, and from code giving 0.2 V and (V_{REF+} − 0.2 V) when the buffer is ON.*
6. *Signal is −0.5dBFS with Fsampling = 1 MHz.*

Figure 50. 12-bit buffered /non-buffered DAC



1. The DAC integrates an output buffer that can be used to reduce the output impedance and to drive external loads directly without the use of an external operational amplifier. The buffer can be bypassed by configuring the BOFFx bit in the DAC_CR register.

6.3.23 Voltage reference buffer characteristics

Table 96. VREFBUF characteristics

Symbol	Parameter	Conditions		Min(1)	Typ(1)	Max(1)	Unit
V_{DDA}	Analog supply voltage	Normal mode	VSCALE = 000	2.8	3.3	3.6	V
			VSCALE = 001	2.4	-	3.6	
			VSCALE = 010	2.1	-	3.6	
			VSCALE = 011	1.8	-	3.6	
		Degraded mode(2)	VSCALE = 000	1.62	-	2.80	
			VSCALE = 001	1.62	-	2.40	
			VSCALE = 010	1.62	-	2.10	
			VSCALE = 011	1.62	-	1.80	

Symbol	Parameter	Conditions		Min(1)	Typ(1)	Max(1)	Unit
V_{REFBUF_OUT}	Voltage Reference Buffer Output	Normal mode at 30°C, I_{LOAD}=100 µA	VSCALE = 000	2.496(3)	2.5000	2.504(3)	V
			VSCALE = 001	2,0460	2.0490	2,0520	
			VSCALE = 010	1,8010	1.8040	1,8060	
			VSCALE = 011	1,4995	1.5015	1,5040	
		Degraded mode(2)	VSCALE = 000	VDDA− 150 mV	-	VDDA	
			VSCALE = 001	VDDA− 150 mV	-	VDDA	
			VSCALE = 010	VDDA− 150 mV	-	VDDA	
			VSCALE = 011	VDDA− 150 mV	-	VDDA	
TRIM	Trim step resolution	-	-	-	±0.05	±0.1	%
C_L	Load capacitor	-	-	0.5	1	1.50	uF
esr	Equivalent Serial Resistor of C_L	-	-	-	-	2	Ω
I_{LOAD}	Static load current	-	-	-	-	4	mA
I_{line_reg}	Line regulation	2.8 V ≤ V_{DDA} ≤ 3.6 V	I_{LOAD} = 500 µA	-	200	-	ppm/V
			I_{LOAD} = 4 mA	-	100	-	
I_{LOAD_reg}	Load regulation	500 µA ≤ I_{LOAD} ≤ 4 mA	Normal Mode	-	50	-	ppm/ mA
T_{coeff}	Temperature coefficient	−40 °C < T_J < +130 °C	-	-	-	Tcoeff VREFINT + 100	ppm/ °C
PSRR	Power supply rejection	DC	-	-	60	-	dB
		100KHz	-	-	40	-	
t_{START}	Startup time	C_L=0.5 µF	-	-	300	-	µs
		C_L=1 µF	-	-	500	-	
		C_L=1.5 µF	-	-	650	-	
I_{INRUSH}	Control of maximum DC current drive on V_{REFBUF_OUT} during startup phase(4)	-		-	8	-	mA
$I_{DDA(VREFBUF)}$	V_{REFBUF} consumption from V_{DDA}	I_{LOAD} = 0 µA	-	-	15	25	µA
		I_{LOAD} = 500 µA	-	-	16	30	
		I_{LOAD} = 4 mA	-	-	32	50	

1. *Guaranteed by design, unless otherwise specified.*
2. *In degraded mode, the voltage reference buffer cannot accurately maintain the output voltage (V_{DDA}−drop voltage).*
3. *Guaranteed by tests in production.*
4. *To properly control VREFBUF I_{INRUSH} current during the startup phase and the change of scaling, VDDA voltage should be in the range of 1.8 V-3.6 V, 2.1 V-3.6 V, 2.4 V-3.6 V and 2.8 V-3.6 V for VSCALE = 011, 010, 001 and 000, respectively.*

6.3.24 Analog temperature sensor characteristics

Table 97. Analog temperature sensor characteristics

Symbol	Parameter	Min	Typ	Max	Unit
T_L (1)	V_{SENSE} linearity with temperature (from V_{SENSOR} voltage)	-	-	3	°C
	V_{SENSE} linearity with temperature (from ADC counter)	-	-	3	
Avg_Slope (2)	Average slope (from V_{SENSOR} voltage)	-	2	-	mV/°C
	Average slope (from ADC counter)	-	2	-	
V_{30} (3)	Voltage at 30°C ± 5 °C	-	0.62	-	V
t_{start_run} (1)	Startup time in Run mode (buffer startup)	-	-	25.2	µs
t_{S_temp} (1)	ADC sampling time when reading the temperature	9	-	-	
I_{sens} (1)	Sensor consumption	-	0.18	0.31	µA
$I_{sensbuf}$ (1)	Sensor buffer consumption	-	3.8	6.5	

1. *Guaranteed by design.*
2. *Guaranteed by characterization results.*
3. *Measured at V_{DDA} = 3.3 V ± 10 mV. The V_{30} ADC conversion result is stored in the TS_CAL1 byte.*

Table 98. Analog temperature sensor calibration values

Symbol	Parameter	Memory address
TS_CAL1	Temperature sensor raw data acquired value at 30 °C, V_{DDA}=3.3 V	0x08FF F814 - 0x08FF F816
TS_CAL2	Temperature sensor raw data acquired value at 130 °C, V_{DDA}=3.3 V	0x08FF F818 - 0x08FF F81A

6.3.25 Digital temperature sensor characteristics

Table 99. Digital temperature sensor characteristics

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)	Unit
f_{DTS} (2)	Output Clock frequency	-	500	750	1150	kHz
T_{LC} (2)	Temperature linearity coefficient	VOS2	1660	2100	2750	Hz/°C
T_{TOTAL_ERROR} (2)	Temperature offset measurement, all VOS	T_J =−40 °C to 30 °C	−13	-	4	°C
		T_J =30 °C to 130 °C	−7	-	2	
T_{VDD_CORE}	Additional error due to supply variation	VOS2	0	-	0	°C
		VOS0, VOS1, VOS3	−1	-	1	
t_{TRIM}	Calibration time	-	-	-	2	ms
t_{WAKE_UP}	Wake-up time from off state until DTS ready bit is set	-	-	67	116.00	µs
I_{DDCORE_DTS}	DTS consumption on V_{CORE}	-	8.5	30	70.0	µA

1. *Guaranteed by design, unless otherwise specified.*
2. *Guaranteed by characterization results.*

6.3.26 Temperature and V_{BAT} monitoring

Table 100. V_{BAT} monitoring characteristics

Symbol	Parameter	Min	Typ	Max	Unit
R	Resistor bridge for V_{BAT}	-	26	-	KΩ
Q	Ratio on V_{BAT} measurement	-	4	-	-
Er(1)	Error on Q	–10	-	+10	%
t_{S_vbat}(1)	ADC sampling time when reading V_{BAT} input	9	-	-	µs
$V_{BAThigh}$	High supply monitoring	-	3.55	-	V
V_{BATlow}	Low supply monitoring	-	1.36	-	

1. *Guaranteed by design.*

Table 101. V_{BAT} charging characteristics

Symbol	Parameter	Condition	Min	Typ	Max	Unit
R_{BC}	Battery charging resistor	VBRS in PWR_CR3= 0	-	5	-	KΩ
		VBRS in PWR_CR3= 1	-	1.5	-	

Table 102. Temperature monitoring characteristics

Symbol	Parameter	Min(1)	Typ(1)	Max(1)	Unit
$TEMP_{high}$	High temperature monitoring	-	117	-	°C
$TEMP_{low}$	Low temperature monitoring	-	–25	-	

1. *Guaranteed by design.*

6.3.27 Voltage booster for analog switch

Table 103. Voltage booster for analog switch characteristics

Symbol	Parameter	Condition	Min(1)	Typ(1)	Max(1)	Unit
V_{DD}	Supply voltage	-	1.62	2.6	3.6	V
$t_{SU(BOOST)}$	Booster startup time	-	-	-	50	µs
$I_{DD(BOOST)}$	Booster consumption	1.62 V ≤ V_{DD} ≤ 2.7 V	-	-	125	µA
		2.7 V < V_{DD} < 3.6 V	-	-	250	

1. *Guaranteed by characterization results.*

6.3.28 Comparator characteristics

Table 104. COMP characteristics

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)	Unit
V_{DDA}	Analog supply voltage	-	1.62	3.3	3.6	V
V_{IN}	Comparator input voltage range	-	0	-	V_{DDA}	
V_{BG}(2)	Scaler input voltage	-		-		
V_{SC}	Scaler offset voltage	-	-	±5	±10	mV

Symbol	Parameter	Conditions		Min(1)	Typ(1)	Max(1)	Unit
$I_{DDA(SCALER)}$	Scaler static consumption from V_{DDA}	BRG_EN=0 (bridge disable)		-	0.2	0.3	µA
		BRG_EN=1 (bridge enable)		-	0.8	1	
t_{START_SCALER}	Scaler startup time	-		-	140	250	µs
t_{START}	Comparator startup time to reach propagation delay specification	High-speed mode		-	2	5	µs
		Medium mode		-	5	20	
		Ultra-low-power mode		-	15	80	
t_D(3)	Propagation delay for 200 mV step with 100 mV overdrive	High-speed mode		-	50	80	ns
		Medium mode		-	0.5	0.9	µs
		Ultra-low-power mode		-	2.5	7	
	Propagation delay for step > 200 mV with 100 mV overdrive only on positive inputs	High-speed mode		-	50	120	ns
		Medium mode		-	0.5	1.2	µs
		Ultra-low-power mode		-	2.5	7	
V_{offset}	Comparator offset error	Full common mode range		-	±5	±20	mV
V_{hys}	Comparator hysteresis	No hysteresis		-	0	-	mV
		Low hysteresis		4	10	22	
		Medium hysteresis		8	20	37	
		High hysteresis		16	30	52	
$I_{DDA(COMP)}$	Comparator consumption from V_{DDA}	Ultra-low-power mode	Static	-	400	600	nA
			With 50 kHz ±100 mV overdrive square signal	-	800	-	
		Medium mode	Static	-	5	7	µA
			With 50 kHz ±100 mV overdrive square signal	-	6	-	
		High-speed mode	Static	-	70	100	
			With 50 kHz ±100 mV overdrive square signal	-	75	-	

1. *Guaranteed by design, unless otherwise specified.*
2. *Refer to Section 6.3.6 Embedded reference voltage.*
3. *Guaranteed by characterization results.*

6.3.29 Operational amplifier characteristics

Table 105. Operational amplifier characteristics

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)	Unit
V_{DDA}	Analog supply voltage Range	-	2	3.3	3.6	V
CMIR	Common Mode Input Range	-	0	-	V_{DDA}	
VI_{OFFSET}	Input offset voltage	25°C, no load on output	-	-	±1.5	mV
		All voltages and temperature, no load	-	-	±2.5	
ΔVI_{OFFSET}	Input offset voltage drift	-	-	±3.0	-	µV/°C
TRIMOFFSETP, TRIMLPOFFSETP	Offset trim step at low common input voltage ($0.1*V_{DDA}$)	-	-	1.1	1.5	mV

Symbol	Parameter	Conditions		Min(1)	Typ(1)	Max(1)	Unit
TRIMOFFSETN, TRIMLPOFFSETN	Offset trim step at high common input voltage (0.9*V_{DDA})	-		-	1.1	1.5	mV
I_{LOAD}	Drive current	-		-	-	500	µA
I_{LOAD_PGA}	Drive current in PGA mode	-		-	-	270	
C_{LOAD}	Capacitive load	-		-	-	50	pF
CMRR	Common mode rejection ratio	-		-	80	-	dB
PSRR	Power supply rejection ratio	C_{LOAD} ≤ 50pf / R_{LOAD} ≥ 4 kΩ(2) at 1 kHz, V_{com}=V_{DDA}/2		50	66	-	dB
GBW	Gain bandwidth for high supply range	200 mV ≤ Output dynamic range ≤ V_{DDA} - 200 mV		4	7.3	12.3	MHz
SR	Slew rate (from 10% and 90% of output voltage)	Normal mode		-	3	-	V/µs
		High-speed mode		-	24	-	
AO	Open loop gain	200 mV ≤ Output dynamic range ≤ V_{DDA} - 200 mV		59	90	129	dB
φm	Phase margin	-		-	55	-	°
GM	Gain margin	-		-	12	-	dB
V_{OHSAT}	High saturation voltage	I_{load}=max or R_{LOAD}=min, Input at V_{DDA}		V_{DDA} −100 mV	-	-	mV
V_{OLSAT}	Low saturation voltage	I_{load}=max or R_{LOAD}=min, Input at 0 V		-	-	100	
t_{WAKEUP}	Wake up time from OFF state	Normal mode	C_{LOAD} ≤ 50pf, R_{LOAD} ≥ 4 kΩ, follower configuration	-	0.8	3.2	µs
		High speed mode	C_{LOAD} ≤ 50pf, R_{LOAD} ≥ 4 kΩ, follower configuration	-	0.9	2.8	
PGA gain	Non inverting gain error value	PGA gain = 2		−1	-	1	%
		PGA gain = 4		−2	-	2	
		PGA gain = 8		−2.5	-	2.5	
		PGA gain = 16		−3	-	3	
	Inverting gain error value	PGA gain = 2		−1	-	1	
		PGA gain = 4		−1	-	1	
		PGA gain = 8		−2	-	2	
		PGA gain = 16		−3	-	3	
	External non-inverting gain error value	PGA gain = 2		−1	-	1	
		PGA gain = 4		−3	-	3	
		PGA gain = 8		−3.5	-	3.5	
		PGA gain = 16		−4	-	4	
$R_{network}$	R2/R1 internal resistance values in non-inverting PGA mode(3)	PGA Gain=2		-	10/10	-	kΩ/ kΩ
		PGA Gain=4		-	30/10	-	
		PGA Gain=8		-	70/10	-	
		PGA Gain=16		-	150/10	-	
	R2/R1 internal resistance values in inverting PGA mode(3)	PGA Gain = -1		-	10/10	-	
		PGA Gain = -3		-	30/10	-	

Symbol	Parameter	Conditions		Min(1)	Typ(1)	Max(1)	Unit
$R_{network}$	R2/R1 internal resistance values in inverting PGA mode(3)	PGA Gain = -7		-	70/10	-	kΩ/ kΩ
		PGA Gain = -15		-	150/10	-	
Delta R	Resistance variation (R1 or R2)	-		−15	-	15	%
PGA BW	PGA bandwidth for different non inverting gain	Gain=2		-	GBW/2	-	MHz
		Gain=4		-	GBW/4	-	
		Gain=8		-	GBW/8	-	
		Gain=16		-	GBW/16	-	
	PGA bandwidth for different inverting gain	Gain = -1		-	5.00	-	MHz
		Gain = -3		-	3.00	-	
		Gain = -7		-	1.50	-	
		Gain = -15		-	0.80	-	
en	Voltage noise density	at 1 KHz	output loaded with 4 kΩ	-	140	-	nV/√Hz
		at 10 KHz		-	55	-	
$I_{DDA(OPAMP)}$	OPAMP consumption from V_{DDA}	Normal mode	no Load, quiescent mode, follower	-	570	1000	µA
		High-speed mode		-	610	1200	

1. *Guaranteed by design, unless otherwise specified.*
2. *R_{LOAD} is the resistive load connected to VSSA or to VDDA.*
3. *R2 is the internal resistance between the OPAMP output and th OPAMP inverting input. R1 is the internal resistance between the OPAMP inverting input and ground. PGA gain = 1 + R2/R1.*

6.3.30 Digital filter for Sigma-Delta Modulators (DFSDM) characteristics

Unless otherwise specified, the parameters given in Table 106. DFSDM measured timing 1.62-3.6 V for DFSDM are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and supply voltage conditions summarized in Table 20. General operating conditions and Section 6.3.1 .

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C_L = 30 pF
- Measurement points are done at CMOS levels: $0.5V_{DD}$
- VOS level set to VOS0

Refer to Section 6.3.16 I/O port characteristics for more details on the input/output alternate function characteristics (DìFSDM_CKINx, DFSDM_DATINx, DFSDM_CKOUT for DFSDM).

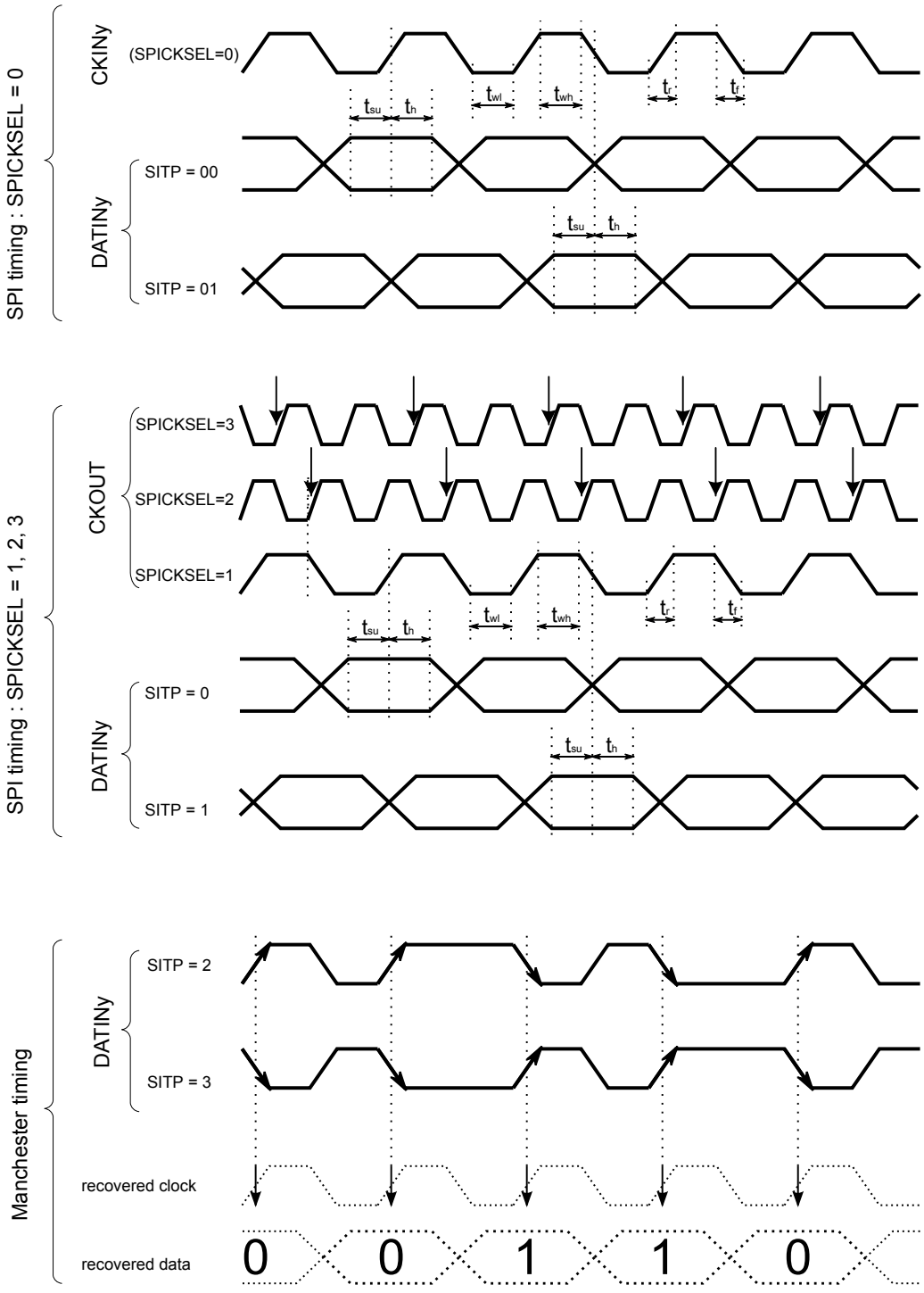
Table 106. DFSDM measured timing 1.62-3.6 V

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
$f_{DFSDMCLK}$	DFSDM clock	1.62 V < V_{DD} < 3.6 V		-	-	(1)	MHz
f_{CKIN} (1/ T_{CKIN})	Input clock frequency	SPI mode (SITP[1:0]=0,1), External clock mode (SPICKSEL[1:0]=0),		-	-	20(2)	
		SPI mode (SITP[1:0]=0,1), Internal clock mode (SPICKSEL[1:0]≠0)		-	-	20(2)	
f_{CKOUT}	Output clock frequency	1.62 < V_{DD} < 3.6 V		-	-	20	
DuCyCKOUT	Output clock frequency duty cycle	1.62 < V_{DD} < 3.6 V	Even division,CKOUTDIV[7:0] = n, 1, 3, 5, ...	45	50	55	%

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
DuCyCKOUT	Output clock frequency duty cycle	1.62 < V_{DD} < 3.6 V	Odd division,CKOUTDIV[7:0] = n, 2, 4, 6, ...	(((n/ 2+1)/(n−1))*100)−5	(((n/2+1)/ (n−1))*100)	(((n/ 2+1)/(n−1))*100)+5	%
$t_{wh(CKIN)}$ $t_{wl(CKIN)}$	Input clock high and low time	SPI mode (SITP[1:0]=0,1), External clock mode (SPICKSEL[1:0]=0), 1.62 < V_{DD} < 3.6 V	-	T_{CKIN}/2 - 0.5	T_{CKIN}/2	-	ns
t_{su}	Data input setup time	SPI mode (SITP[1:0]=0,1), External clock mode (SPICKSEL[1:0]=0), 1.62 < V_{DD} < 3.6 V	-	4	-	-	
t_h	Data input hold time	SPI mode (SITP[1:0]=0,1), External clock mode (SPICKSEL[1:0]=0), 1.62 < V_{DD} < 3.6 V	-	0.5	-	-	
$T_{Manchester}$	Manchester data period (recovered clock period)	Manchester mode (SITP[1:0]=2,3), Internal clock mode (SPICKSEL[1:0]'0), 1.62 < V_{DD} < 3.6 V	-	(CKOUTDIV[7:0]+1) x $T_{DFSDMCLK}$	-	(2*CKOUTDIV[7:0]) x $T_{DFSDMCLK}$(3)	

1. *The maximum DFSDM kernel clock is specified in Section 6.3.1 .*
2. *The internal DFSDMCLK clock must be at least 4 times faster than the external CKIN clock.*
3. *The internal DFSDMCLK must be at least 6 times faster than the Manchester data frequency.*

Figure 51. Channel transceiver timing diagrams



6.3.31 Camera interface (DCMI) timing specifications

Unless otherwise specified, the parameters given in Table 107. DCMI characteristics for DCMI are derived from tests performed under the ambient temperature, f_{HCLK} frequency and VDD supply voltage summarized in Table 20. General operating conditions and Section 6.3.1 , with the following configuration:

- DCMI_PIXCLK polarity: falling
- DCMI_VSYNC and DCMI_HSYNC polarity: high

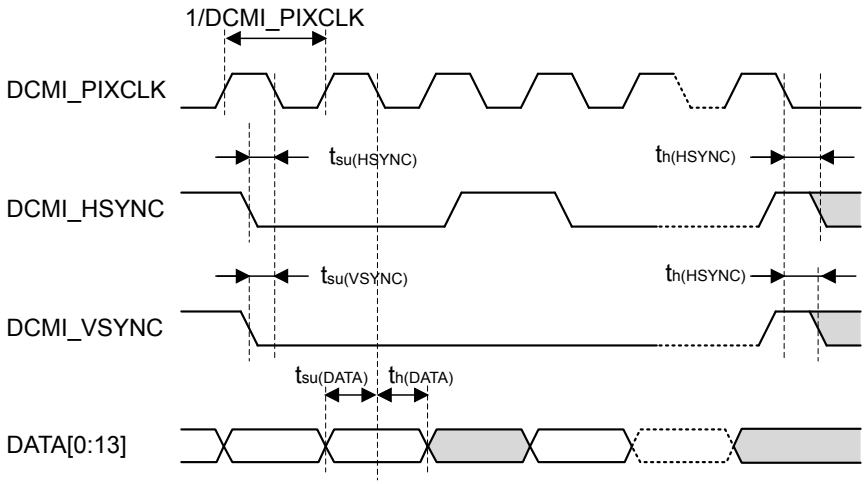
- Data formats: 14 bits
- Capacitive load C_L=30 pF
- Measurement points are done at CMOS levels: $0.5V_{DD}$
- Output speed is set to OSPEEDRy[1:0] = 11
- VOS level set to VOS0

Table 107. DCMI characteristics

Symbol	Parameter	Min(1)	Max(1)	Unit
-	Frequency ratio DCMI_PIXCLK/f_{HCLK}	-	0.4	-
DCMI_PIXCLK	Pixel Clock input	-	80	MHz
D_{pixel}	Pixel Clock input duty cycle	30	70	%
t_{su}(DATA)	Data input setup time	2.5	-	-
t_h(DATA)	Data hold time	1	-	
tsu(HSYNC), tsu(VSYNC)	DCMI_HSYNC/ DCMI_VSYNC input setup time	3	-	ns
th(HSYNC), th(VSYNC)	DCMI_HSYNC/ DCMI_VSYNC input hold time	1	-	-

1. *Guaranteed by design.*

Figure 52. DCMI timing diagram



6.3.32 PSSI interface characteristics

Unless otherwise specified, the parameters given in Table 108 and 109for PSSI are derived from tests performed under the ambient temperature, f_{HCLK} frequency and V_{DD} supply voltage summarized in Table 20. General operating conditions and Section 6.3.1 , with the following configuration:

- PSSI_PDCK polarity: falling
- PSSI_RDY and PSSI_DE polarity: low
- Bus width : 16 lines
- DATA width : 32 bits
- Capacitive load C=30 pF
- Measurement points are done at CMOS levels: $0.5V_{DD}$
- Output speed is set to OSPEEDRy[1:0] = 11

Note: At VOS1, the performance in Transmit mode can be degraded by up to 5 % compared to VOS0. This is indicated by a footnote when applicable.

Table 108. PSSI transmit characteristics

Guaranteed by characterization results.

Symbol	Parameter	Min	Max(1)	Unit
-	Frequency ratio PSSI_PDCK/f_{HCLK}	-	0.4	-
PSSI_PDCK	PSSI clock input	-	50	MHz
D_{pixel}	PSSI clock input duty cycle	30	70	%
$t_{dv(DATA)}$	Data output valid time	-	10	ns
$t_{dh(DATA)}$	Data output hold time	5	-	
$t_{dv((DE)}$	DE output valid time	-	14	
$t_{dh(DE)}$	DE output hold time	6	-	
$t_{su(RDY)}$	RDY input setup time	3	-	
$t_{h(RDY)}$	RDY input hold time	0	-	

1. *At VOS1, these values are degraded by up to 5 %.*

Table 109. PSSI receive characteristics

Guaranteed by characterization results.

Symbol	Parameter	Min	Max
-	Frequency ratio PSSI_PDCK/f_{HCLK}	-	0.4
PSSI_PDCK	PSSI clock input	-	100
D_{pixel}	PSSI clock input duty cycle	30	70
$t_{su(DATA)}$	Data input setup time	2	-
$t_{h(DATA)}$	Data input hold time	1	-
$t_{su((DE)}$	DE input setup time	3	-
$t_{h(DE)}$	DE input hold time	1	-
$t_{ov(RDY)}$	RDY output valid time	-	10
$t_{oh(RDY)}$	RDY output hold time	4.5	-

Figure 53. **PSSI timing diagram in Transmit mode**

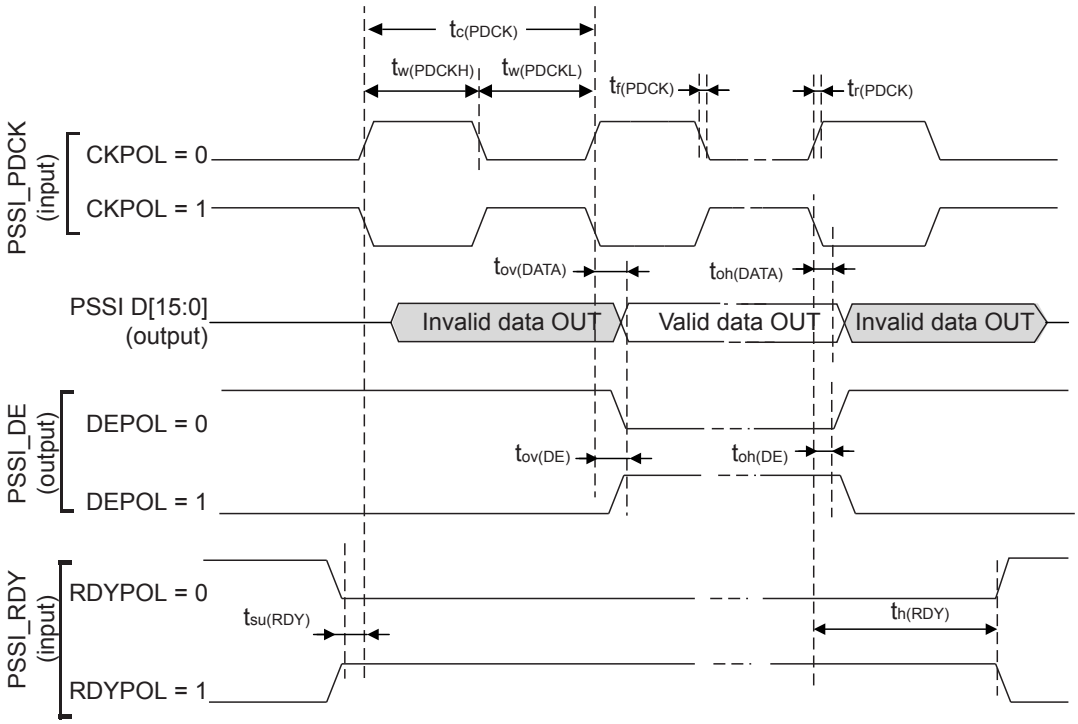
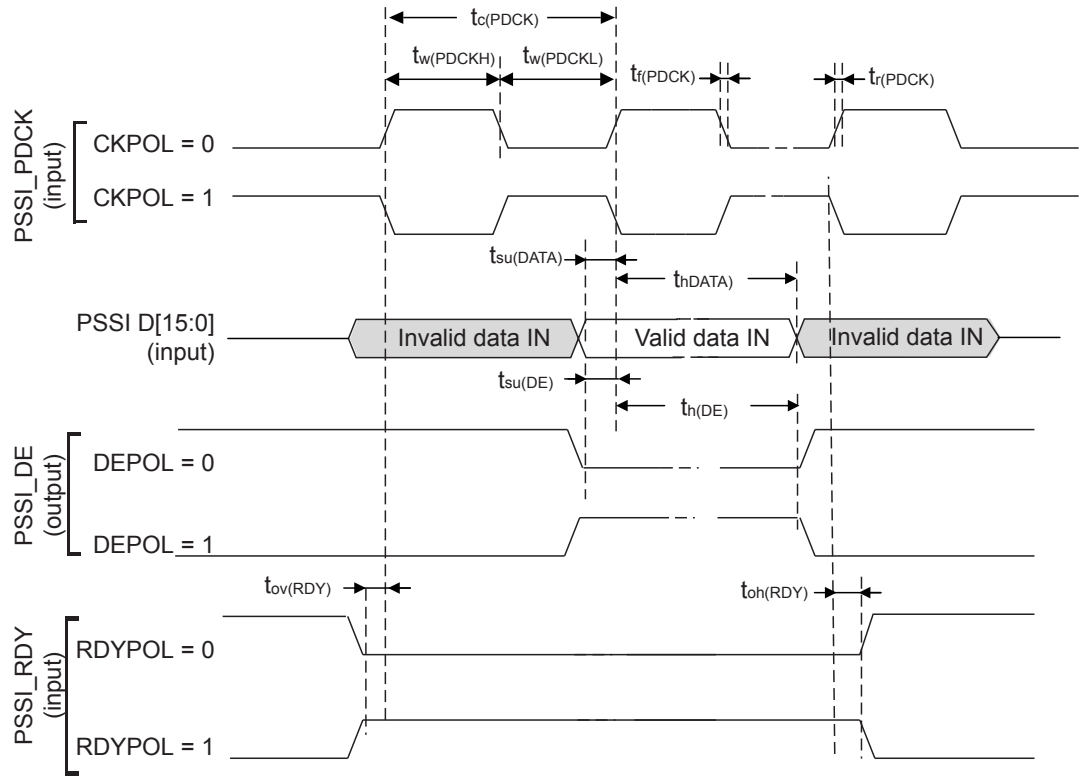


Figure 54. **PSSI timing diagram in Receive mode**



6.3.33 LCD-TFT controller (LTDC) characteristics

Unless otherwise specified, the parameters given in Table 110 for LCD-TFT are derived from tests performed under the ambient temperature, f_{HCLK} frequency and VDD supply voltage summarized in Table 20. General operating conditions and Section 6.3.1 , with the following configuration:

- LCD_CLK polarity: high
- LCD_DE polarity: low
- LCD_VSYNC and LCD_HSYNC polarity: high
- Pixel formats: 24 bits
- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C_L=30 pF
- Measurement points are done at CMOS levels: 0.5VDD
- IO Compensation cell activated.
- HSLV activated when $V_{DD} \leq 2.7$ V
- VOS level set to VOS 0

Note: At VOS1, the performance can be degraded by up to 5 % compared to VOS0. This is indicated by a footnote when applicable.

Table 110. LTDC characteristics

Symbol	Parameter	Conditions	Min	Max(1)	Unit
f_{CLK}	LTDC clock output frequency	2.7 V < V_{DD} < 3.6 V, 20 pF	-	140	MHz
		2.7 V < V_{DD} < 3.6 V	-	133	
		1.62 V < V_{DD} < 3.6 V	-	66.5	
D_{CLK}	LTDC clock output duty cycle	-	45	55	%
$t_{w(CLKH)}$, $t_{w(CLKL)}$	Clock High time, low time	-	$t_{w(CLK)}$/2−0.5	$t_{w(CLK)}$/2+0.5	ns
$t_{v(DATA)}$	Data output valid time	2.7 V < V_{DD} < 3.6 V	-	3.0	
		1.62 V < V_{DD} < 3.6 V	-	7.5	
$t_{h(DATA)}$	Data output hold time	-	0	-	
$t_{v(HSYNC)}$, $t_{v(VSYNC)}$, $t_{v(DE)}$	HSYNC/VSYNC/DE output valid time	2.7 V < V_{DD} < 3.6 V	-	3.0	
		1.62 V < V_{DD} < 3.6 V	-	7.5	
$t_{h(HSYNC)}$, $t_{h(VSYNC)}$, $t_{h(DE)}$	HSYNC/VSYNC/DE output hold time	-	0	-	

1. *At VOS1, these values are degraded by up to 5 %.*

Figure 55. **LCD-TFT horizontal timing diagram**

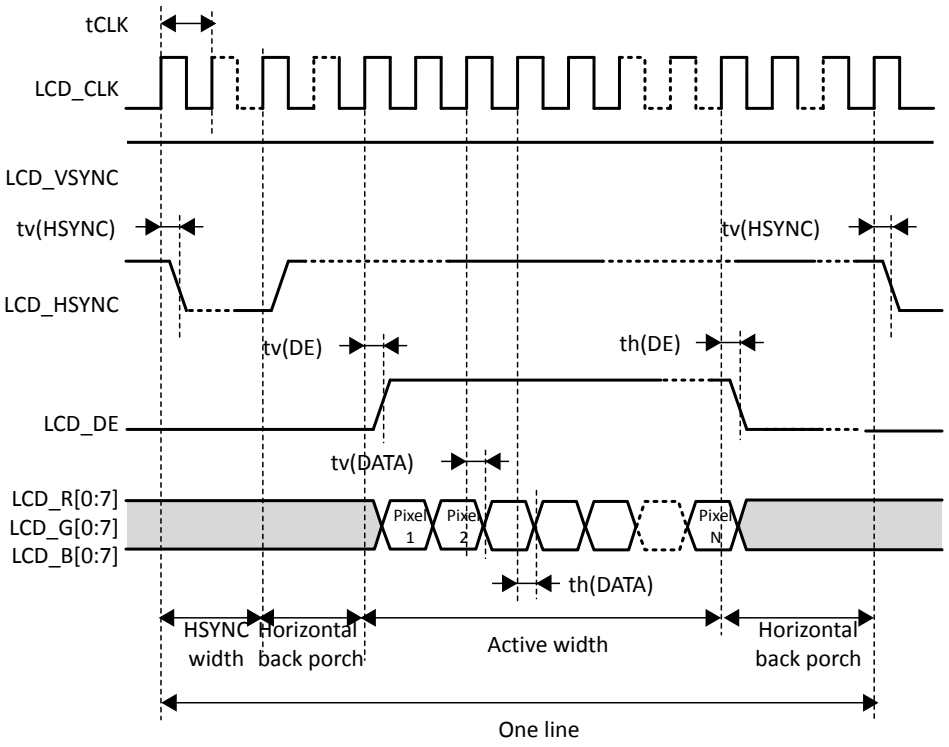
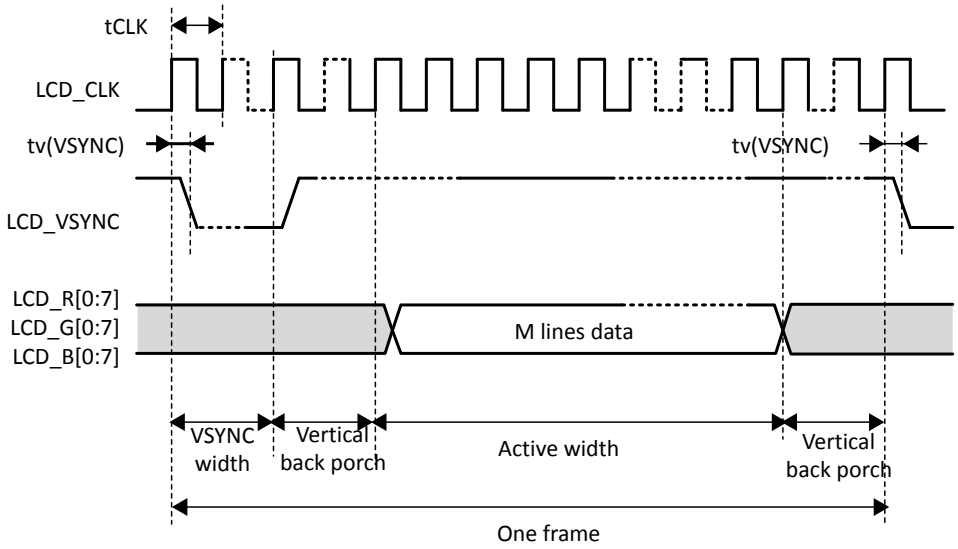


Figure 56. **LCD-TFT vertical timing diagram**



6.3.34 Timer characteristics

The parameters given in Table 111. TIMx characteristics are guaranteed by design.

Refer to Section 6.3.16 I/O port characteristics for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Table 111. TIMx characteristics

Symbol	Parameter	Conditions[1]	Min[2]	Max[2]	Unit
$t_{res(TIM)}$	Timer resolution time	AHB/APBx prescaler=1 or 2 or 4, $f_{TIMxCLK}$ = 280 MHz	1	-	$t_{TIMxCLK}$
		AHB/APBx prescaler>4, $f_{TIMxCLK}$ = 140 MHz	1	-	$t_{TIMxCLK}$
f_{EXT}	Timer external clock frequency on CH1 to CH4	$f_{TIMxCLK}$ = 280 MHz	0	$f_{TIMxCLK}/2$	MHz
Res_{TIM}	Timer resolution		-	16/32	bit
t_{MAX_COUNT}	Maximum possible count with 32-bit counter	-	-	65536 × 65536	$t_{TIMxCLK}$

1. *The maximum timer frequency on APB1 or APB2 is up to 280 MHz, by setting the TIMPRE bit in the RCC_CFGR register. If APBx prescaler is 1 or 2 or 4, then TIMxCLK = rcc_hclk1, otherwise TIMxCLK = 4x $F_{rcc_pclkx_d2}$.*
2. *Guaranteed by design.*

6.3.35 Low-power timer characteristics

Table 112. LPTIMx characteristics

Symbol	Parameter	Min	Max	Unit
$t_{res(TIM)}$	Timer resolution time	1	-	$t_{TIMxCLK}$
$f_{LPTIMxCLK}$	Timer kernel clock	0	100	MHz
f_{EXT}	Timer external clock frequency on Input1 and Input2	0	f_{LPTIM}xCLK/2	
Res_{TIM}	Timer resolution	-	16	bit
t_{MAX_COUNT}	Maximum possible count	-	65536	$t_{TIMxCLK}$

6.3.36 Communication interfaces

6.3.36.1 *I²C interface characteristics*

The I²C interface meets the timings requirements of the I2C-bus specification and user manual revision 03 for:

- Standard-mode (Sm): with a bit rate up to 100 kbit/s
- Fast-mode (Fm): with a bit rate up to 400 kbit/s
- Fast-mode Plus (Fm+): with a bit rate up to 1 Mbit/s.

The parameters given in Table 113 and Table 114are obtained with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 00

Note: At VOS1, the performance can be degraded by up to 5 % compared to VOS0. This is indicated by a footnote when applicable.

The I²C timings requirements are guaranteed by design when the I²C peripheral is properly configured (refer to RM0455 reference manual) and when the i2c_ker_ck frequency is greater than the minimum shown in the table below:

Table 113. Minimum i2c_ker_ck frequency in all I²C modes

Symbol	Parameter	Condition		Min	Unit
f_{I2CCLK}	I2CCLK frequency	Standard-mode	-	2	MHz
		Fast-mode	Analog Filtre ON, DNF=0	9	
			Analog Filtre OFF, DNF=1	9	
		Fast-mode Plus	Analog Filtre ON, DNF=0	19	

Symbol	Parameter	Condition		Min	Unit
f_{I2CCLK}	I2CCLK frequency	Fast-mode Plus	Analog Filtre OFF, DNF=1	16	-

The SDA and SCL I/O requirements are met with the following restrictions:

- The SDA and SCL I/O pins are not “true” open-drain. When configured as open-drain, the PMOS connected between the I/O pin and V_{DDIOx} is disabled, but still present.
- The 20 mA output drive requirement in Fast-mode Plus is not supported. This limits the maximum load C_{Load} supported in Fm+, which is given by these formulas:

$t_{r(SDA/SCL)}=0.8473xR_PxC_{Load}$

$R_{P(min)}= (V_{DD}-V_{OL(max)})/I_{OL(max)}$

Where R_P is the I2C lines pull-up. Refer to Section 6.3.16 I/O port characteristics for the I^2C I/Os characteristics.

All I^2C SDA and SCL I/Os embed an analog filter. Refer to the table below for the analog filter characteristics:

Table 114. I^2C analog filter characteristics

Symbol	Parameter	Min(1)	Max(1)	Unit
t_{AF}	Maximum pulse width of spikes that are suppressed by analog filter	50(2)	260(3)	ns

1. *Guaranteed by design.*
2. *Spikes whose width is lower than $t_{AF(min)}$ are filtered.*
3. *Spikes whose width is higher than $t_{AF(max)}$ are not filtered.*

6.3.36.2 *USART interface characteristics*

Unless otherwise specified, the parameters given in Table 115 for USART are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage conditions summarized in Table 20. General operating conditions and Section 6.3.1 , with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C_L = 30 pF
- Measurement points are done at CMOS levels: $0.5V_{DD}$
- IO Compensation cell activated.
- VOS level set to VOS0

Note: At VOS1, the performance can be degraded by up to 5 % compared to VOS0. This is indicated by a footnote when applicable.

Refer to Section 6.3.16 I/O port characteristics for more details on the input/output alternate function characteristics (NSS, CK, TX, RX for USART).

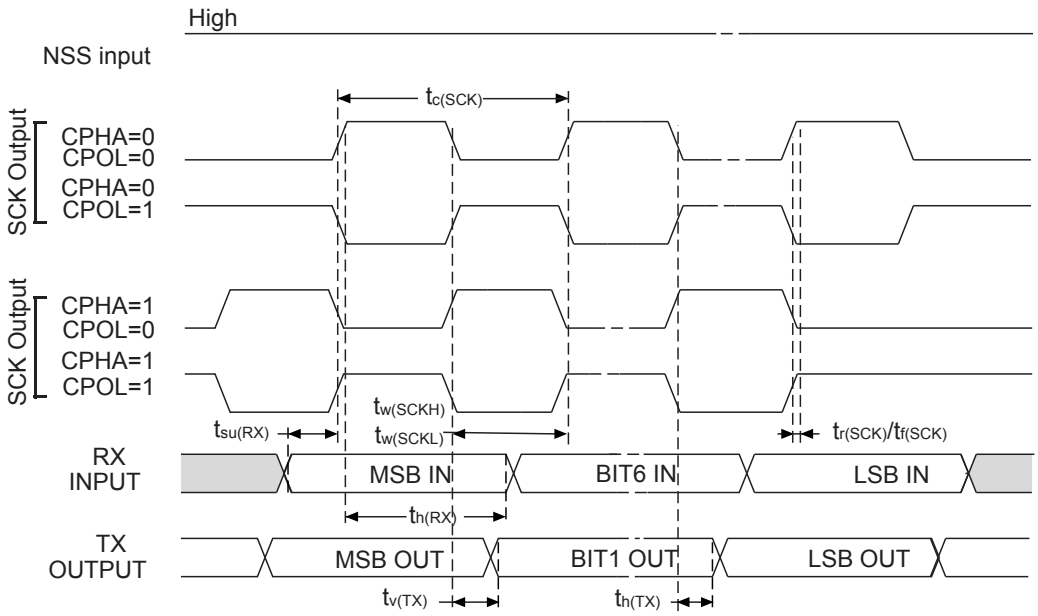
Table 115. USART characteristics

Symbol	Parameter	Conditions	Min	Typ	Max(1)	Unit
f_{CK}	USART clock frequency	Master mode	-	-	35	MHz
		Slave receiver mode			93.0	
		Slave mode transmitter mode, 2.7 V < V_{DD} < 3.6 V			29.0	
		Slave mode transmitter mode, 1.62 V < V_{DD} < 3.6 V			22.0	
$t_{su(NSS)}$	NSS setup time	Slave mode	t_{ker}+2	-	-	-
$t_{h(NSS)}$	NSS hold time	Slave mode	2	-	-	
$t_{w(SCKH)}$, $t_{w(SCKL)}$	CK high and low time	Master mode	$1/f_{ck}/2-2$	$1/f_{ck}/2$	$1/f_{ck}/2+2$	
$t_{su(RX)}$	Data input setup time	Master mode	17	-	-	ns
		Slave mode	1	-	-	

Symbol	Parameter	Conditions	Min	Typ	Max(1)	Unit
$t_{h(RX)}$	Data input hold time	Master mode	0	-	-	ns
		Slave mode	1.5	-	-	
$t_{v(TX)}$	Data output valid time	Slave mode transmitter mode, 1.62 V < V_{DD} < 3.6 V	-	15.5	22	
		Slave mode transmitter mode, 2.7 V < V_{DD} < 3.6 V	-	15.5	17	
		Master mode	-	1.5	2	
$t_{h(TX)}$	Data output hold time	Slave mode	12	-	-	
		Master mode	1	-	-	

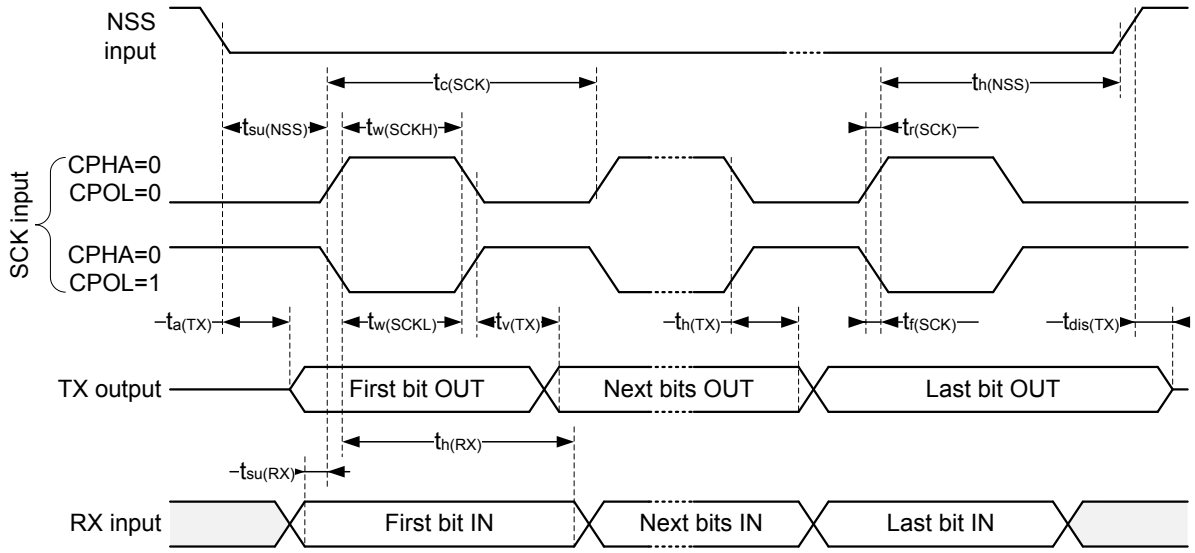
1. *At VOS1, these values are degraded by up to 5 %.*

Figure 57. USART timing diagram in Master mode



1. Measurement points are done at 0.5V_{DD} and with external C_L = 30 pF.

Figure 58. **USART timing diagram in Slave mode**



6.3.36.3 *SPI interface characteristics*

Unless otherwise specified, the parameters given in Table 116 for SPI are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage conditions summarized in Table 20. General operating conditions and Section 6.3.1 , with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C_L = 30 pF
- Measurement points are done at CMOS levels: $0.5V_{DD}$
- IO Compensation cell activated.
- HSLV activated when $V_{DD} \leq 2.7$ V
- VOS level set to VOS0

Note: At VOS1, the performance can be degraded by up to 5 % compared to VOS0. This is indicated by a footnote when applicable.

Refer to Section 6.3.16 I/O port characteristics for more details on the input/output alternate function characteristics (SS, SCK, MOSI, MISO for SPI).

Table 116. **SPI dynamic characteristics**

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)(2)	Unit
f_{SCK}	SPI clock frequency	Master mode 2.7 < V_{DD}< 3.6 V, SPI1, 2, 3	-	-	125/100(3)	MHz
		Master mode, 2.7 < V_{DD}<3.6 V, SPI4, 5, 6			100	
		Master mode, 1.62 < V_{DD} < 3.6 V, SPI4, 5, 6			75/38(3)	
		Slave receiver mode, 1.62 < V_{DD} < 3.6 V			100	
		Slave mode transmitter/full duplex, 2.7 < V_{DD} < 3.6 V			45/31(3)	
		Slave mode transmitter/full duplex, 1.62 <V_{DD} < 3.6 V			29/18(3)	
$t_{su(SS)}$	SS setup time	Slave mode	2	-	-	ns
$t_{h(SS)}$	SS hold time	Slave mode	1	-	-	
$t_{su(MI)}$	Data input setup time	Master mode	3	-	-	ns

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)(2)	Unit
$t_{su(SI)}$	Data input setup time	Slave mode	2	-	-	ns
$t_{h(MI)}$	Data input hold time	Master mode	3	-	-	
$t_{h(SI)}$		Slave mode	1	-	-	
$t_{a(SO)}$	Data output access time	Slave mode	9	13	27	
$t_{dis(SO)}$	Data output disable time	Slave mode	0	1	5	
$t_{v(SO)}$	Data output valid time	Slave mode, 2.7 < V_{DD} < 3.6 V	-	9/15(3)	11/16(3)	
		Slave mode, 1.62 < V_{DD} < 3.6 V	-	9/15(3)	17/27(3)	
$t_{v(MO)}$		Master mode, 2.7 < V_{DD} < 3.6 V	-	1/5(3)	1.5/7(3)	
		Master mode, 1.62 < V_{DD} < 3.6 V	-	1/5(3)	2/13(3)	
$t_{h(SO)}$	Data output hold time	Slave mode, 1.62 < V_{DD} < 3.6 V	7	-	-	
$t_{h(MO)}$		Master mode	0	-	-	

1. *Guaranteed by characterization results.*
2. *At VOS1, these values are degraded by up to 5 %.*
3. *Using PC3_C / PC2_C (not available on all packages).*

Figure 59. SPI timing diagram - slave mode and CPHA = 0

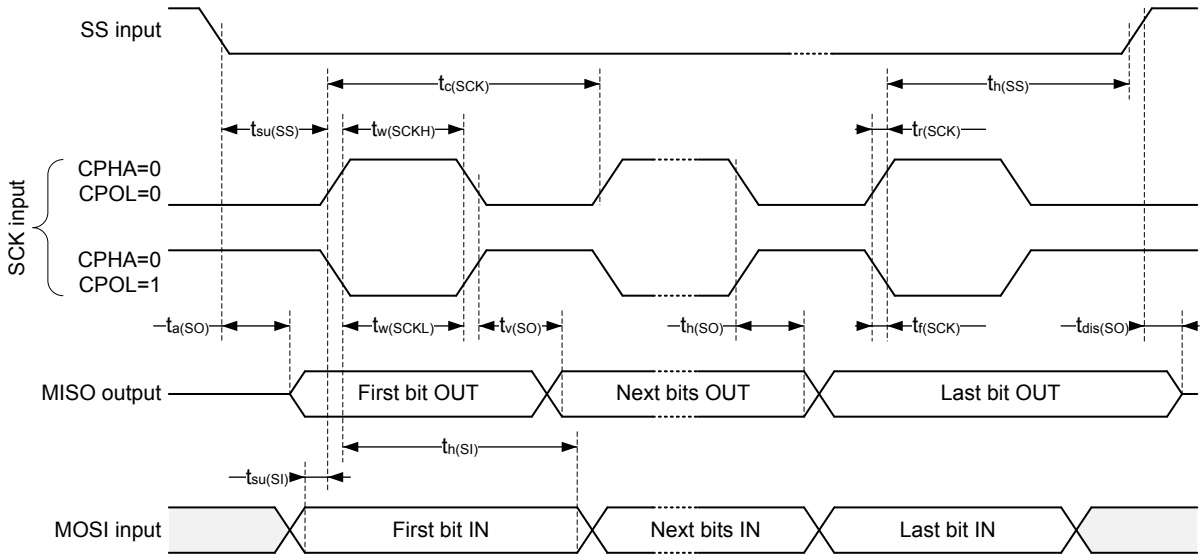
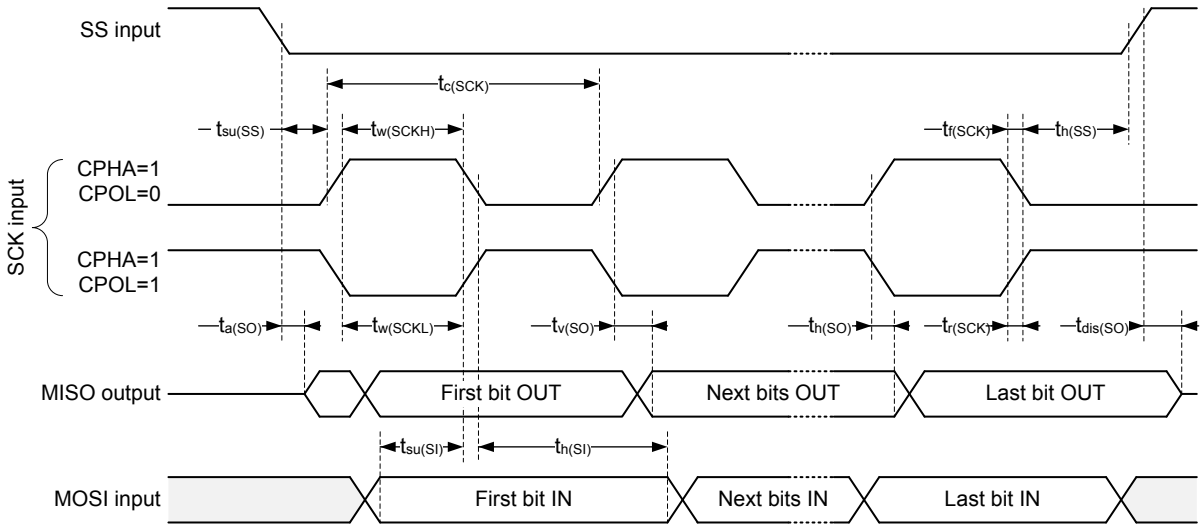
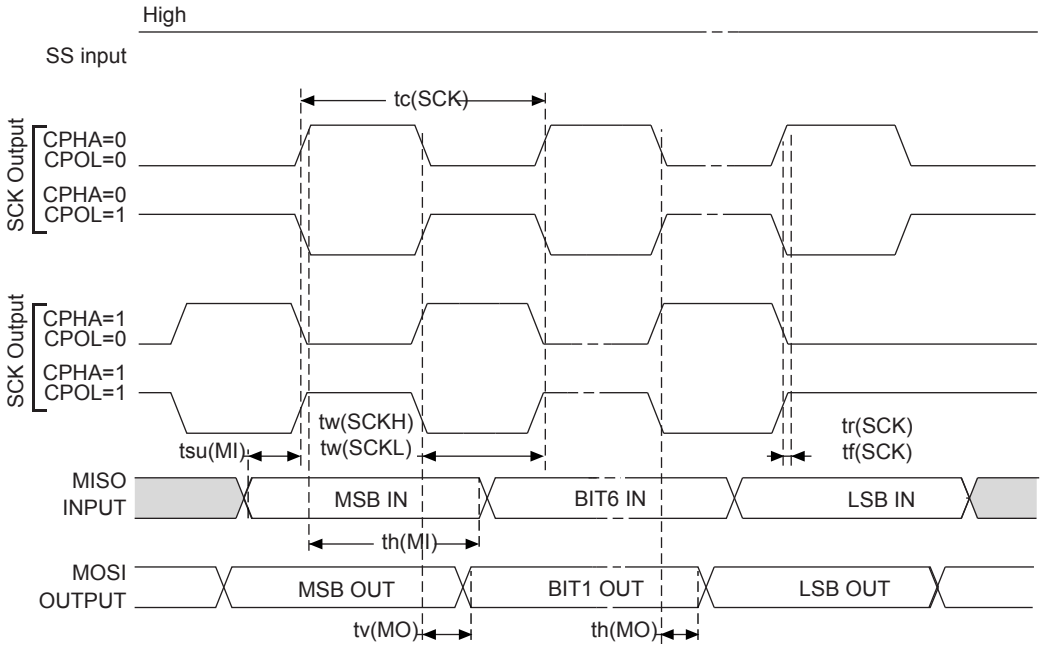


Figure 60. SPI timing diagram - slave mode and CPHA = 1(1)



1. Measurement points are done at $0.5V_{DD}$ and with external C_L = 30 pF.

Figure 61. SPI timing diagram - master mode(1)



1. Measurement points are done at $0.5V_{DD}$ and with external C_L = 30 pF.

6.3.36.4 *I²S Interface characteristics*

Unless otherwise specified, the parameters given in Table 117 for I²S are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage conditions summarized in Table 20. General operating conditions and Section 6.3.1 , with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10

- Capacitive load C_L = 30 pF
- Measurement points are done at CMOS levels: $0.5V_{DD}$
- IO Compensation cell activated.
- HSLV activated when VDD ≤ 2.7 V
- VOS level set to VOS0

Note: At VOS1, the performance can be degraded by up to 5 % compared to VOS0. This is indicated by a footnote when applicable.

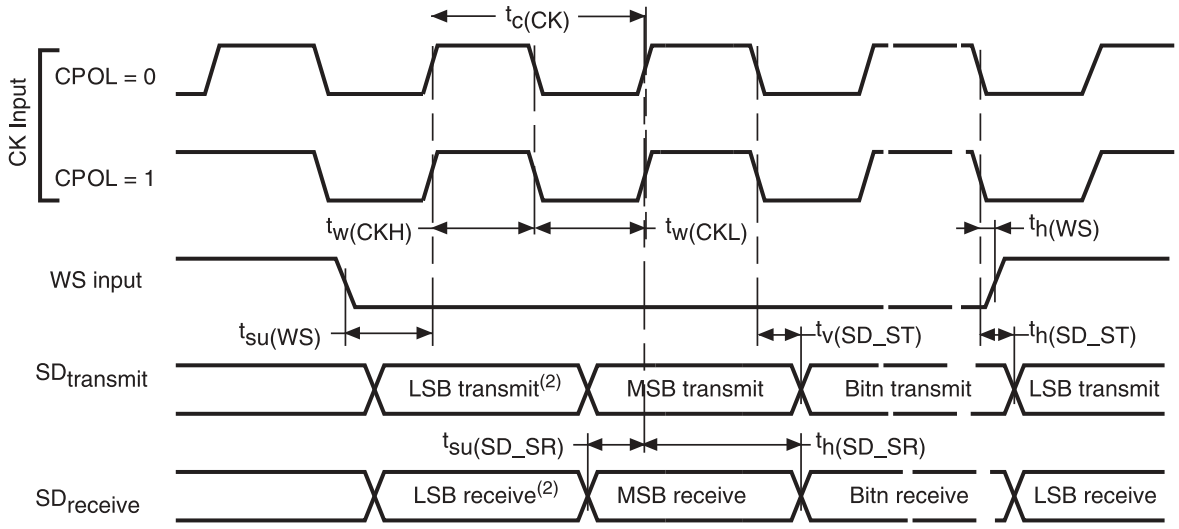
Refer to Section 6.3.16 I/O port characteristics for more details on the input/output alternate function characteristics (CK,SD,WS).

Table 117. I²S dynamic characteristics

Symbol	Parameter	Conditions	Min(1)	Max(1)(2)	Unit
f_{MCK}	I²S main clock output	-	-	50	MHz
f_{CK}	I²S clock frequency	Master Tx	-	50/33(3)	MHz
		Master Rx	-	40	
		Slave Tx	-	31/18.5(3)	
		Slave Rx	-	50	
$t_{v(WS)}$	WS valid time	Master mode	-	5.5	ns
$t_{h(WS)}$	WS hold time	Master mode	0	-	
$t_{su(WS)}$	WS setup time	Slave mode	2	-	
$t_{h(WS)}$	WS hold time	Slave mode	1	-	
$t_{su(SD_MR)}$	Data input setup time	Master receiver	2	-	
$t_{su(SD_SR)}$		Slave receiver	2	-	
$t_{h(SD_MR)}$	Data input hold time	Master receiver	4.5	-	
$t_{h(SD_SR)}$		Slave receiver	1	-	
$t_{v(SD_ST)}$	Data output valid time	Slave transmitter (after enable edge)	-	16/27(3)	
$t_{v(SD_MT)}$		Master transmitter (after enable edge)	-	4/15(3)	
$t_{h(SD_ST)}$	Data output hold time	Slave transmitter (after enable edge)	7	-	
$t_{h(SD_MT)}$		Master transmitter (after enable edge)	0	-	

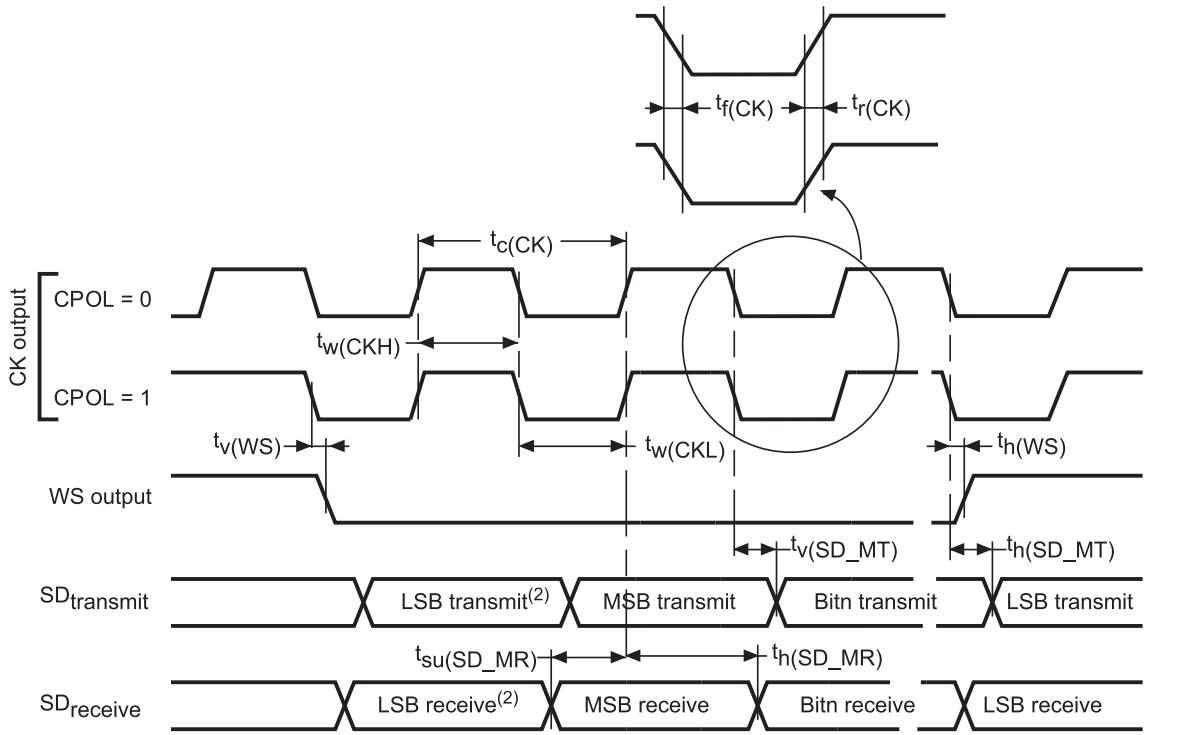
1. *Guaranteed by characterization results.*
2. *At VOS1, these values are degraded by up to 5 %.*
3. *Using PC3_C / PC2_C (not available on all packages).*

Figure 62. I²S slave timing diagram (Philips protocol)(1)



1. LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

Figure 63. I²S master timing diagram (Philips protocol)(1)



1. LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

6.3.36.5 *SAI characteristics*

Unless otherwise specified, the parameters given in Table 118 for SAI are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and VDD supply voltage conditions summarized in Table 20. General operating conditions and Section 6.3.1 , with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C_L = 30 pF
- IO Compensation cell activated.
- Measurement points are done at CMOS levels: 0.5VDD
- VOS level set to VOS0

Note: At VOS1, the performance can be degraded by up to 5 % compared to VOS0. This is indicated by a footnote when applicable.

Refer to Section 6.3.16 I/O port characteristics for more details on the input/output alternate function characteristics (SCK,SD,WS).

Table 118. SAI characteristics

Symbol	Parameter	Conditions	Min(1)	Max(1)(2)	Unit
f_{MCK}	SAI Main clock output	-	-	50	MHz
f_{CK}	SAI clock frequency	Master transmitter, 2.7 ≤ V_{DD} ≤ 3.6 V	-	34	
		Master transmitter, 1.62 ≤ V_{DD} ≤ 3.6 V	-	27	
		Master receiver, 1.6 ≤ V_{DD} ≤ 3.6 V	-	27	
		Slave transmitter, 2.7 ≤ VDD ≤ 3.6 V	-	37	
		Slave transmitter, 1.62 ≤ VDD ≤ 3.6 V	-	30	
		Slave receiver, 1.62 ≤ V_{DD}≤ 3.6 V	-	50	
$t_{v(FS)}$	F_S valid time	Master mode,2.7 ≤ V_{DD} ≤ 3.6 V	-	14.5	ns
		Master mode, 1.62 ≤ V_{DD} ≤ 3.6 V	-	18.5	
$t_{su(FS)}$	F_S setup time	Slave mode	8	-	
$t_{h(FS)}$	F_S hold time	Master mode	1	-	
	F_S hold time	Slave mode	2	-	
$t_{su(SD_A_MR)}$	Data input setup time	Master receiver	0.5	-	
$t_{su(SD_B_SR)}$		Slave receiver	1	-	
$t_{h(SD_A_MR)}$	Data input hold time	Master receiver	5.5	-	
$t_{h(SD_B_SR)}$		Slave receiver	3	-	
$t_{v(SD_B_ST)}$	Data output valid time	Slave transmitter (after enable edge), 2.7 ≤ V_{DD} ≤ 3.6 V	-	13.5	
		Slave transmitter (after enable edge), 1.62 ≤ V_{DD} ≤ 3.6 V	-	16.5	
$t_{h(SD_B_ST)}$	Data output hold time	Slave transmitter (after enable edge)	8	-	
$t_{v(SD_A_MT)}$	Data output valid time	Master transmitter (after enable edge), 2.7 ≤ V_{DD} ≤ 3.6 V	-	14	
		Master transmitter (after enable edge), 1.62 ≤ V_{DD} ≤ 3.6 V	-	18	
$t_{h(SD_A_MT)}$	Data output hold time	Master transmitter (after enable edge)	7.5	-	

1. *Guaranteed by characterization results.*
2. *At VOS1, these values are degraded by up to 5 %.*
3. *APB clock frequency must be at least twice SAI clock frequency.*

Figure 64. **SAI master timing waveforms**

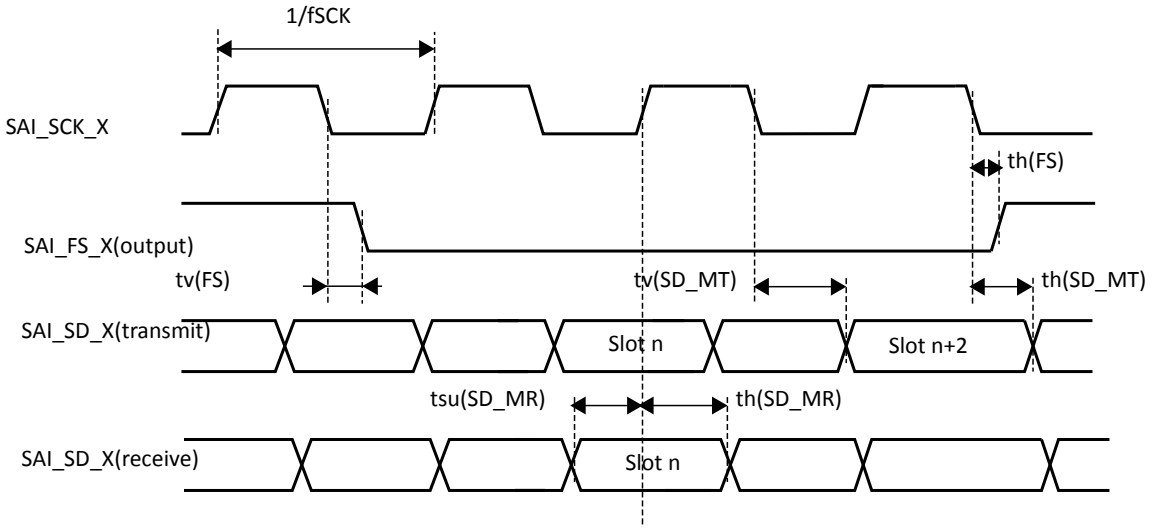
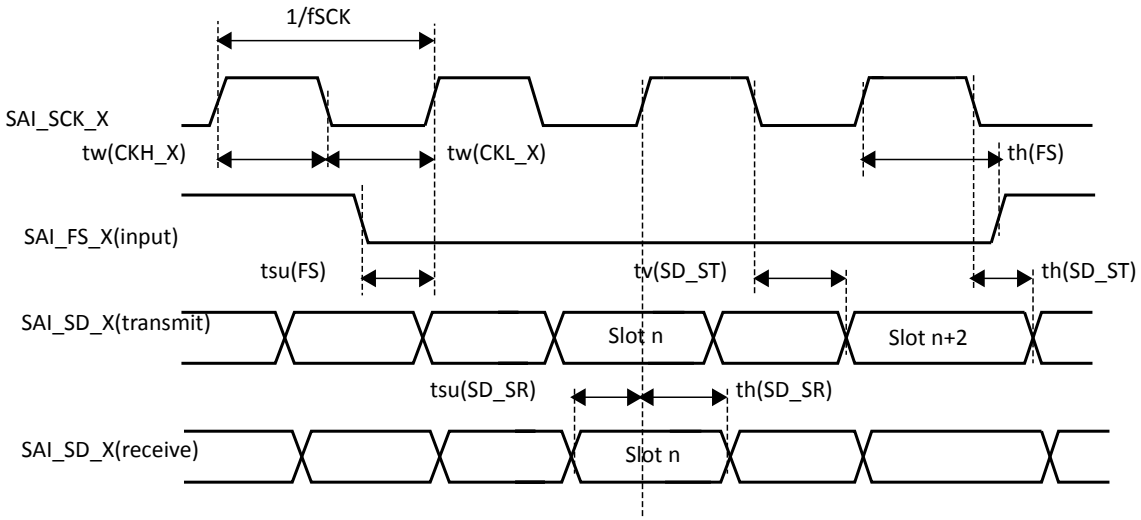


Figure 65. **SAI slave timing waveforms**



6.3.36.6 *MDIO characteristics*

Unless otherwise specified, the parameters given in Table 119 are derived from tests performed under the ambient temperature, f_{HCLK} frequency and V_{DD} supply voltage summarized in Table 20. General operating conditions, with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Measurement points are done at CMOS levels: $0.5V_{DD}$
- I/O compensation cell activated.
- HSLV activated when $V_{DD} \leq 2.7$ V
- VOS level set to VOS0

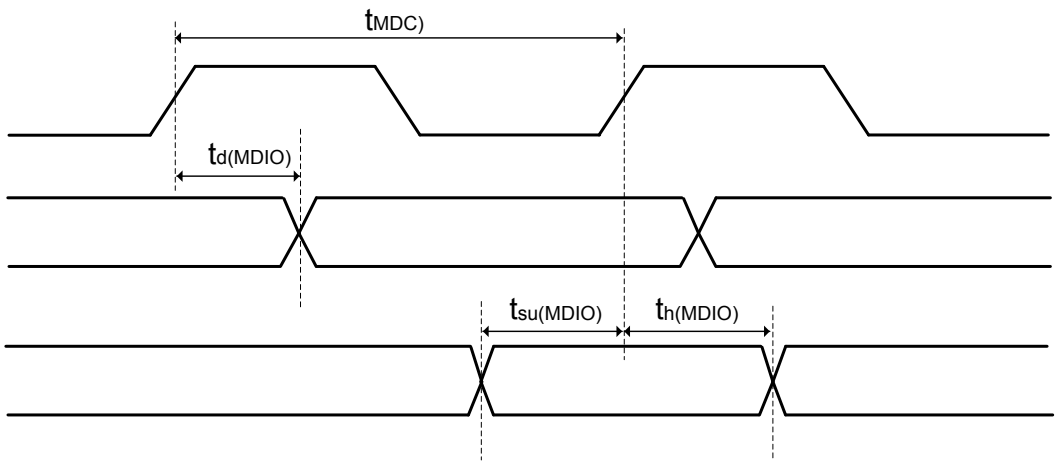
Note: At VOS1, the performance can be degraded by up to 5 % compared to VOS0. This is indicated by a footnote when applicable.

Table 119. MDIO Slave timing parameters

Symbol	Parameter	Min	Typ	Max(1)	Unit
F_{MDC}	Management Data Clock	-	-	30	MHz
$t_{d(MDIO)}$	Management Data Iput/output output valid time	9	11	21	ns
$t_{su(MDIO)}$	Management Data Iput/output setup time	2.5	-	-	
$t_{h(MDIO)}$	Management Data Iput/output hold time	1	-	-	

1. *At VOS1, these values are degraded by up to 5 %.*

Figure 66. MDIO Slave timing diagram



6.3.36.7 *SD/SDIO MMC card host interface (SDMMC) characteristics*

Unless otherwise specified, the parameters given in Table 120 and Table 121 for SDIO are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage summarized in Table 20. General operating conditions and Section 6.3.1 , with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C_L=30 pF
- Measurement points are done at CMOS levels: $0.5V_{DD}$
- IO Compensation cell activated.
- HSLV activated when $V_{DD} \leq 2.7$ V
- VOS level set to VOS0

Note: At VOS1, the performance can be degraded by up to 5 % compared to VOS0. This is indicated by a footnote when applicable.

Refer to Section 6.3.16 I/O port characteristics for more details on the input/output characteristics.

Table 120. Dynamics characteristics: SDMMC characteristics, V_{DD}=2.7 to 3.6 V

Above 100 MHz, C_L = 20 pF.

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)(2)	Unit
f_{PP}	Clock frequency in data transfer mode	-	0	-	133	MHz
-	SDIO_CK/f_{PCLK2} frequency ratio	-	-	-	8/3	-
$t_{W(CKL)}$	Clock low time	f_{PP} =52 MHz	8.5	9.5	-	ns
$t_{W(CKH)}$	Clock high time	f_{PP} =52 MHz	8.5	9.5	-	
CMD, D inputs (referenced to CK) in eMMC legacy/SDR/DDR and SD HS/SDR/DDR mode						

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Max(1)(2)	Unit
t_{ISU}	Input setup time HS	-	2.5	-	-	ns
t_{IH}	Input hold time HS	-	0.5	-	-	
t_{IDW}(3)	Input valid window (variable window)	-	3.0	-	-	-
CMD, D outputs (referenced to CK) in eMMC legacy/SDR/DDR and SD HS/SDR/DDR mode						
t_{OV}	Output valid time HS	-	-	6	6.5	ns
t_{OH}	Output hold time HS	-	5	-	-	
CMD, D inputs (referenced to CK) in SD default mode						
t_{ISUD}	Input setup time SD	-	2.5	-	-	ns
t_{IHD}	Input hold time SD	-	0.5	-	-	
CMD, D outputs (referenced to CK) in SD default mode						
t_{OVD}	Output valid default time SD	-	-	1	1.5	ns
t_{OHD}	Output hold default time SD	-	0	-	-	

1. *Guaranteed by characterization results.*
2. *At VOS1, these values are degraded by up to 5 %.*
3. *The minimum window of time where the data needs to be stable for proper sampling in tuning mode.*

Table 121. Dynamics characteristics: eMMC characteristics VDD=1.71V to 1.9V

Above 100 MHz, C_L = 20 pF.

Symbol	Parameter	Conditions	Min(1)	Typ(1)	Ma(1)(2)	Unit
f_{PP}	Clock frequency in data transfer mode	-	0	-	85	MHz
$t_{W(CKL)}$	Clock low time	f_{PP} =52 MHz	8.5	9.5	-	ns
$t_{W(CKH)}$	Clock high time	f_{PP} =52 MHz	8.5	9.5	-	
CMD, D inputs (referenced to CK) in eMMC mode						
t_{ISU}	Input setup time HS	-	2.5	-	-	ns
t_{IH}	Input hold time HS	-	0.5	-	-	
t_{IDW}(3)	Input valid window (variable window)	-	3.5	-	-	
CMD, D outputs (referenced to CK) in eMMC mode						
t_{OVD}	Output valid time HS	-	-	6	6.5	ns
t_{OHD}	Output hold time HS	-	5.5	-	-	

1. *Guaranteed by characterization results.*
2. *At VOS1, these values are degraded by up to 5 %.*
3. *The minimum window of time where the data needs to be stable for proper sampling in tuning mode.*

Figure 67. **SDIO high-speed mode**

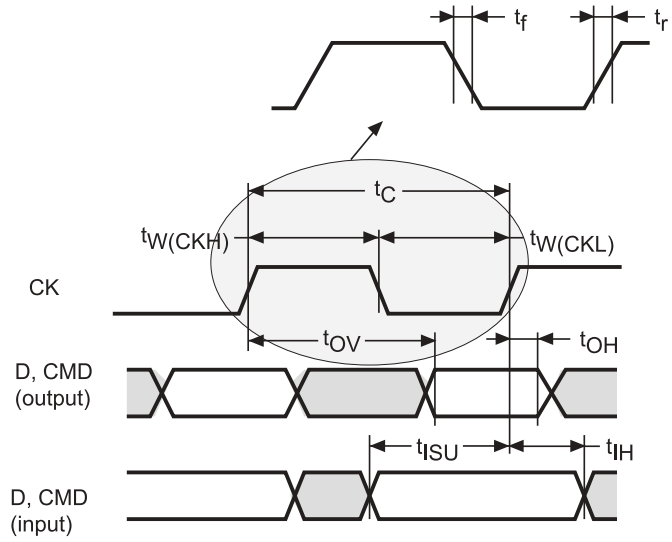


Figure 68. **SD default mode**

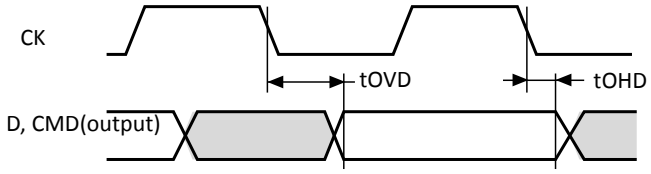
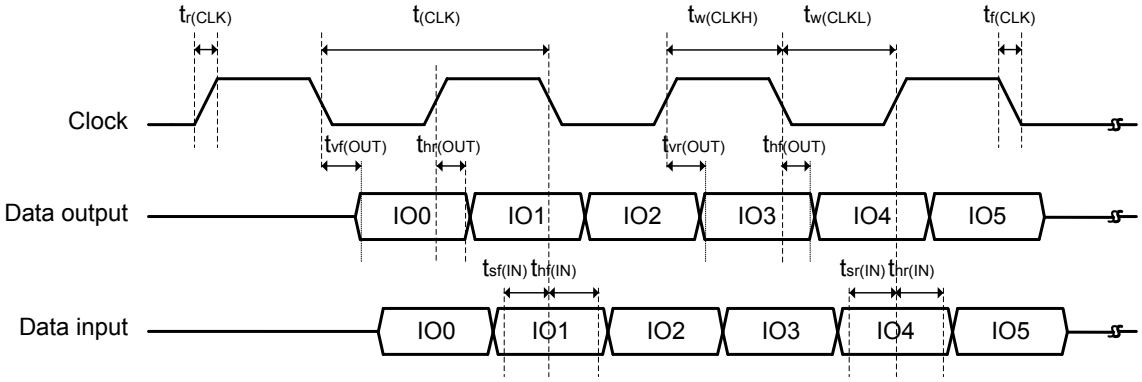


Figure 69. **DDR mode**



6.3.36.8 *USB OTG_FS characteristics*

Unless otherwise specified, the parameters given in Table 122. Dynamics characteristics: USB OTG_FS for ULPI are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage summarized in Table 20. General operating conditions and Section 6.3.1 , with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C_L=20 pF

- Measurement points are done at CMOS levels: $0.5V_{DD}$
- IO Compensation cell activated.
- VOS level set to VOS0

Refer to Section 6.3.16 I/O port characteristics for more details on the input/output characteristics.

Table 122. Dynamics characteristics: USB OTG_FS

Symbol	Parameter	Condition	Min	Typ	Max	Unit
$V_{DD33USB}$	USB transceiver operating voltage	-	3.0(1)	-	3.6	V
R_{PUI}	Embedded USB_DP pull-up value during idle	-	900	1250	1600	Ω
R_{PUR}	Embedded USB_DP pull-up value during reception	-	1400	2300	3200	
Z_{DRV}	Output driver impedance(2)	Driver high and low	28	36	44	

1. *The USB functionality is ensured down to 2.7 V but not the full USB electrical characteristics that are degraded in the 2.7 to 3.0 V voltage range.*
2. *No external termination series resistors are required on USB_DP (D+) and USB_DM (D-); the matching impedance is already included in the embedded driver.*

6.3.36.9 *USB OTG_HS characteristics*

Unless otherwise specified, the parameters given in Table 123 for ULPI are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage summarized in Table 20. General operating conditions and Section 6.3.1 , with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C_L=20 pF
- Measurement points are done at CMOS levels: $0.5V_{DD}$
- IO Compensation cell activated.
- VOS level set to VOS0

Note: At VOS1, the performance can be degraded by up to 5 % compared to VOS0. This is indicated by a footnote when applicable.

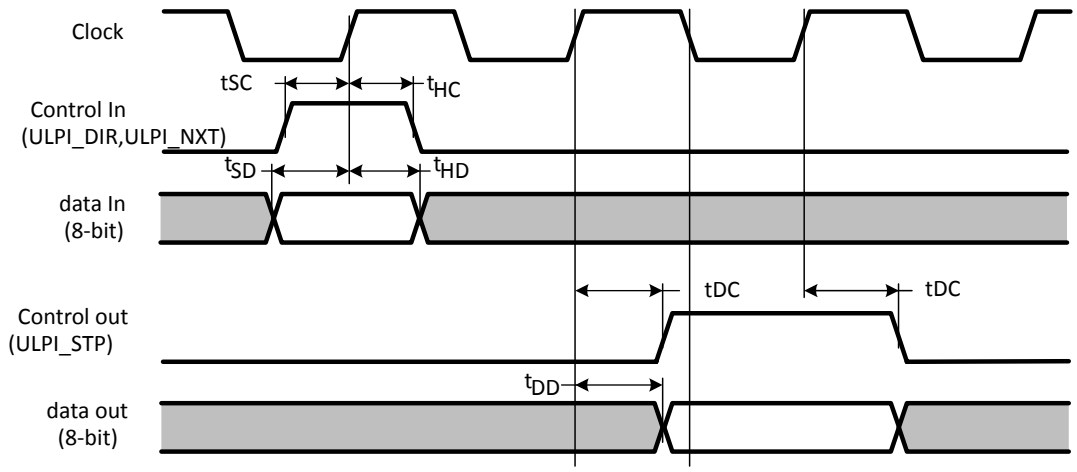
Refer to Section 6.3.16 I/O port characteristics for more details on the input/output characteristics.

Table 123. Dynamics characteristics: USB ULPI

Symbol	Parameter	Condition	Min(1)	Typ(1)	Max(1)(2)(3)	Unit
t_{SC}	Control in (ULPI_DIR , ULPI_NXT) setup time	-	3.5	-	-	ns
t_{HC}	Control in (ULPI_DIR, ULPI_NXT) hold time	-	2	-	-	
t_{SD}	Data in setup time	-	3	-	-	
t_{HD}	Data in hold time	-	0	-	-	
t_{DC}/t_{DD}	Control/Datal output delay	$2.7 < V_{DD} < 3.6$ V, C_L=20 pF	-	7	8.5	
		$1.71 < V_{DD} < 3.6$ V, C_L=15 pF	-	9	13	

1. *Guaranteed by characterization results.*
2. *At VOS1, these values are degraded by up to 5 %.*
3. *For external ULPI transceivers operating at 1.8 V, check carefully the timing values for compatibility.*

Figure 70. **ULPI timing diagram**



6.3.36.10 *JTAG/SWD interface characteristics*

Unless otherwise specified, the parameters given in Table 124 and Table 125 for JTAG/SWD are derived from tests performed under the ambient temperature, $f_{rcc_cpu_ck}$ frequency and V_{DD} supply voltage summarized in Table 20. General operating conditions and Section 6.3.1 , with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C_L=30 pF
- Measurement points are done at CMOS levels: $0.5V_{DD}$
- VOS level set to VOS0

Note: At VOS1, the performance can be degraded by up to 5 % compared to VOS0. This is indicated by a footnote when applicable.

Refer to Section 6.3.16 I/O port characteristics for more details on the input/output characteristics:

Table 124. **Dynamics JTAG characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max(1)	Unit
F_{pp}	T_{CK} clock frequency	2.7 V <V_{DD}< 3.6 V	-	-	35	MHz
$1/t_{c(TCK)}$		1.62 V <V_{DD}< 3.6 V	-	-	27.5	ns
$t_{isu(TMS)}$	TMS input setup time	-	1	-	-	
$t_{ih(TMS)}$	TMS input hold time	-	1	-	-	
$t_{isu(TDI)}$	TDI input setup time	-	1.5	-	-	
$t_{ih(TDI)}$	TDI input hold time	-	1	-	-	
$t_{ov(TDO)}$	TDO output valid time	2.7 V <V_{DD}< 3.6 V	-	8	14	
		1.62 V <V_{DD}< 3.6 V	-	8	18	
$t_{oh(TDO)}$	TDO output hold time	-	7	-	-	

1. *At VOS1, these values are degraded by up to 5 %.*

Table 125. **Dynamics SWD characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max(1)	Unit
F_{pp}	SWCLK clock frequency	2.7V <V_{DD}< 3.6 V	-	-	76	MHz
$1/t_{c(SWCLK)}$		1.62 <V_{DD}< 3.6 V	-	-	55.5	

Symbol	Parameter	Conditions	Min	Typ	Max(1)	Unit
$t_{isu(SWDIO)}$	SWDIO input setup time	-	2	-	-	ns
$t_{ih(SWDIO)}$	SWDIO input hold time	-	1	-	-	
$t_{ov(SWDIO)}$	SWDIO output valid time	2.7V <V_{DD}< 3.6 V	-	8.5	13	
		1.62 <V_{DD}< 3.6 V	-	8.5	18	
$t_{oh(SWDIO)}$	SWDIO output hold	-	8	-	-	

1. *At VOS1, these values are degraded by up to 5 %.*

Figure 71. JTAG timing diagram

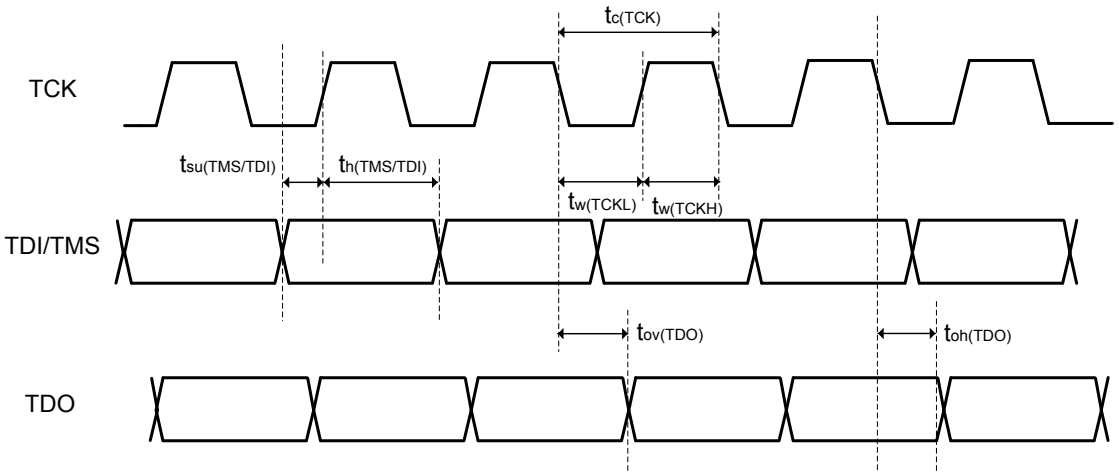
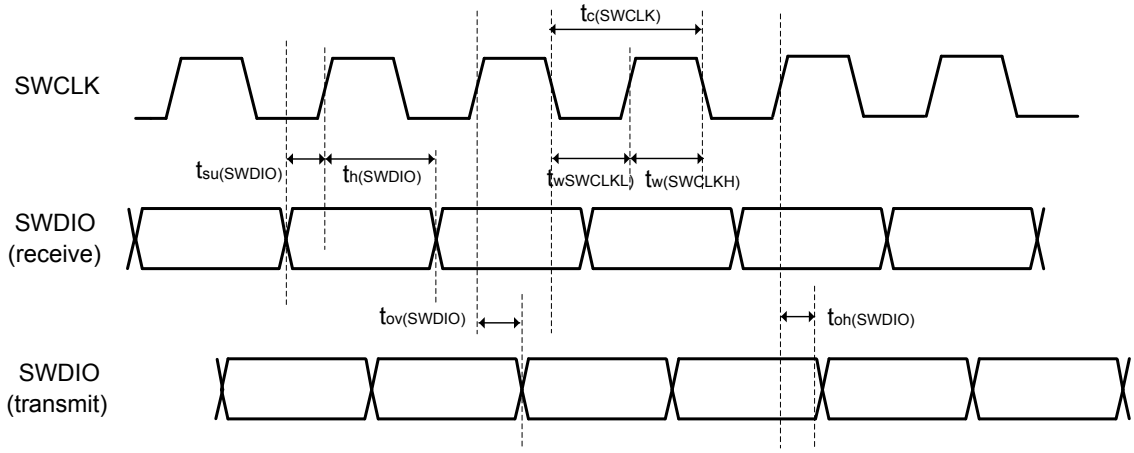


Figure 72. SWD timing diagram



7 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: www.st.com. ECOPACK is an ST trademark.

7.1 LQFP64 package information

This is a 64-pins, 10 x 10 mm, low-profile quad flat package.

Note: See list of notes in the notes section.

Figure 73. LQFP64 - Outline(15.)

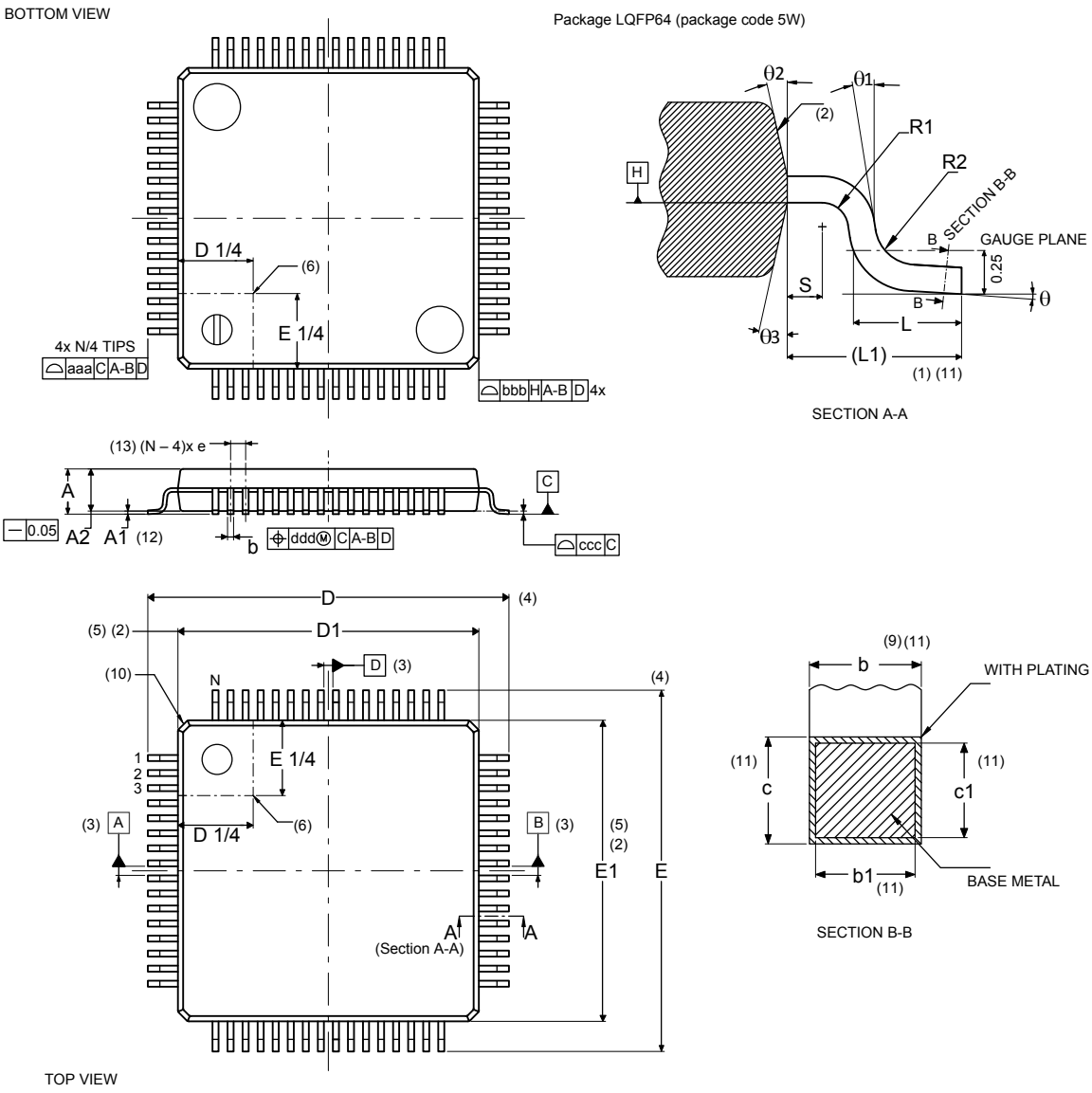


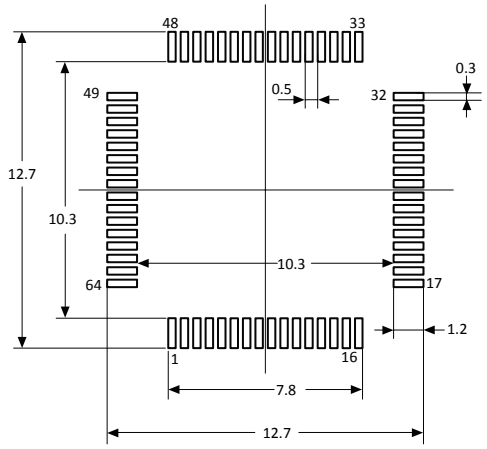
Table 126. **LQFP64 - Mechanical data**

Symbol	millimeters			inches[14.]		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.60	-	-	0.0630
A1[12.]	0.05	-	0.15	0.0020	-	0.0059
A2	1.35	1.40	1.45	0.0531	0.0551	0.0571
b[9.][11.]	0.17	0.22	0.27	0.0067	0.0087	0.0106
b1[11.]	0.17	0.20	0.23	00067	0.0079	0.0091
c[11.]	0.09	-	0.20	0.0035	-	0.0079
c1[11.]	0.09	-	0.16	0.0035	-	0.0063
D[4.]	12.00 BSC			0.4724 BSC		
D1[2.][5.]	10.00 BSC			0.3937 BSC		
E[4.]	12.00 BSC			0.4724 BSC		
E1[2.][5.]	10.00 BSC			0.3937 BSC		
e	0.500 BSC			0.0197 BSC		
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
N[13.]	64					
Θ	0°	3.5°	7°	0°	3.5°	7°
Θ1	0°	-	-	0°	-	-
Θ2	10°	12°	14°	10°	12°	14°
Θ3	10°	12°	14°	10°	12°	14°
R1	0.08	-	-	0.0031	-	-
R2	0.08	-	0.20	0.0031	-	0.0079
S	0.20	-	-	0.0079	-	-
aaa[1.]	0.20			0.0079		
bbb[1.]	0.20			0.0079		
ccc[1.]	0.08			0.0031		
ddd[1.]	0.08			0.0031		

Notes

1. Dimensioning and tolerancing schemes conform to ASME Y14.5M-1994.
2. The top package body size may be smaller than the bottom package size by as much as 0.15 mm.
3. Datums A-B and D to be determined at datum plane H.
4. To be determined at seating datum plane C.
5. Dimensions D1and E1 do not include mold flash or protrusions. Allowable mold flash or protrusions is “0.25 mm” per side. D1 and E1 are Maximum plastic body size dimensions including mold mismatch.
6. Details of pin 1 identifier are optional but must be located within the zone indicated.
7. All dimensions are in millimeters.
8. No intrusion allowed inwards the leads.
9. Dimension “b” does not include dambar protrusion. Allowable dambar protrusion shall not cause the lead width to exceed the maximum “b” dimension by more than 0.08 mm. Dambar cannot be located on the lower radius or the foot. Minimum space between protrusion and an adjacent lead is 0.07 mm for 0.4 mm and 0.5 mm pitch packages.
10. Exact shape of each corner is optional.
11. These dimensions apply to the flat section of the lead between 0.10 mm and 0.25 mm from the lead tip.
12. A1 is defined as the distance from the seating plane to the lowest point on the package body.
13. “N” is the number of terminal positions for the specified body size.
14. Values in inches are converted from mm and rounded to 4 decimal digits.
15. Drawing is not to scale.

Figure 74. LQFP64 - Recommended footprint



1. Dimensions are expressed in millimeters.

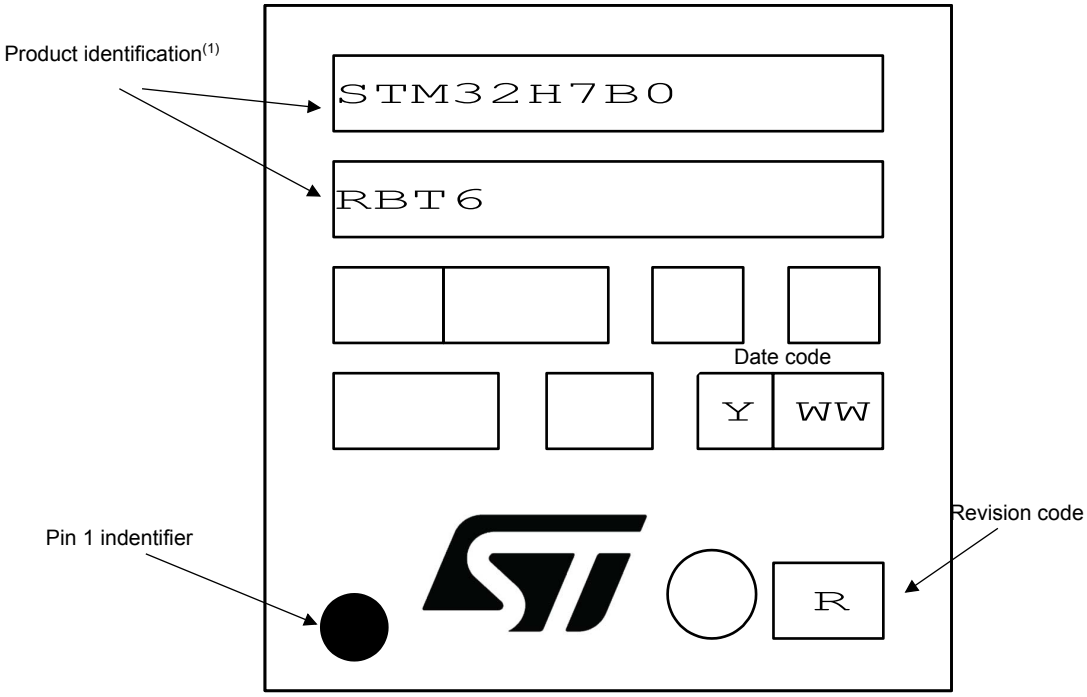
7.1.1 Device marking for LQFP64

The following figure gives an example of topside marking versus pin 1 position identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

Figure 75. **LQFP64 marking example (package top view)**



1. Parts marked as “ES”, “E” or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST’s Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.2 LQFP100 package information

This LQFP is a 100 pins, 14 x 14 mm, low-profile quad flat package.

Note: See list of notes in the notes section.

Figure 76. LQFP100 - Outline(15.)

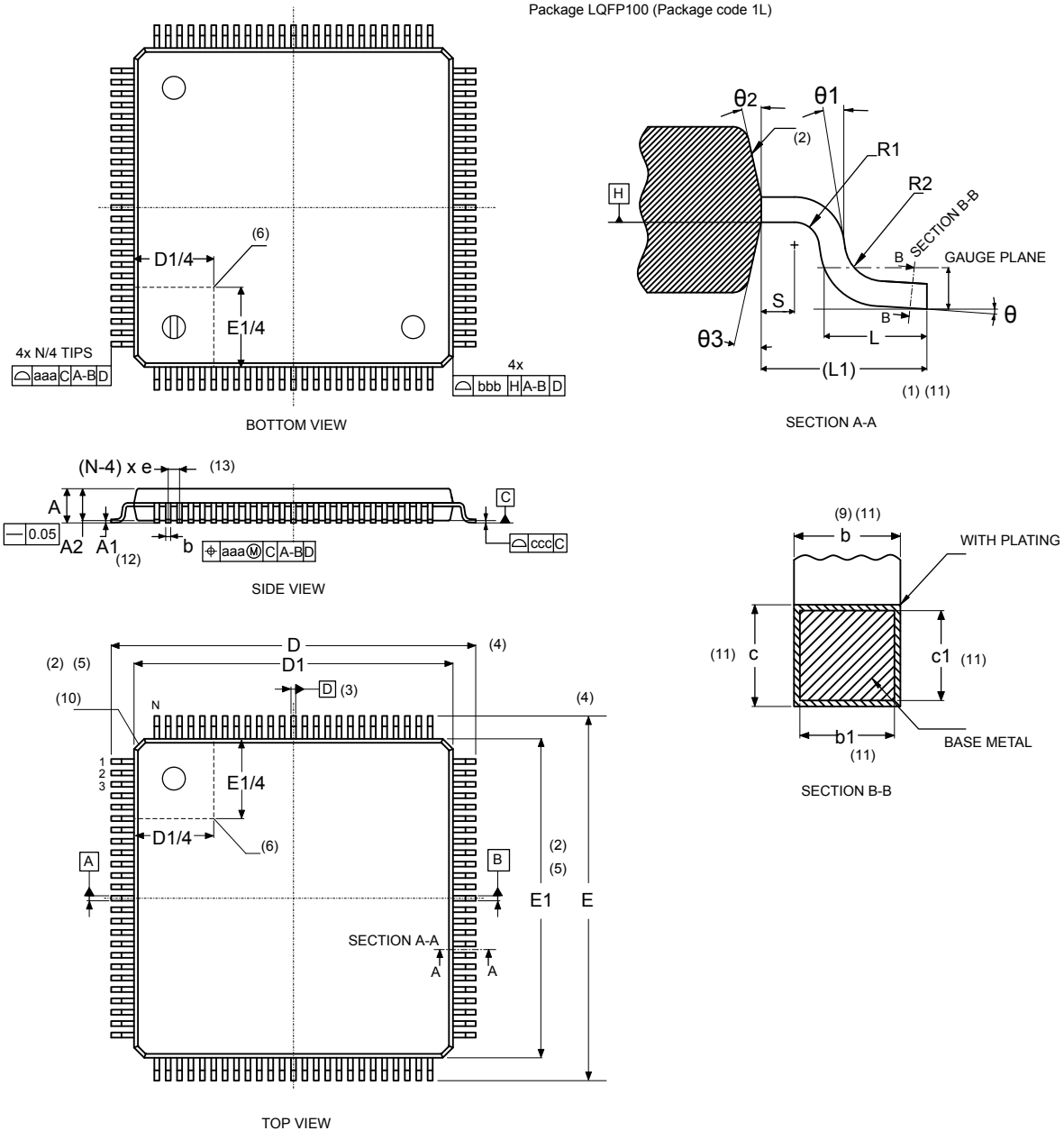


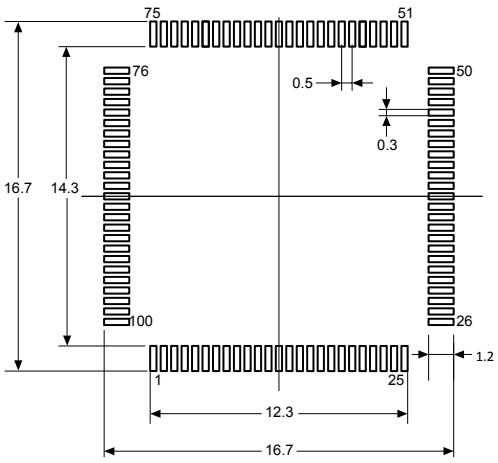
Table 127. **LQFP100 - Mechanical data**

Symbol	millimeters			inches[14.]		
	Min	Typ	Max	Min	Typ	Max
A	-	1.50	1.60	-	0.0590	0.0630
A1[12.]	0.05	-	0.15	0.0020	-	0.0059
A2	1.35	1.40	1.45	0.0531	0.0551	0.0571
b[9.][11.]	0.17	0.22	0.27	0.0067	0.0087	0.0106
b1[11.]	0.17	0.20	0.23	0.0067	0.0079	0.0090
c[11.]	0.09	-	0.20	0.0035	-	0.0079
c1[11.]	0.09	-	0.16	0.0035	-	0.0063
D[4.]	16.00 BSC			0.6299 BSC		
D1[2.][5.]	14.00 BSC			0.5512 BSC		
E[4.]	16.00 BSC			0.6299 BSC		
E1[2.][5.]	14.00 BSC			0.5512 BSC		
e	0.50 BSC			0.0197 BSC		
L	0.45	0.60	0.75	0.0177	0.0236	0.0295
L1[1.][11.]	-	1.00	-	-	0.0394	-
N[13.]	100					
Θ	0°	3.5°	7°	0°	3.5°	7°
Θ1	0°	-	-	0°	-	-
Θ2	10°	12°	14°	10°	12°	14°
Θ3	10°	12°	14°	10°	12°	14°
R1	0.08	-	-	0.0031	-	-
R2	0.08	-	0.20	0.0031	-	0.0079
S	0.20	-	-	0.0079	-	-
aaa[1.]	0.20			0.0079		
bbb[1.]	0.20			0.0079		
ccc[1.]	0.08			0.0031		
ddd[1.]	0.08			0.0031		

Notes

1. Dimensioning and tolerancing schemes conform to ASME Y14.5M-1994.
2. The top package body size may be smaller than the bottom package size by as much as 0.15 mm.
3. Datums A-B and D to be determined at datum plane H.
4. To be determined at seating datum plane C.
5. Dimensions D1and E1 do not include mold flash or protrusions. Allowable mold flash or protrusions is “0.25 mm” per side. D1 and E1 are Maximum plastic body size dimensions including mold mismatch.
6. Details of pin 1 identifier are optional but must be located within the zone indicated.
7. All dimensions are in millimeters.
8. No intrusion allowed inwards the leads.
9. Dimension “b” does not include dambar protrusion. Allowable dambar protrusion shall not cause the lead width to exceed the maximum “b” dimension by more than 0.08 mm. Dambar cannot be located on the lower radius or the foot. Minimum space between protrusion and an adjacent lead is 0.07 mm for 0.4 mm and 0.5 mm pitch packages.
10. Exact shape of each corner is optional.
11. These dimensions apply to the flat section of the lead between 0.10 mm and 0.25 mm from the lead tip.
12. A1 is defined as the distance from the seating plane to the lowest point on the package body.
13. “N” is the number of terminal positions for the specified body size.
14. Values in inches are converted from mm and rounded to 4 decimal digits.
15. Drawing is not to scale.

Figure 77. LQFP100 - Recommended footprint



1. Dimensions are expressed in millimeters.

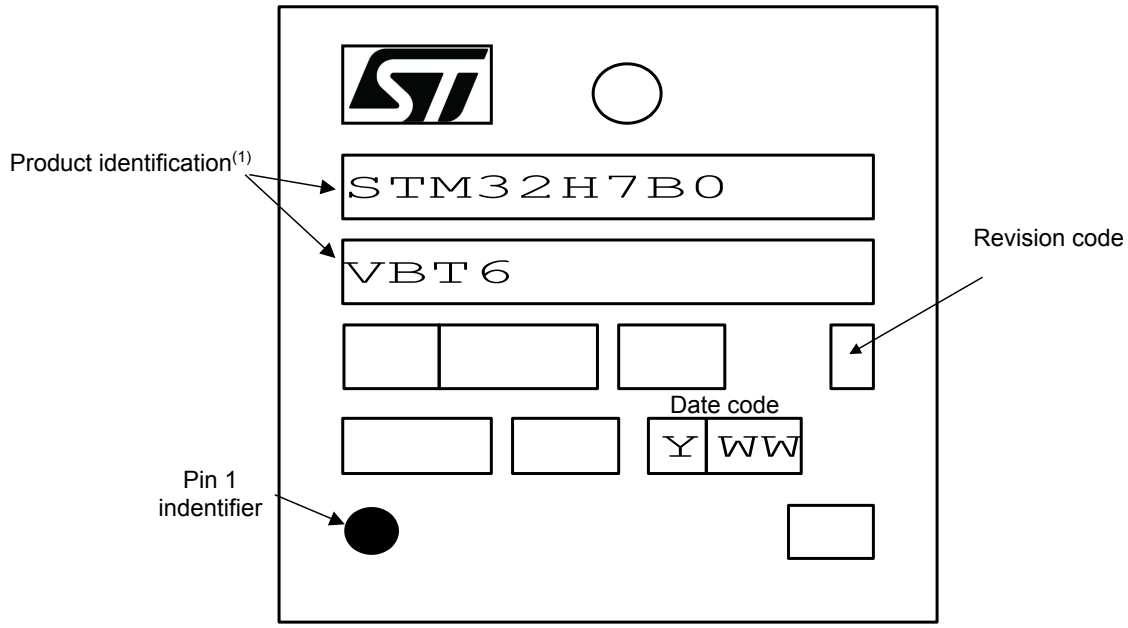
7.2.1 Device marking for LQFP100

The following figure gives an example of topside marking versus pin 1 position identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

Figure 78. LQFP100 marking example (package top view)



1. Parts marked as “ES”, “E” or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST’s Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.3 LQFP144 package information

LQFP144 is a 144-pin, 20 x 20 mm low-profile quad flat package.

Note: See list of notes in the notes section.

Figure 79. LQFP144 - Outline(15.)

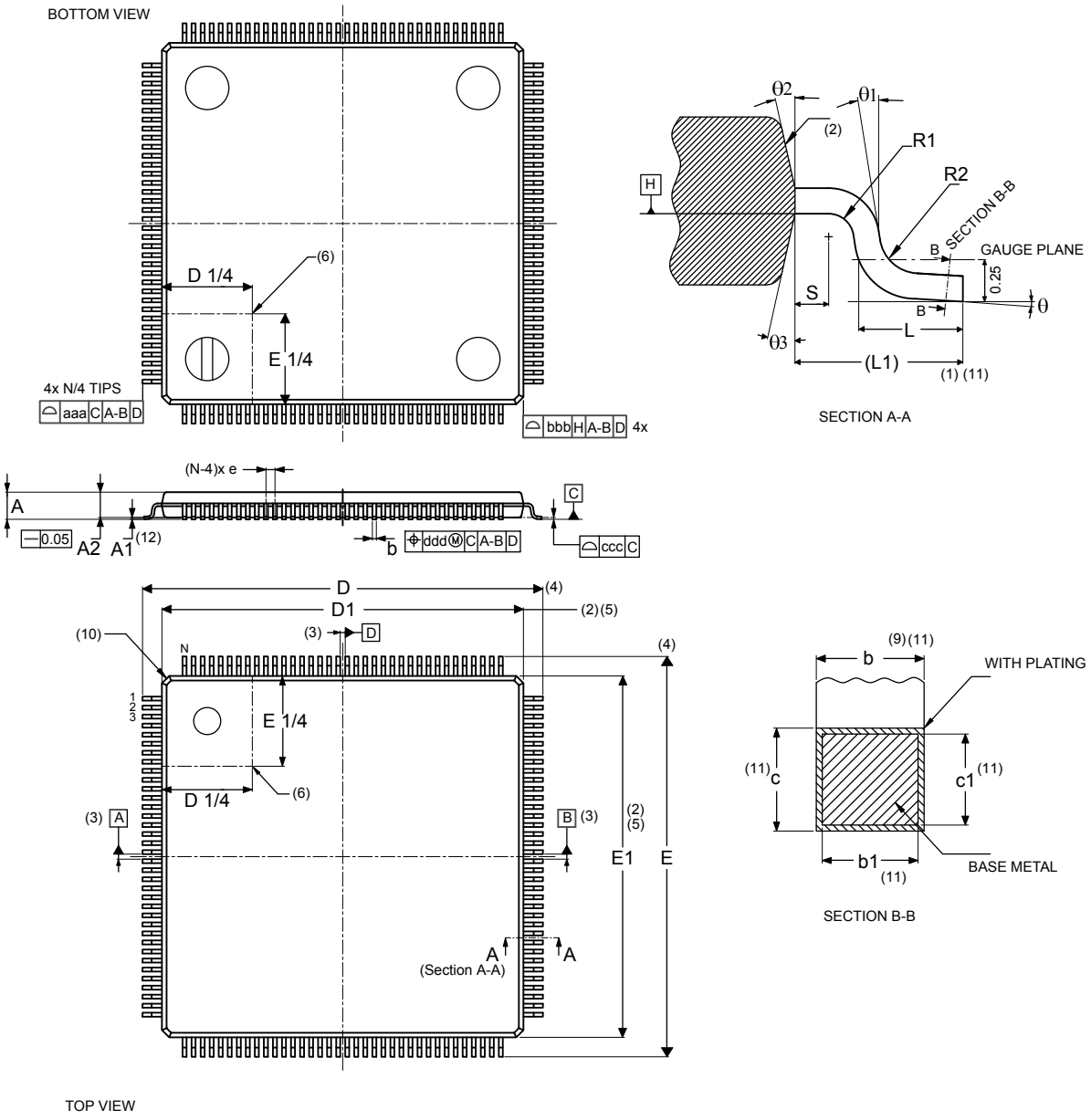


Table 128. **LQFP144 - Mechanical data**

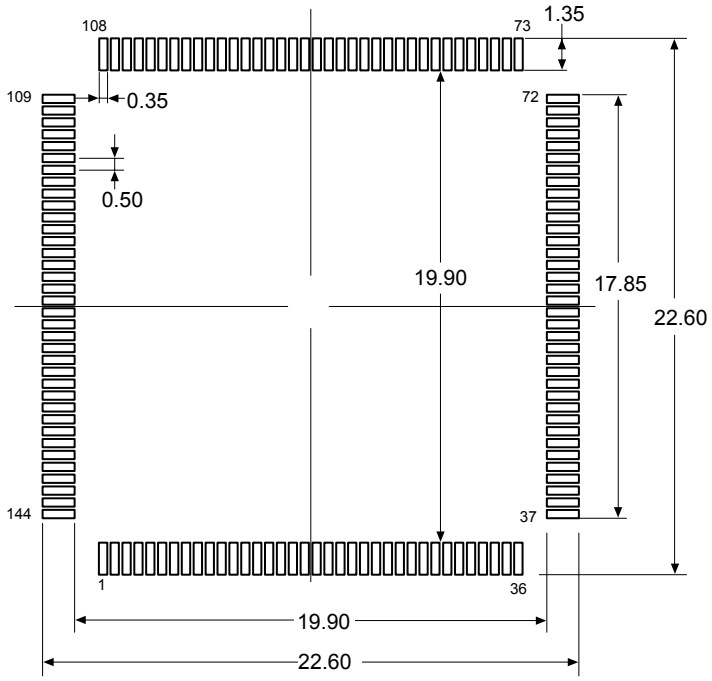
Symbol	millimeters			inches[14.]		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.60	-	-	0.0630
A1[12.]	0.05	-	0.15	0.0020	-	0.0059
A2	1.35	1.40	1.45	0.0531	0.0551	0.0571
b[9.][11.]	0.17	0.22	0.27	0.0067	0.0087	0.0106
b1[11.]	0.17	0.20	0.23	0.0067	0.0079	0.0090
c[11.]	0.09	-	0.20	0.0035	-	0.0079
c1[11.]	0.09	-	0.16	0.0035	-	0.0063
D[4.]	22.00 BSC			0.8661 BSC		
D1[2.][5.]	20.00 BSC			0.7874 BSC		
E[4.]	22.00 BSC			0.8661 BSC		
E1[2.][5.]	20.00 BSC			0.7874 BSC		
e	0.50 BSC			0.0197 BSC		
L	0.45	0.60	0.75	0.0177	0.0236	0.0295
L1	1.00 REF			0.0394 REF		
N[13.]	144					
Θ	0°	3.5°	7°	0°	3.5°	7°
Θ1	0°	-	-	0°	-	-
Θ2	10°	12°	14°	10°	12°	14°
Θ3	10°	12°	14°	10°	12°	14°
R1	0.08	-	-	0.0031	-	-
R2	0.08	-	0.20	0.0031	-	0.0079
S	0.20	-	-	0.0079	-	-
aaa	0.20			0.0079		
bbb	0.20			0.0079		
ccc	0.08			0.0031		
ddd	0.08			0.0031		

Notes:

1. Dimensioning and tolerancing schemes conform to ASME Y14.5M-1994.
2. The top package body size may be smaller than the bottom package size by as much as 0.15 mm.
3. Datums A-B and D to be determined at datum plane H.
4. To be determined at seating datum plane C.
5. Dimensions D1and E1 do not include mold flash or protrusions. Allowable mold flash or protrusions is “0.25 mm” per side. D1 and E1 are Maximum plastic body size dimensions including mold mismatch.
6. Details of pin 1 identifier are optional but must be located within the zone indicated.
7. All dimensions are in millimeters.
8. No intrusion allowed inwards the leads.
9. Dimension “b” does not include dambar protrusion. Allowable dambar protrusion shall not cause the lead width to exceed the maximum “b” dimension by more than 0.08 mm. Dambar cannot be located on the lower radius or the foot. Minimum space between protrusion and an adjacent lead is 0.07 mm for 0.4 mm and 0.5 mm pitch packages.

10. Exact shape of each corner is optional.
11. These dimensions apply to the flat section of the lead between 0.10 mm and 0.25 mm from the lead tip.
12. A1 is defined as the distance from the seating plane to the lowest point on the package body.
13. “N” is the number of terminal positions for the specified body size.
14. Values in inches are converted from mm and rounded to 4 decimal digits.
15. Drawing is not to scale.

Figure 80. LQFP144 - Recommended footprint

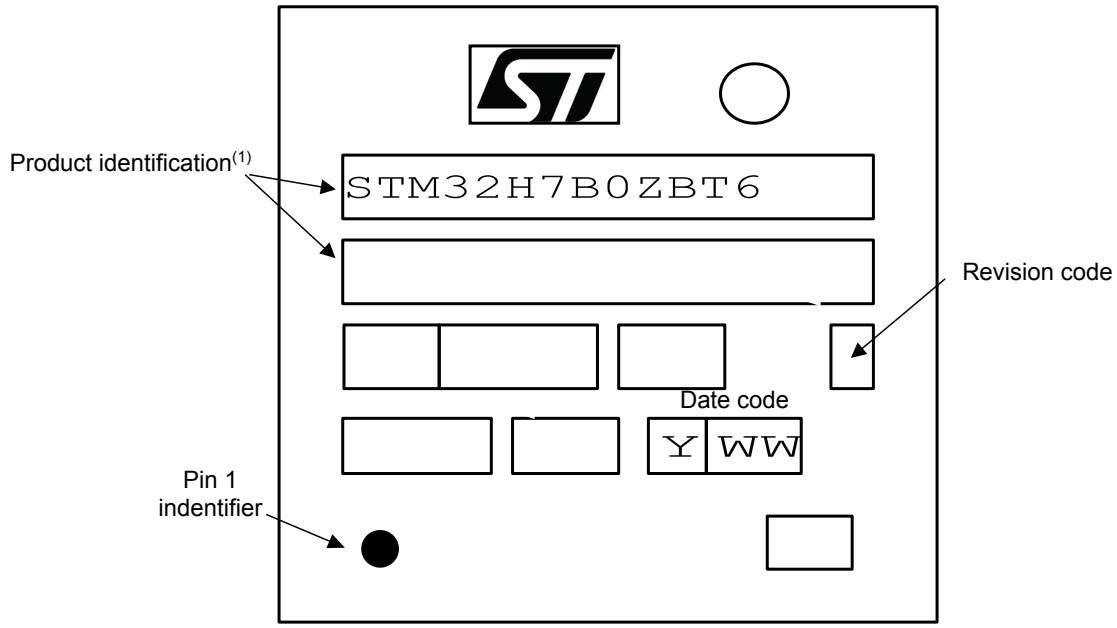


1. Dimensions are expressed in millimeters.

7.3.1 Device marking for LQFP144

The following figure gives an example of topside marking versus pin 1 position identifier location.
The printed markings may differ depending on the supply chain.
Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

Figure 81. LQFP144 marking example (package top view)



1. Parts marked as “ES”, “E” or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST’s Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.4 LQFP176 package information

This LQFP is a 176-pin, 24 x 24 mm, 0.5 mm pitch, low profile quad flat package.

Note: See list of notes in the notes section.

Figure 82. LQFP176 - Outline(15.)

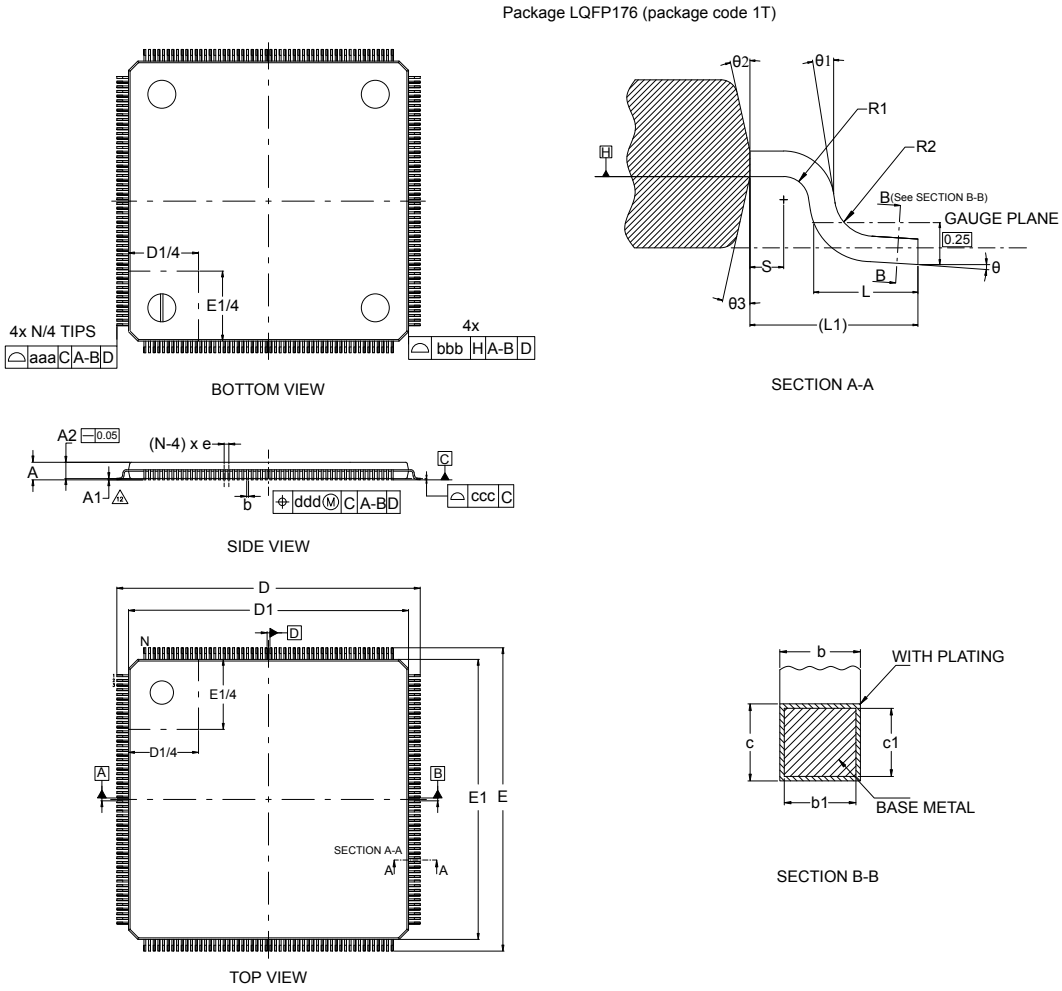


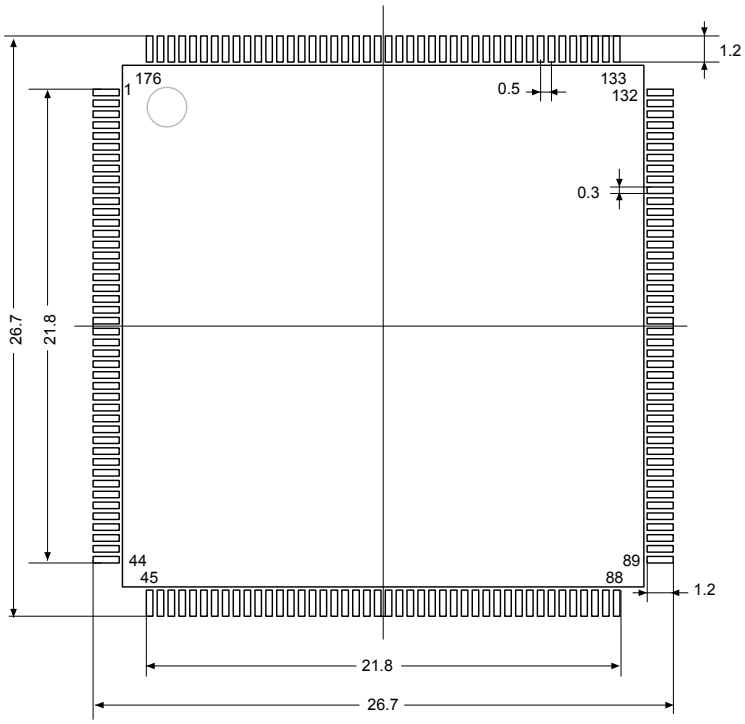
Table 129. **LQFP176 - Mechanical data**

Symbol	millimeters			inches[14.]		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.600	-	-	0.0630
A1[12.]	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b[9.][11.]	0.170	0.220	0.270	0.0067	0.0087	0.0106
b1[11.]	0.170	0.200	0.230	0.0067	0.0079	0.0091
c[11.]	0.090	-	0.200	0.0035	-	0.0079
c1[11.]	0.090	-	0.160	0.0035	-	0.063
D[4.]	26.000			1.0236		
D1[2.][5.]	24.000			0.9449		
E[4.]	26.000			0.0197		
E1[2.][5.]	24.000			0.9449		
e	0.500			0.1970		
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1[1.][11.]	1 REF			0.0394 REF		
N[13.]	176					
θ	0°	3.5°	7°	0°	3.5°	7°
θ1	0°	-	-	0°	-	-
θ2	10°	12°	14°	10°	12°	14°
θ3	10°	12°	14°	10°	12°	14°
R1	0.080	-	-	0.0031	-	-
R2	0.080	-	0.200	0.0031	-	0.0079
S	0.200	-	-	0.0079	-	-
aaa[1.]	0.200			0.0079		
bbb[1.]	0.200			0.0079		
ccc[1.]	0.080			0.0031		
ddd[1.]	0.080			0.0031		

Notes

1. Dimensioning and tolerancing schemes conform to ASME Y14.5M-1994.
2. The top package body size may be smaller than the bottom package size by as much as 0.15 mm.
3. Datums A-B and D to be determined at datum plane H.
4. To be determined at seating datum plane C.
5. Dimensions D1and E1 do not include mold flash or protrusions. Allowable mold flash or protrusions is “0.25 mm” per side. D1 and E1 are Maximum plastic body size dimensions including mold mismatch.
6. Details of pin 1 identifier are optional but must be located within the zone indicated.
7. All dimensions are in millimeters.
8. No intrusion allowed inwards the leads.
9. Dimension “b” does not include dambar protrusion. Allowable dambar protrusion shall not cause the lead width to exceed the maximum “b” dimension by more than 0.08 mm. Dambar cannot be located on the lower radius or the foot. Minimum space between protrusion and an adjacent lead is 0.07 mm for 0.4 mm and 0.5 mm pitch packages.
10. Exact shape of each corner is optional.
11. These dimensions apply to the flat section of the lead between 0.10 mm and 0.25 mm from the lead tip.
12. A1 is defined as the distance from the seating plane to the lowest point on the package body.
13. “N” is the number of terminal positions for the specified body size.
14. Values in inches are converted from mm and rounded to 4 decimal digits.
15. Drawing is not to scale.

Figure 83. LQFP176 - Recommended footprint



Note: Dimensions are expressed in millimeters.

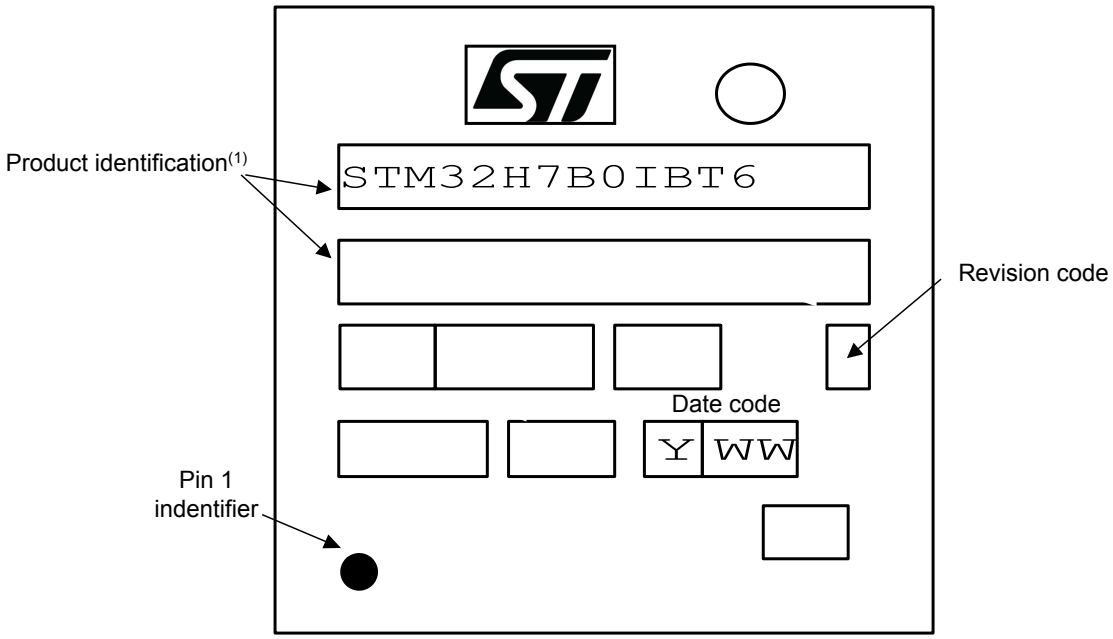
7.4.1 Device marking for LQFP176

The following figure gives an example of topside marking versus pin 1 position identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

Figure 84. **LQFP176 marking example (package top view)**

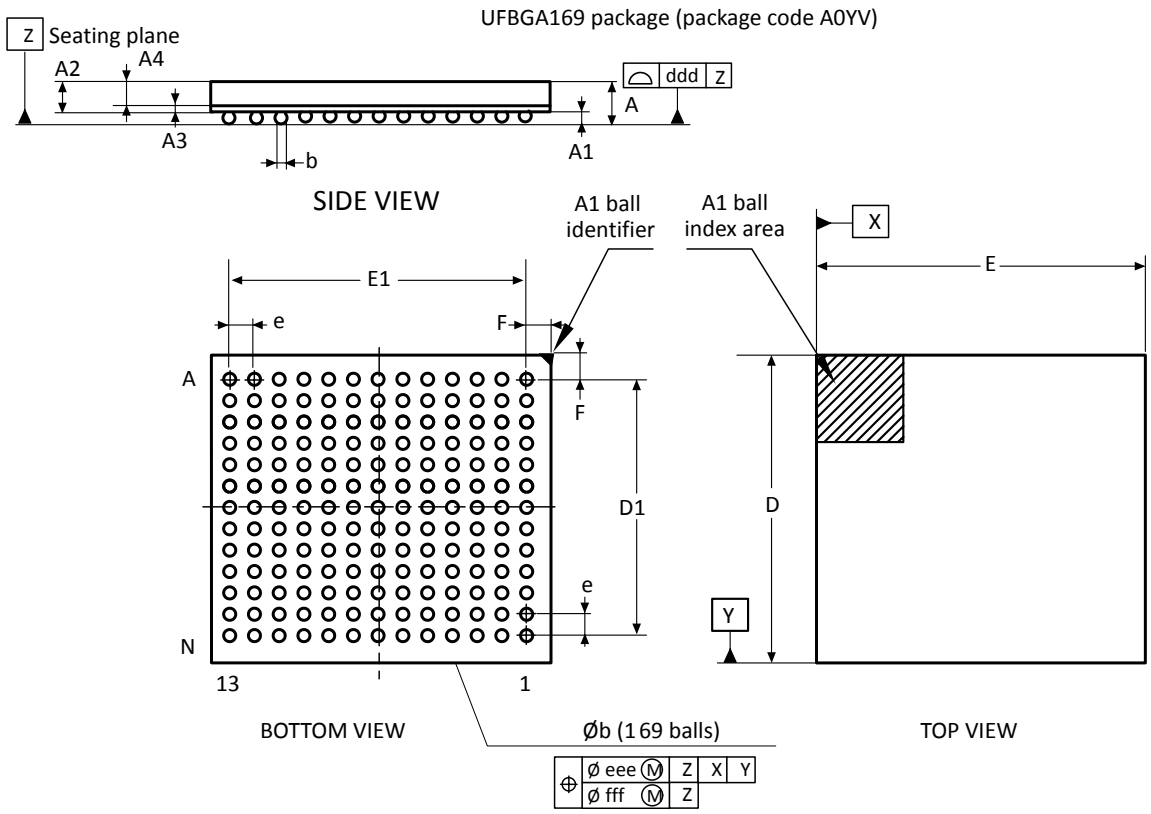


1. Parts marked as “ES”, “E” or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST’s Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.5 UFBGA169 package information

This UFBGA is a 169 balls, 7 x 7 mm, 0.50 mm pitch, ultra thin profile fine pitch ball grid array package

Figure 85. UFBGA169 - Outline



1. Drawing is not to scale.
2. – The terminal A1 corner must be identified on the top surface by using a corner chamfer, ink or metalized markings, or other feature of package body or integral heat slug.
   – A distinguishing feature is allowable on the bottom surface of the package to identify the terminal A1 corner. Exact shape of each corner is optional.

Table 130. UFBGA169 - Mechanical data

Symbol	millimeters			inches [1]		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A[2]	-	-	0.600	-	-	0.0236
A1	-	-	0.110	-	-	0.0043
A2	-	0.130	-	-	0.0051	-
A4	-	0.320	-	-	0.0126	-
b[3]	0.230	0.280	0.330	0.0091	0.0110	0.0130
D	6.850	7.000	7.150	0.2697	0.2756	0.2815
D1	-	6.000	-	-	0.2362	-
E	6.850	7.000	7.150	0.2697	0.2756	0.2815
E1	-	6.000	-	-	0.2362	-
e	-	0.500	-	-	0.0197	-
F	-	0.500	-	-	0.0197	-
ddd	-	-	0.080	-	-	0.0031
eee [4]	-	-	0.015	-	-	0.0059
fff[5]	-	-	0.050	-	-	0.0020

1. *Values in inches are converted from mm and rounded to four decimal digits.*
2. *Ultra Thin profile: 0.50 < A ≤ 0.65 mm / Fine pitch: e < 1.00 mm pitch.*
   - *The total profile height (dim A) is measured from the seating plane to the top of the component*
   - *The maximum total package height is calculated by the following methodology:*
   - *A Max = A1 Typ + A2 Typ + A4 Typ + √ (A1²+A2²+A4² tolerance values)*
3. *The typical balls diameters before mounting is 0.20 mm.*
4. *The tolerance of position that controls the location of the pattern of balls with respect to datum A and B. For each ball there is a cylindrical tolerance zone eee perpendicular to datum C and located on true position with respect to datum A and B as defined by e. The axis perpendicular to datum C of each ball must lie within this tolerance zone.*
5. *The tolerance of position that controls the location of the balls within the matrix with respect to each other. For each ball there is a cylindrical tolerance zone fff perpendicular to datum C and located on true position as defined by e. The axis perpendicular to datum C of each ball must lie within this tolerance zone. Each tolerance zone fff in the array is contained entirely in the respective zone eee above The axis of each ball must lie simultaneously in both tolerance zones.*

Figure 86. UFBGA169 - Recommended footprint

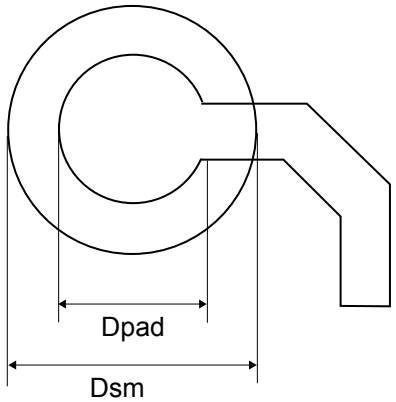


Table 131. **UFBGA169 - recommended PCB design rules (0.5 mm pitch BGA)**

Dimension	Recommended values
Pitch	0.5 mm
Dpad	0.27 mm
Dsm	0.35 mm typ. (depends on the soldermask registration tolerance)
Solder paste	0.27 mm aperture diameter

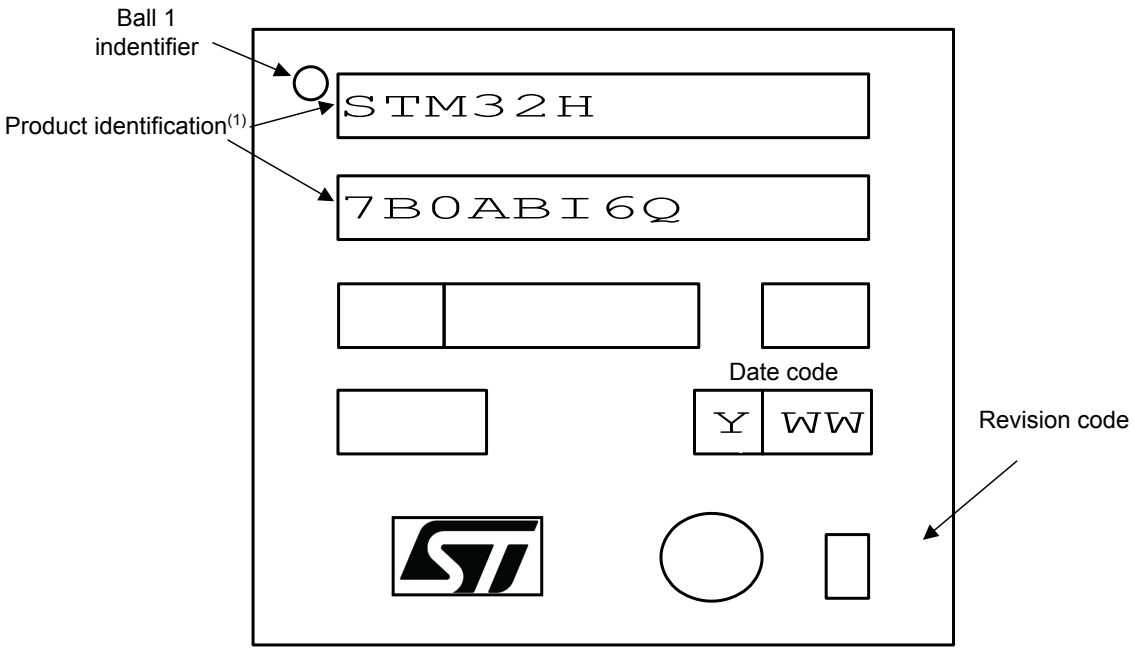
7.5.1 Device marking for UFBGA169

The following figure gives an example of topside marking versus pin 1 position identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

Figure 87. **UFBGA169 marking example (package top view)**

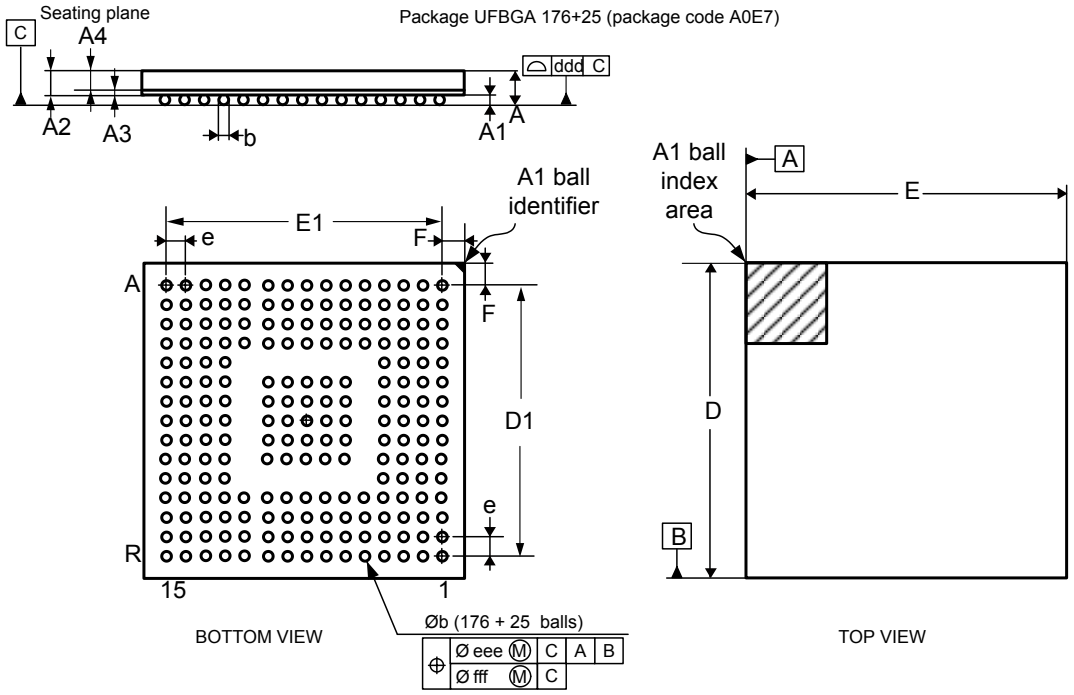


1. Parts marked as “ES”, “E” or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST’s Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.6 UFBGA(176+25) package information

This UFBGA is a 176+25 balls, 10 x 10 mm, 0.65 mm pitch, ultra thin profile fine pitch ball grid array package.

Figure 88. UFBGA(176+25) - Outline



1. Drawing is not to scale.
2. The terminal A1 corner must be identified on the top surface by using a corner chamfer, ink or metalized markings, or other feature of package body or integral heat slug. A distinguishing feature is allowable on the bottom surface of the package to identify the terminal A1 corner. Exact shape of each corner is optional.

Table 132. UFBGA(176+25) - Mechanical data

Symbol	millimeters			inches [1]		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A[2]	-	-	0.60	-	-	0.0236
A1	0.05	0.08	0.11	0.0020	0.0031	0.0043
A2	-	0.45	-	-	0.0177	-
A3	-	0.13	-	-	0.0051	-
A4	-	0.32	-	-	0.0126	-
b	0.24	0.29	0.34	0.0094	0.0114	0.0134
D	9.85	10.00	10.15	0.3878	0.03937	0.3996
D1	-	9.10	-	-	0.3583	-
E	9.85	10.00	10.15	0.3878	0.03937	0.3996
E1	-	9.10	-	-	0.3583	-
e	-	0.65	-	-	0.0256	-
F	-	0.45	-	-	0.0177	-
ddd	-	-	0.08	-	-	0.0031
eee [3]	-	-	0.15	-	-	0.0059
fff[4]	-	-	0.05	-	-	0.0020

1. *Values in inches are converted from mm and rounded to four decimal digits.*
2. *Ultra thin profile: 0.50 < A Max \u0001 0.65 mm / Fine pitch: e < 1.00 mm. The total profile height (Dim.A) is measured from the seating plane “C” to the top of the component.*
3. *The tolerance of position that controls the location of the pattern of balls with respect to datum A and B. For each ball there is a cylindrical tolerance zone eee perpendicular to datum C and located on true position with respect to datum A and B as defined by e. The axis perpendicular to datum C of each ball must lie within this tolerance zone.*
4. *The tolerance of position that controls the location of the balls within the matrix with respect to each other. For each ball there is a cylindrical tolerance zone fff perpendicular to datum C and located on true position as defined by e. The axis perpendicular to datum C of each ball must lie within this tolerance zone. Each tolerance zone fff in the array is contained entirely in the respective zone eee above The axis of each ball must lie simultaneously in both tolerance zones.*

Figure 89. UFBGA(176+25) - Recommended footprint

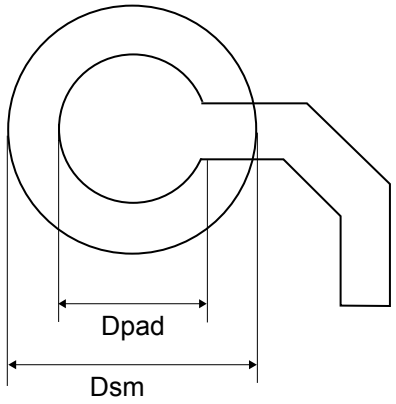


Table 133. UFBGA(176+25) - Recommended PCB design rules (0.65 mm pitch BGA)

Dimension	Recommended values
Pitch	0.65 mm
Dpad	0.300 mm
Dsm	0.400 mm typ. (depends on the soldermask registration tolerance)
Stencil opening	0.300 mm aperture diameter
Stencil thickness	Between 0.100 mm and 0.125 mm
Pad trace width	0.100 mm

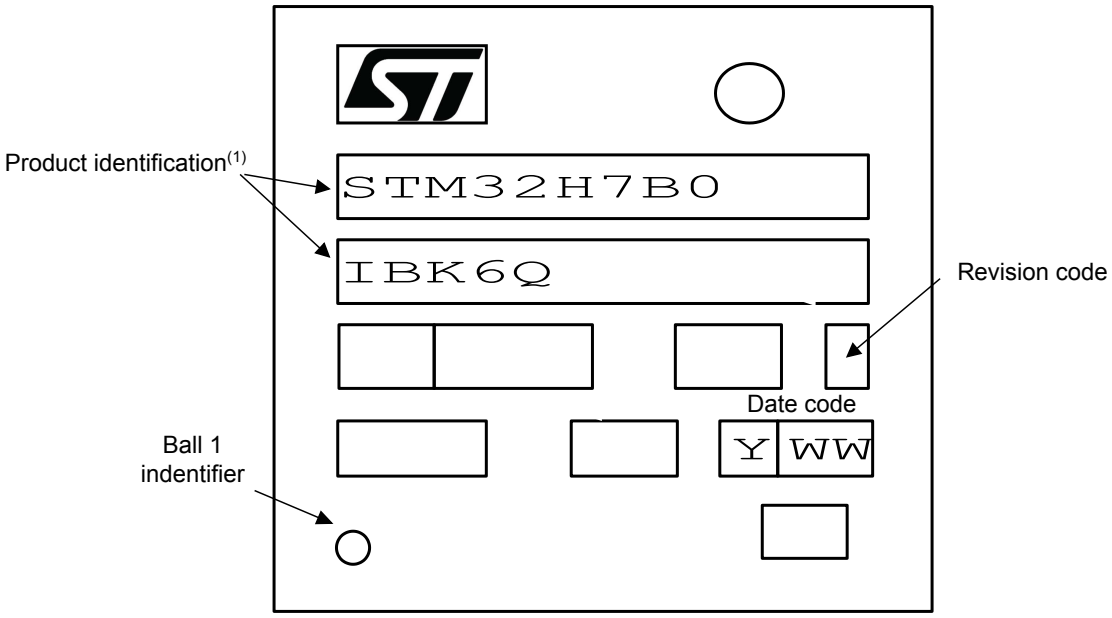
7.6.1 Device marking for UFBGA176+25

The following figure gives an example of topside marking versus pin 1 position identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

Figure 90. UFBGA176+25 marking example (package top view)



1. Parts marked as “ES”, “E” or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST’s Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.7 Thermal characteristics

The maximum chip-junction temperature, TJ max, in degrees Celsius, may be calculated using the following equation:

$T_J max = T_A max + (P_D max \times \Theta_{JA})$

Where:

- $T_A$max is the maximum ambient temperature in °C,
- Θ_{JA} is the package junction-to-ambient thermal resistance, in °C/W,
- $P_D$max is the sum of P_{INT}max and $P_{I/O}$max ($P_D$max = P_{INT}max + $P_{I/O}$max),
- P_{INT}max is the product of I_{DD} and V_{DD}, expressed in Watts. This is the maximum chip internal power.

$P_{I/O}$max represents the maximum power dissipation on output pins where:

$P_{I/O}$max = Σ (V_{OL} × I_{OL}) + Σ((V_{DD} – V_{OH}) × I_{OH}),

taking into account the actual V_{OL} / I_{OL} and V_{OH} / I_{OH} of the I/Os at low and high level in the application.

Table 134. Thermal characteristics

Symbol	Definition	Parameter	value	unit
Θ_{JA}	Thermal resistance junction-ambiant	Thermal resistance junction-ambient LQFP64 - 10 x 10 mm /0.5 mm pitch	48.8	°C/W
		Thermal resistance junction-ambient LQFP100 - 14 x 14 mm /0.5 mm pitch	47.4	
		Thermal resistance junction-ambient LQFP144 - 20 x 20 mm /0.5 mm pitch	46	
		Thermal resistance junction-ambient LQFP176 - 24 x 24 mm /0.5 mm pitch	43.6	
		Thermal resistance junction-ambient UFBGA169 - 7 x 7 mm /0.5 mm pitch	41.4	
		Thermal resistance junction-ambient UFBGA176+25 - 10 x 10 mm /0.65 mm pitch	44.4	
Θ_{JB}	Thermal resistance junction-board	Thermal resistance junction-board LQFP64 - 10 x 10 mm /0.5 mm pitch	37.2	°C/W
		Thermal resistance junction-board LQFP100 - 14 x 14 mm /0.5 mm pitch	39.2	
		Thermal resistance junction-board LQFP144 - 20 x 20 mm /0.5 mm pitch	41.3	
		Thermal resistance junction-board LQFP176 - 24 x 24 mm /0.5 mm pitch	40.2	
		Thermal resistance junction-board UFBGA169 - 7 x 7 mm /0.5 mm pitch	15.3	
		Thermal resistance junction-board UFBGA176+25 - 10 x 10 mm /0.65 mm pitch	25	
Θ_{JC}	Thermal resistance junction-case	Thermal resistance junction-case LQFP64 - 10 x 10 mm /0.5 mm pitch	13	°C/W
		Thermal resistance junction-case LQFP100 - 14 x 14 mm /0.5 mm pitch	12.8	
		Thermal resistance junction-case LQFP144 - 20 x 20 mm /0.5 mm pitch	12.6	
		Thermal resistance junction-case LQFP176 - 24 x 24 mm /0.5 mm pitch	11.5	
		Thermal resistance junction-case UFBGA169 - 7 x 7 mm /0.5 mm pitch	19.9	
		Thermal resistance junction-case UFBGA176+25 - 10 x 10 mm /0.65 mm pitch	18.9	

7.7.1 Reference documents

- JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from www.jedec.org.
- For information on thermal management, refer to application note “Thermal management guidelines for STM32 32-bit Arm Cortex MCUs applications” (AN5036) available from *www.st.com*.

8 Ordering information

Example: STM32 H 7B0 Z B T 6 Q TR

Device family

STM32 = Arm-based 32-bit microcontroller

Product type

H = High performance

Device subfamily

7B0 = STM32H7B0 Value line with cryptographic accelerator

Pin count

R = 64 pins

V = 100 pins/balls

Z = 144 pins

A = 169 balls

I = 176 or 176 + 25 pins/balls

Flash memory size

B = 128 Kbytes

Package

T = LQFP ECOPACK2

K = UFBGA 0.65 mm pitch ECOPACK2

I = UFBGA 0.5 mm pitch ECOPACK2

Temperature range

6 = Industrial temperature range, –40 to 85 °C

Option

Q = with SMPS

Blank = without SMPS

Packing

TR = tape and reel

No character = tray or tube

For a list of available options (such as speed and package) or for further information on any aspect of this device, contact your nearest ST sales office.

Important security notice

The STMicroelectronics group of companies (ST) places a high value on product security, which is why the ST product(s) identified in this documentation may be certified by various security certification bodies and/or may implement our own security measures as set forth herein. However, no level of security certification and/or built-in security measures can guarantee that ST products are resistant to all forms of attacks. As such, it is the responsibility of each of ST's customers to determine if the level of security provided in an ST product meets the customer needs both in relation to the ST product alone, as well as when combined with other components and/or software for the customer end product or application. In particular, take note that:

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Revision history

Table 135. **Document revision history**

Date	Revision	Changes
20-Dec-2019	1	Initial release.
27-Apr-2020	2	Updated Octo-SPI interface in Table 1. STM32H7B0xB features and peripheral counts. Updated Figure 2. Power-up/power-down sequence in Section 6.1.6 Power supply scheme. Updated HSLV feature description in Section 3.8 General-purpose input/outputs (GPIOs). In Section 5 Pin descriptions: updated Table 6. Legend/abbreviations used in the pinout table; changed SPDIFRX into SPDIFRX1 and updated all SPDIFRX1 pin names. Updated Table 17. Voltage characteristics to add V_{REF+} in the list of external main supply voltage. Removed clock frequencies from Table 20. General operating conditions and added new Section 6.3.1 . Changed condition for $t_{RSTTEMPO}$ in Table 27. Reset and power control block characteristics. Added $I_{DD50USB}$ in Table 30. USB regulator characteristics. Updated Table 38. Typical current consumption in System Stop mode, added Table 39. Typical current consumption RAM shutoff in Stop mode, and added IWDG and changed SPDIFRX into SPDIFRX1 in Table 42. Peripheral current consumption in Run mode. Table 56. Flash memory programming: updated t_{ME} description and unit. In the whole Section 6.3.18 FMC characteristics, replaced sentence "the T_{KERCK} is the fmc_ker_ck clock period" by "the $T_{fmc_ker_ck}$ is the kernel clock period". Section 6.3.19 Octo-SPI interface characteristics: added parameter measurement conditions, updated Table 88. OCTOSPI characteristics in SDR mode and Table 89. OCTOSPI characteristics in DTR mode (with DQS)/Octal and Hyperbus, updated Figure 42. OctoSPI Hyperbus clock, Figure 43. OctoSPI Hyperbus read and Figure 44. OctoSPI Hyperbus write. Updated Figure 48. Power supply and reference decoupling (V_{REF+} not connected to V_{DDA}), note 1. and note 1.. Section 6.3.30 Digital filter for Sigma-Delta Modulators (DFSDM) characteristics, Section 6.3.31 Camera interface (DCMI) timing specificationsSection 6.3.33 LCD-TFT controller (LTDC) characteristics, Section 6.3.36.2 USART interface characteristics, Section 6.3.36.3 SPI interface characteristics, Section 6.3.36.4 I2S Interface characteristics, Section 6.3.36.5 SAI characteristics, Section 6.3.36.7 SD/SDIO MMC card host interface (SDMMC) characteristics, Section 6.3.36.8 USB OTG_FS characteristics, Section 6.3.36.9 USB OTG_HS characteristics, Section 6.3.36.10 JTAG/SWD interface characteristics: changed VOS level to VOS0 in the parameter measurement conditions.
08-Jul-2020	3	Updated note related to ULPI interface availability on packages that do not feature PC2 and PC3 I/Os in Table 1. STM32H7B0xB features and peripheral counts. Updated Table 18. Current characteristics, Table 19. Thermal characteristics and Figure 13. Current consumption measurement scheme. Updated Figure 12. Power supply scheme. Added note to V_{REFINT} in Table 28. Embedded reference voltage.Added Table 31. Inrush current and inrush electric charge characteristics for LDO and SMPS. Updated Table 44. Low-power mode wakeup timings, Table 32. Typical and maximum current consumption in Run mode, code with data processing running from ITCM, regulator ON and Table 33. Typical and maximum current consumption in Run mode, code with data processing running from flash memory, cache ON.

Date	Revision	Changes
		Updated Table 66. Output timing characteristics (HSLV OFF) and Table 67. Output timing characteristics (HSLV ON). Updated Table 60. ESD absolute maximum ratings. Added notes related to performance degradation at VOS1 in Section 6.3.18 FMC characteristics, Section 6.3.19 Octo-SPI interface characteristics, Section 6.3.32 PSSI interface characteristics, Section 6.3.33 LCD-TFT controller (LTDC) characteristics, Section 6.3.36.2 USART interface characteristics, Section 6.3.36.3 SPI interface characteristics, Section 6.3.36.4 I2S Interface characteristics, Section 6.3.36.5 SAI characteristics, Section 6.3.36.6 MDIO characteristics, Section 6.3.36.7 SD/SDIO MMC card host interface (SDMMC) characteristics, Section 6.3.36.9 USB OTG_HS characteristics and Section 6.3.36.10 JTAG/SWD interface characteristics. Updated $F_{(CLK)}$ measurement conditions in Table 88. OCTOSPI characteristics in SDR mode and Table 89. OCTOSPI characteristics in DTR mode (with DQS)/Octal and Hyperbus. Added Figure 45. ADC conversion timing diagram. Added Section 6.3.32 PSSI interface characteristics. Updated Figure 57. USART timing diagram in Master mode and Figure 58. USART timing diagram in Slave mode. Added note related to ULPI transceivers operating at 1.8 V in Table 123. Dynamics characteristics: USB ULPI.
16-Sep-2020	4	Updated VDDMMC separate supply pad in Section 2 Description. In Section 3.31 True random number generator (RNG), changed "random number generator" in "true random number generator" and description updated. Updated Figure 12. Power supply scheme and Section 6.3.2 VCAP external capacitor. Added V_{BAT} in Section 6.3.1 General operating conditions. Updated High-speed external user clock generated from an external source and Low-speed external user clock generated from an external source. Added reference to application note AN4899 in Section 6.3.16 I/O port characteristics. Updated DuCyCKOUT in Table 106. DFSDM measured timing 1.62-3.6 V Updated and .
28-Sep-2020	5	Updated V_{HSEL} maximum value in Table 45. High-speed external user clock characteristics.
04-May-2021	6	Added indication that patents apply to the devices in Section Features. Added reference to errata sheet in Section 1 Introduction. Updated WKUP signals in Figure 1. STM32H7B0xB block diagram and in Table 7. STM32H7B0xB pin/ball definition. Updated Section 3.40 Serial peripheral interfaces (SPI)/integrated interchip sound interfaces (I2S). Added note to TRIM parameter in Table 50. HSI oscillator characteristics. Extended Figure 46. ADC accuracy characteristics to both ADC resolutions and updated Figure 47. Typical connection diagram using the ADC with FT/TT pins featuring analog switch function. Replace 110 °C by 130 °C in Table 98. Analog temperature sensor calibration values. Updated $t_{su(ADV-CLKH)}$, $t_{h(CLKH-ADV)}$, $t_{su(NWAIT-CLKH)}$ and $t_{h(CLKH-NWAIT)}$ minimum values in Table 9. Updated $t_{su(DV-CLKH)}$, $t_{h(CLKH-DV)}$, $t_{(NWAIT-CLKH)}$ and $t_{h(CLKH-NWAIT)}$ minimum values in Table 80. Synchronous non-multiplexed NOR/PSRAM read timings. Updated $t_{su(SDCLKH_Data)}$ and $t_{h(SDCLKH_Data)}$ minimum values in Table 84. SDRAM read timings. Updated $t_{su(SDCLKH_Data)}$ and $t_{h(SDCLKH_Data)}$ minimum values in Table 85. LPSDR SDRAM read timings.

Date	Revision	Changes
		Section 6.3.19 Octo-SPI interface characteristics: • Updated $t_{S(IN)}$/$t_{H(IN)}$ conditions and minimum values in Table 88. OCTOSPI characteristics in SDR mode and updated Figure 40. OctoSPI timing diagram - SDR mode. • Updated Table 89. OCTOSPI characteristics in DTR mode (with DQS)/ Octal and Hyperbus and Figure 41. OctoSPI timing diagram - DTR mode. • Updated Figure 42. OctoSPI Hyperbus clock, Figure 43. OctoSPI Hyperbus read and Figure 44. OctoSPI Hyperbus write. Table 116. SPI dynamic characteristics: • Changed $t_{su(MI)}$ and $t_{h(MI)}$ minimum values in Master mode. • Removed $t_{w(SCKH)}$, $t_{w(SCKL)}$. Table 120. Dynamics characteristics: SDMMC characteristics, V_{DD}=2.7 to 3.6 V: • Modified t_{ISU} and t_{IH} minimum values for CMD/ D inputs in High-speed mode. • Modified t_{ISUD} and t_{IH} minimum values for CMD/ D inputs in Default mode. Table 121. Dynamics characteristics: eMMC characteristics VDD=1.71V to 1.9V: • Removed SDIO_CK/f_{PCLK2} frequency ratio. • Modified t_{ISU} and t_{IH} minimum values in CMD, D inputs (referenced to CK) in eMMC mode. Section 7.5 UFBGA169 package information: • Added note 2 below Figure 85. UFBGA169 - Outline. • Updated Table 130. UFBGA169 - Mechanical data. • Removed note related to non-solder mask below Figure 86. UFBGA169 - Recommended footprint. Section 7.6 UFBGA(176+25) package information: • Added note 2 below . • Updated .
19-May-2022	7	Renamed SPIx_NSS signal into SPIx_SS. Updated Table 18. Current characteristics. Removed $f_{TraceCK}$ / f_{JTCK} from Table 21. Maximum allowed clock frequencies. Updated note 3 in Table 38. Typical current consumption in System Stop mode. Updated Table 64. Output voltage characteristics for all I/Os except PC13, PC14, PC15 and PI8 and Table 65. Output voltage characteristics for PC13, PC14, PC15 and PI8. Added Table 68. Pxy_C and Pxy analog switch characteristics. Added note to $t_{d(SDCLKL- SDNE)}$ maximum value in Table 84. SDRAM read timings, Table 85. LPSDR SDRAM read timings, Table 86. SDRAM write timings and Table 87. LPSDR SDRAM write timings. Updated Figure 47. Typical connection diagram using the ADC with FT/TT pins featuring analog switch function and notes below figures. Added note to t_{SAMP} in Table 94. DAC characteristics. Updated Table 106. DFSDM measured timing 1.62-3.6 V. In Table 115. USART characteristics, replaced $t_{su(SI)}$ and $t_{su(MI)}$ by $t_{su(RX)}$, $t_{h(SI)}$ and $t_{h(MI)}$ by $t_{h(RX)}$, $t_{v(SO)}$ and $t_{v(MO)}$ by $t_{v(TX)}$, and $t_{h(SO)}$ and $t_{h(MO)}$ by $t_{h(TX)}$. Updated Section 7.1 LQFP64 package information, Section 7.2 LQFP100 package information, Section 7.5 UFBGA169 package information and Section 7.6 UFBGA(176+25) package information. Added Section Important security notice.

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